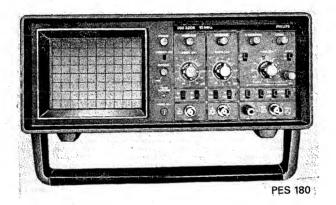
15 MHz Dual trace oscilloscope PM 3206

Service Manual

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GENERAL INFORMATION

Read these pages carefully before installation and use of the instrument.

The following clauses contain information, cautions and warnings which must be followed to ensure safe operation and to retain the instrument in a safe condition. Adjustment, maintenance and repair of the instrument shall be carried out only by qualified personnel.

1.1. Safety Precautions

For the correct and safe use of this instrument it is essential that both operating and servicing personnel follow generally-accepted safety procedures in addition to the safety precautions specified in this manual.

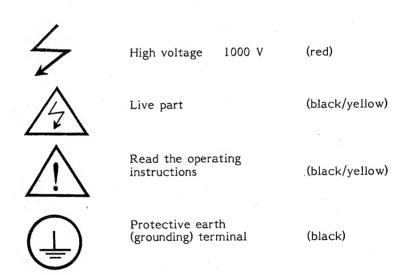
Specific warning and caution statements, where they apply, will be found throughout the manual. Where necessary, the warning and caution statements and/or symbols are marked on the apparatus.

1.2. Caution and Warning Statements

<u>CAUTION</u>: is used to indicate correct operating or maintenance procedures in order to prevent damage to or destruction of the equipment or other property.

<u>WARNING</u>: calls attention to a potential danger that requires correct procedures or practices in order to prevent personal injury.

1.3. Symbols



1.4. Impaired Safety-Protection

Whenever it is likely that safety-protection has been impaired, the instrument <u>must</u> be made inoperative and be secured against any unintended operation. The matter should then be referred to qualified technicians.

Safety protection is likely to be impaired if, for example, the instrument fails to perform the intended measurements or shows visible damage.

1.5. General Clauses

1.5.1. WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts and accessible terminals which can be dangerous to life.

- 1.5.2. The instrument shall be diconnected from all voltage sources before it is opened.
- 1.5.3. Bear in mind that capacitors inside the instrument can hold their charge even if the instrument has been separated from all voltage sources.
- 1.5.4. WARNING: Any interruption of the protective earth conductor inside or outside the instrument, or disconnection of the protective earth terminal, is likely to make the instrument dangerous. Intentional interruption is prohibited.

1.6. **CHARACTERISTICS**

A. Performance Characteristics

Properties expressed in numerical values with stated tolerance are guaranteed by PHILIPS. Specified non-tolerance numerical values indicate those that could be <u>nominally</u> expected from the mean of a range of identical instruments.

This specification is valid after the instrument has warmed up for 15 minutes (reference temperature 23°C).

B. Safety Characteristics

This apparatus has been designed and tested in accordance with Safety Class I requirements of IEC Publication 348, and has been supplied in a safe condition.

C. Initial Characteristics

Overall dimensions:

Height Width (including handle) Depth

Maximum weight (Mass)

Operation position

142 mm

378 mm 348 mm

5 kg

a) horizontally on bottom feet

b) vertically on rear feet

c) one fixed tilted position between a) and b)

1.6.1 C.R.T.

Type

150 BTB 31 rectangular tube with 2KV accelerating

voltage

Screen type

P 31 phosphor standard

Useful screen area

8 x 10 div of 1 cm

Graticule

Internal graticule with Centimeter division and 2 mm

divisions along the central axes.

1.6.2. **VERTICAL OR Y-AXIS**

Response

DC: 0 Hz ... 15 MHz (-3dB) AC: 10 Hz ... 15 MHz (-3dB)

Risetime

23 ns approximately

Deflection coefficient

5 mV ... 20V/div. calibrated steps, 1-2-5 sequence.

Accuracy

Display modes

Α

:

A & B in chopped or alternate mode

Input impedance

1 M Ω // 35 pF

AC, DC

Input coupling

Maximum rated input

400 V (dc + ac peak)

voltage

test voltage 570V(rms)

50 Hz during 1 min according to IEC 348

1.6.3. HORIZONTAL OR X-AXIS

Horizontal display modes

Time base

X-Y operation with X deflection via A-input

DC: 0 Hz ... 1 MHz (-3dB) Response

Deflection coefficients See Y-axes Phase error 3° at 10 kHz

1.6.5. TIME BASE

> $0.2s/div. 0.5\mu s/div.in 2 \times 9$ calibrated steps in 1-2-5 Time coefficients

sequence. Variable sweep rate facility at any time/div

setting.

X 5 magnifier extends max. sweep rate to 100ns/div

Accurancy •

Additional error for magnifier: ± 2%

1.6.6. TRIGERRING

> Trigger source Internal: A or B

External

Trigger coupling Normal (AC Coupled)

T۷

Slope : + or -

Trigger sensitivity Internal: 1 div. at 100 kHz

External: 0.75V at 100 kHz

Trigger level range ± 8 div.

External trigger

input impedence I M Ω // 35pF

400V (dc + ac peak) Max. rated input voltage

test voltage 570 V(rms)

50 Hz during 1 min according to IEC348

1.6.7. **Z-MOD INPUT**

> Trace blanking TTL High blanks trace.

OV or not connected no trace blanking.

Max. rated input voltage + 25V and - 10V.

1.6.8. **CALIBRATION**

Signal available for probe adjustment

POWER 1.6.9.

> Line voltage (ac) and freq : 108...132V, 45...66Hz

198...242V, 45...66 Hz 216...264V, 45...66 Hz

28VA maximum. The insulation between Power consumption

PM 3206 and line fulfills the safety requirements of

IEC 348 for Class I instruments.

ENVIRONMENTAL CAPABILITIES 1.6.10.

The environmental data are valid only if the instrument is checked in accordance with official checking procedure. Details on these procedures and failures criteria are supplied on request by PHILIPS Organisation in your country or by PHILIPS, scientific & Industrial Equipment Division, Eindhoven, The Netherlands.

Ambient	temperature
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Rated range of use: 5°C...+40°C Limits for operating: -10°C...+55°C Storage and transport: -40°C...+70°C

Altitude

Operating: to 5000m (15000 ft) Non-operating: to 15000m (45000 ft)

Humidity

In accordance with IEC 68 Db

Shock.

300m/s²(30g): half sine wave shock of 11ms. duration: 3 shocks per direction for a total of 12 shocks.

Vibration

30m/s²(3g)vibrations in three directions with a maximum of 15 min. per direction; 10 mins. with a frequency of 15-25 Hz and a peak-peak altitude of 1 mm. Unit mounted on vibration table without shock absorbing

Recovery time

Operates within 60 min. coming from - 10°C soak, going into 60% relative humidity at +20°C room condi-

tions.

hocks.

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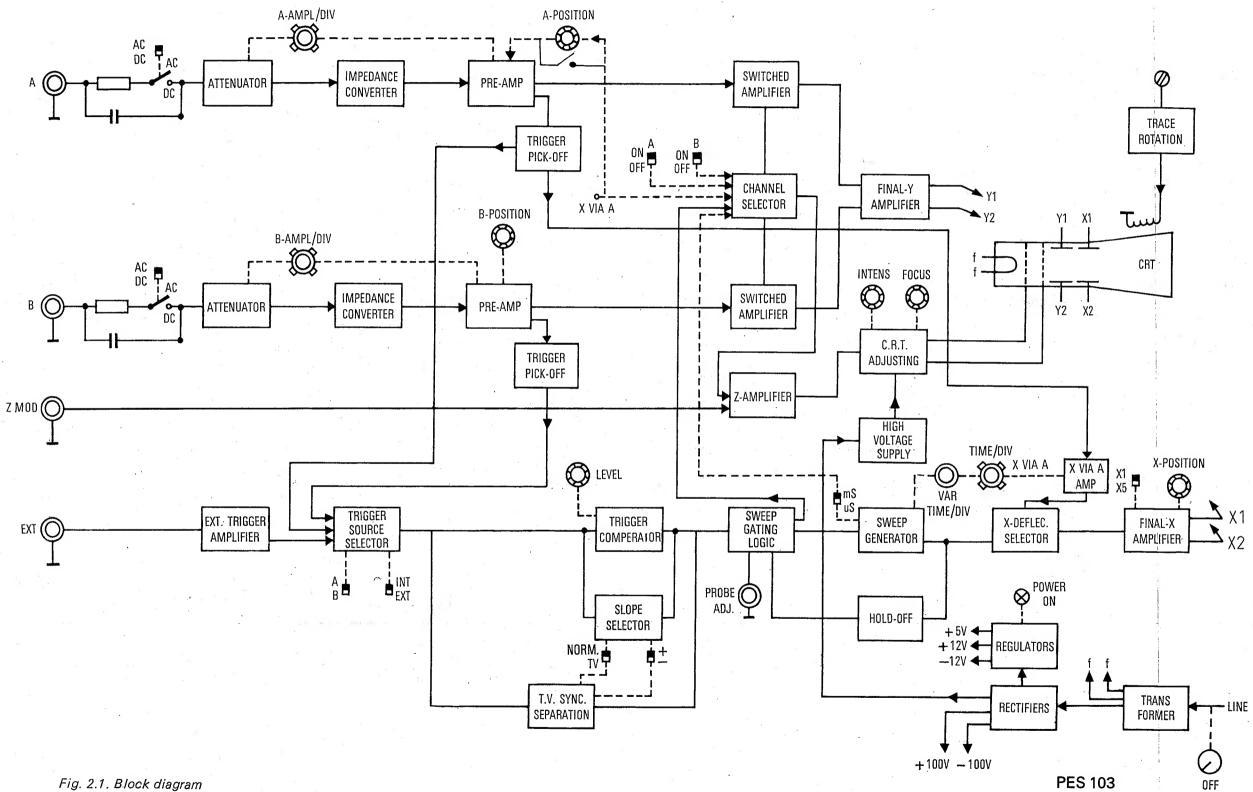
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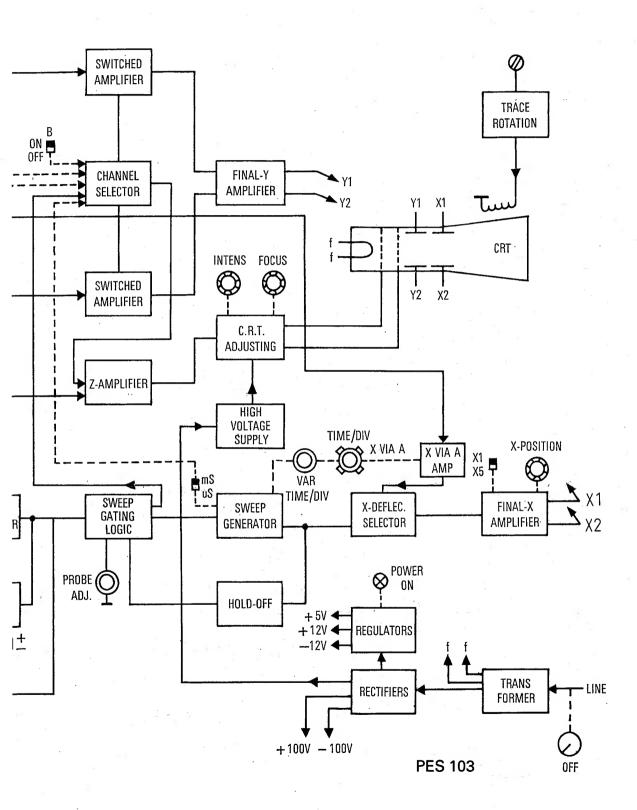
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2.1.2





2. CIRCUIT DESCRIPTIONS

In Chapter 2.1 the block diagram description is given and in the Chapters 2.2 - 2.5, the detailed circuit information is described

2.1. Block Diagram Description

The circuit is located on printed circuit boards and is subdivided into the following parts:

- vertical amplifier channels A and B
- final vertical amplifier and Z amplifier
- time base and horizontal amplifier
- c.r.t. circuits
- power supply unit

The block diagram of the PM 3206 is given in Fig. 2.1. The subdivisions of the diagram do not necessarily relate to the circuit areas of the printed circuit boards.

2.1.1 Vertical Deflection

As the A and B channels are almost identical, only channel A is described. The signal at the input socket is applied either directly or via a d.c. blocking capacitor, depending upon the position of the AC/DC coupling switch to the high-impedance A ATTENUATOR stage. This stage incorporates the 1, 10, 100 and 1000 times attenuator coefficients, which are selected by the A AMPL/DIV switch. The A attenuator is followed by an IMPEDANCE CONVERTER and PREAMPLIFIER, which provides the following functions:

- the 1-2-5 attenuator sequence in conjunction with the basic attenuation coefficients
- an adjustment of DC balance
- Y position control for channel A. This control is disabled inthe X via A mode.

The PREAMPLIFIER in channel A has two outputs, one of which feeds the TRIGGER PICK-OFF amplifier and the other the SWITCHED AMPLIFIER. The CH. A TRIGGER PICKOFF Amplifier feeds the trigger source selector and the X via A Amplifier.

Depending on the command signal from the CHANNEL SELECTOR, the SWITCHED AMPLIFIER routes the PREAMPLIFIER signal to the FINAL Y AMPLIFIER. The FINAL Y AMPLIFIER feeds the YI and Y2 vertical deflection plates of the CATHODE RAY TUBE (C.R.T). The CHANNEL SELECTOR signals are generated by a multivibrator controlled by logic gates from the front-panel channel selection switches A ON-OFF, B ON-OFF, µs/ms (ALT./CHOP.), X VIA A. For the µs (ALT) mode, an input signal derived from the time-base is used for switching purposes. The following modes are possible:

- Single Channel Operation: One channel is permanently to the final Y amplifier while the other channel is blocked.
- Dual Channel Operation in μs (ALT) mode: the final Y channel input is switched from one channel to the other at the end of each time-base sweep.
- Dual Channel Operation in ms (CHOP) mode: the final Y amplifier input is switched from one channel to the other at a frequency of approx. 60 KHZ.
- X via A mode: the channel B is switched through to the final Y amplifier.

2.1.2 Triggering and Horizontal Deflection

The signal source required for time-base triggering is selected by the TRIGGER SOURCE SELECTOR under the command of the front panel trigger selector switches. The trigger sources are available from the following:

- The channel A or B TRIGGER PICK-OFF amplifiers which are activated by the A and B switches.
- The EXTERNAL TRIGGER AMPLIFIER in order to trigger on a signal applied to the external input X5 activated by the EXT switch.

The TRIGGER SOURCE SELECTOR output is applied to the TRIGGER COMPARATOR, where it is compared with the input from the LEVEL control. This enables the selected level to be compared with the incoming trigger signal to determine the level at which the time-base starts. With switch NORM/TV in TV position the TRIGGER SOURCE SELECTOR output is also fed to the TV SYNC SEPARATION stage. In case of positive video signal the +/- switch should be in + position.

The TV SYNC SEPARATOR stage allows triggering on TV frame pulses (ms position of ms/µs switch) or TV line pulses (µs position of ms/µs switch). The output of the TRIGGER COMPARATOR feeds the SWEEPGATING LOGIC. Depending on the position of the +/- switch the SWEEP GENERATOR starts the positive or negative going slope of the input signal. With the NORM/TV switch in TV position the TV SYNC SEPARATOR output is applied to the SWEEP GATING LOGIC and the TRIGGER COMPARATOR output is blocked. The HOLD-OFF circuit prevents the SWEEP GENERATOR from responding to a trigger command before the time base capacitors are fully discharged.

The SWEEP GATING logic, in addition to feeding the SWEEP GENERATOR, also feeds the CHANNEL SELECTOR in order to control the ALTERNATE vertical display mode switching. An output also controls the Z AMPLIFIER in order to blank the trace during the flyback of the time-base.

The SWEEP GENERATOR produces the sawtooth waveform that is used for horizontal deflection. The time-base sweep period can be adjusted by the step control TIME/DIV. The VAR. TIME/DIV control provides the variable sweep facility.

The X DEFLECTION SELECTOR enables the input to the FINAL X AMPLIFIER to be selected by diode switching networks. Normally, the internal time-base produced by the SWEEP GENERATOR is routed to the FINAL X AMPLIFIER, but in the X via A mode, an output signal from the X via A AMPLIFIER is selected.

The FINAL X AMPLIFIER, which drives the X1 and X2 horizontal deflection plates of the C.R.T has a X POSITION control potentiometer. In addition the stage has x5 magnifier switch facility, which increases the horizontal gain by a factor of 5.

2.1.3 CRT Display Section

The Z AMPLIFIER receives an input blanking pulse which originates in the channel selector multivibrator and the sweep gating logic. There is also another input to this amplifier from the Z-MOD input socket. Normally, blanking of the trace takes place during flyback of the time-base and also in the chopped mode during switching from one channel to the other. The INTENS control determines the d.c. level fed to the cathode of the CRT. The output of the Z AMPLIFIER is capacitively coupled to the CRT control electrodes.

A HIGH VOLTAGE multiplier provides the final anode potential (-2kV) of the CRT. The TRACE ROTATION circuit, operated by a front panel control, enables adjustment of the sense and strength of current through the trace rotation coil of the CRT. This allows alignment of the trace with the horizontal graticule lines.

2.1.4 Power Supply

The mains supply is transformed by means of a TRANSFORMER and rectified to give d.c. supplies of $\pm 100 \, \text{V}$ and $\pm 100 \, \text{V}$.

VOLTAGE REGULATOR stages provides low voltage d.c. outputs of +5V, +12V and feed the various circuits of the oscilloscope.

A 6.3V a.c. secondary winding of the mains transformer supplies the filament of the CRT.

2.2 DESCRIPTION OF THE VERTICAL SECTION

As channels A and B are identical, only channel A is described.

2.2.1. Input coupling stage

In the DC position (switch S9 closed), the signal applied to input socket XI is fed to the high impedance attenuator via R20 and R21.

In the AC position (switch S9 open), the signal applied to input socket XI is fed to the high impedance attenuator via R21 and d.c. blocking capacitor C20.

Resistor R20 discharges C20 when switch S9 is changed from the AC to the DC position.

2.2.2. High impedance attenuator

This section of the circuit comprises of the 1,10,100 and 1000 times attentuator.

The 1000 times attenuator is active in the 5,10 and 20V/div attenuator switch (S4) positions; i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising of R32, R33, R35 and parallel capacitors. The signal, reduced by an attenuation factor of 1000 is fed via K25 contact to the PREAMPLIFIER.

The 100 times attenuator is active in the 0.5,1 and 2V/div attenuator switch (S4) positions; i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising R28, R30, R31 and parallel capacitors. The signal, reduced by an attenuation factor of 100 is fed via K23 contact to the PREAMPLIFIER.

The 10 times attenuator is active in the 50mv, 0.1V and 0.2V/div attenuator switch (S4) positions i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising R25, R26, R27 and parallel capacitors. The signal, reduced by an attenuation factor of 10 is fed via K22 contact to the PREAMPLIFIER.

The 1 x attenuator is active in the 5mV, 10mV and 20mV/div positions of S4 i.e. the output signal from the coupling stage is applied via K20 to the attenuator section comprising of R22, R23 and parallel capacitors. In conjunction with these four basic attenuator coefficients switched by reed relays K20 (5mV-20mV), K22 (50mV-0.2V), K23 (0.5V-2V) and K25 (5V-20V), the 1-2-5 attenuator sequence of adjacent ranges is provided by gain switching of PREAMPLIFIER by relays K26 and K27.

2.2.3. Preamplifier

The output of the HIGH IMPEDANCE ATTENUATOR is connected via resistor R36 to the input of a symmetrical impedance converter consisting of two matched FET's V21 in source follower configuration.

Diode V20 protects the FET input against excessive negative voltages. The output of the Impedance Converter is fed to a transistor array D20 which uses series feedback in the emitter for gain control. The current source for this circuit is obtained with a transistor biased by resistors R50, R51 and R52. Resistors R47 and R48, with K26 and K27 released, determine the gain of the stage in the attenuator switch (S4) positions 20mV, 0.2V, 2V and 20V/div.

- When contact K26 closes, R56 is switched into circuit and the gain of the stage increases 2x. This occurs in the attenuator switch positions 10mV, 0.1V, 1V and 10V/div.
- When contact K27 closes, R57 is switched into circuit and the gain of the stage increases 4x. This occurs in the attenuator switch positions 5mV, 50mV, 0.5V and 5V/div.

Resistor R55 in this preamplifier stage is used to calibrate the gain in 5mV/div range. D.C. balance control potentiometer R60 adjusts the D.C. balance for the 20mV/div switch position. Potentiometer R40 is used to adjust the D.C. balance for 5mV/div switch position.

The final stage of the preamplifier circuit uses two transistors V22 and V23 with shunt feedback resistors R61 and R62. The Y position control circuit is implemented here using transistors V25 and V26.

Resistor R2 is used for Y position control. The collector current drawn by the transistor array D20 is fixed and is dependent on the current source feeding the emitters of D20. The collectors of V25 and V26 are also connected to the collectors of D20. The change in the collector currents of V25 and V26 due to variation of R2 will now be reflected in the collector voltages of V22 and V23 which are the output voltages of the preamplifier circuit. Relay K28 (operated in X via A mode) shorts the Y position control R2 in the X via A mode only. The outputs of the preamplifier stage are d.c. coupled to the trigger amplifier circuit via resistors R75 and R76.

2.2.4. Switched amplifier

The output of the preamplifier stage goes to the series stage of the output amplifier consisting of transistors V30, V31 with series feedback resistors R80, R81, R82. R83 is used for gain adjustment in the 20mV/div attenuator switch position. In this stage itself channel switching is incorporated with the help of a switching current source (V32). When the output Q of the flipflop D200 in the CHANNEL SWITCHING circuit goes high, the current source (V32) conducts and switches channel A 'ON', while output $\overline{\mathbb{Q}}$ of the same flipflop switches channel B 'OFF'.

The output of this stage goes to the final Y amplifier circuit via resistors R91 and R92.

2.2.5. Channel flipflop and logic circuits

The logic circuits used in this unit can have two logic output levels:

A low level or logic 0 between 0V and 0.8V and a high level or logic 1 between 2V and 5V. The unit has two outputs (pins 9 & 8 of D200) which are the Q and \bar{Q} outputs of a D flipflop. A logic 1 in the Q and \bar{Q} of this flipflop switches channel A and channel B respectively. The various switching modes that control these outputs are CH. A ON/OFF, CH. B ON/OFF, the ms/µs (for selecting ALT/CHOP mode of operation) and X via A modes.

CH. A ON

When channel A switch S8 is ON, input 3 of NOR gate D202 is at logic 0. Unless the X via A mode is used, the other input 2 of this gate is also at logic 0. Hence the output of this gate is at logic 1.

If channel B switch S10 is OFF, then input 1 of NAND gate D201 is at logic 1 and hence the output 3 of D201 is at logic 0. This causes the D flipflop D200 to be set and its Q output 9 goes high switching ON channel A.

CH, B ON

When channel B switch S10 is ON, input 1 of NAND gate D201 is at logic 0. If channel A switch S8 is OFF, then input 3 of NOR gate D202 is at logic 1 and its output 1 is at logic 0. This means that reset input 13 of flipflop D200 is at logic 0 and hence its $\overline{\mathbb{Q}}$ output goes high, switching channel B ON.

CH. A ON, CH. B ON and switch SI in us position (ALT mode). Here the +12V on SI is fed via resistors R231 and R232 to provide a logic 1 at input 5 of NOR gate D202. This makes the output of this gate go to logic 0 which blocks the CHOP oscillator D201/13, 12, 11 and the output D201-11 is at logic 1.

Output 5 of flipflop D303, connected to input 4 of flipflop D200 is at logic 0 during the time base sweep and at logic 1 during the hold-off period. Thus input 9 of NAND gate D201 is at logic 1 during the time base sweep and at logic 0 during the hold-off period. Therefore, the CLOCK input (pin 11) of flipflop D200 goes from logic 0 to logic 1 at the end of every sweep and changes the state of the flipflop. In this way the display switches alternately between the channels.

CH. A ON, CH. B ON and switch SI in ms position (CHOP mode). Here the input 5 of NOR gate D202 is at logic 0 since switch SI is in 'ms' position. Since both inputs 2 and 3 of NOR gate D202 is at logic 0, its output 1 is at logic 1. Since channel B is ON, input 1 of NAND gate D201 is at logic 0. Hence output 3 of the same gate is at logic 1. Since inputs 4 and 5 of NAND gate D201 is at logic 1, therefore its output 6 is at logic 0. Hence output 4 of NOR gate D202 is at logic 1. This enables the CHOP oscillator D201/13, 12, 11. This oscillator is a NAND - schmitt trigger with a RC feedback loop, which produces a 120 KHZ square wave signal on its output pin 11. This is at logic 1 if the oscillator is switched off in single channel or ALT mode.

The oscillator output is fed to input 10 of NAND gate D201.

During the time base sweep the other input (pin 9) of D201 is at logic 1; therefore the inverted chopper pulses are fed to the CLOCK input (pin 11) of flipflop D200. As both the clear and Preset inputs of the flipflop are at logic 1 (switches S8 and S10 ON) they are inactive. Therefore, due to the feedback connection between output pin 8 and pin 12, the flipflop changes state at every clock pulse. In this way the display switches between the A and B channels at a frequency of 60 KHZ.

The CLOCK input 11 of flipflop D200, which gets the inverted chopper pulses, is also taken to the Z amplifier to blank the display when switching over between the A and B channels.

	D201/3	D202/1	D201/11	D200/10	D200/13	D200/9	D200/8
				PRESET	CLEAR	Q	Q
AON/BOFF BON/AOFF A & B ON - in ALT m	1	1 0 1	- - 1	0 1 1 (sta	1 0 1 te changes a	1 0 1/0 t end of every	0 1 0/1 sweep)
(us) - in CHOP mode (ms)	1	1	1/0	l (sta	1 ite changes	1/0 at chopper frec	0/1 Juency)

2.2.6 Final Amplifier

The final Y amplifier consists of V200, V202, V207, V210 which drive the Y1 deflection plate and are balanced by transistors V201, V203, V208, V211 which drive deflection plate Y2. In order to increase stability, the deflection plates are driven via resistors R220 and R221.

In the Y1 plate drive section, V207 & V210 function as a current source. Transistors V200 and V202 form a shunt feedback stage. Two transistors are used in each case so as not to exceed the maximum permissible current and voltage limits of the transistors. In the Y2 plate drive section, V208 and V211 are the current source and V201 and V203 for the shunt feedback stage.

2.3 HORIZONTAL DEFLECTION AND TIME BASE

2.3.1 CH.A trigger pick off

The trigger signal picked off from the output stage of the CH. A PRE-AMPLIFIER is fed via resistors R75 and R76 to the trigger amplifier consisting of V300 and V301 for which V302 forms a current source. When channel A triggering is selected (S12 to A and S14 to INT), then the collector of V300 is applied via diode V308 in the trigger source select to the base of transistor V315. The channel B pick off and EXT trigger input are inhibited by the -12V switched supply that switches off diods V311 and V310.

The channel A trigger amplifier has another output from the collector of V301 that is used to drive the horizontal amplifier, via the shunt feedback stage using V418, in the X via A mode.

2.3.2 CH. B trigger pick off

The trigger signal picked off from the channel B pre-amplifier is fed to the trigger amplifier consisting of V303 and V305 for which V306 forms the current source. The collector signal from V303 is routed via diode V311 in the trigger source selector to the base of transistor V315. The A channel pick off and the EXT trigger input are inhibited by the -12V switched supply that switches off diodes V308 and V307.

2.3.3 External trigger amplifier

The signal applied to the external trigger input socket (X5) is attenuated by a voltage divider network R342/C302, R343/C303 in the base circuit of emitter followers V326 and V327 connected in cascade. Capacitor C305 serves for d.c. blocking and diode V325 protects transistor V326 against excessive positive input voltage swings.

The emitter of V327 is coupled via C306 to the series feedback stage V328/V330. The collector current of V328 is connected via diode V312 to the base of transistor V315. The A and B

internal pick offs are inhibited in the EXT position of S14 by the -12V switched supply via diodes V310 and V307 that blocks V311 and V308 respectively.

2.3.4. Trigger source selector amplifier

The diode networks referred to in the foregoing descriptions of the trigger pick offs and the external trigger amplifier are all associated with the trigger source selector switches S12 and S14, and the method of selecting each trigger source has been described under these headings. The selected trigger signal to the base of transistor V315 is amplified and fed to emitter follower V318. Electrolytic capacitor C300 connects this signal to the comparator circuit which follows.

2.3.5. Level control

The level voltage control R5 permits variation of the trigger level of the signal.

2.3.6. Comparator and sweep gating logic

The trigger level selected by the LEVEL control is applied to the base of V320, which together with V321 forms a differential amplifier comparator circuit, for which V322 acts as a current source. The trigger signal from the emitter follower V318 is fed directly to the base of V321. When this trigger signal exceeds the reference set by the LEVEL control, V320 conducts less and the collector current decreases so the voltage of the shunt feedback stage V323 increases.

This signal is applied to input 9 of NAND schmitt - trigger D300 in the sweep gating logic, and if S15 is in NORM position the inverted signal appears at D300. With the +/- switch S13 in "+" position, output 8 of D301 is logic 1 and the EX-OR D301 (output 6) inverts the signal applied to input 5, so triggering is effected on the positive slope. If switch S13 is in "-" position the EX-OR D301 input 4 is at logic 0 so the signal applied to input 5 appears at pin 6. Now triggering is effected on the negative slope.

2.3.7. T.V. Slope selector

The output signal from the trigger source selector is fed to the base of transistor V331 which is balanced by transistor V332.

In the negative slope position of S13 the signal on collector of V332 is routed via diode V336 to the base of transistor V338 and the inverted signal on collector of V331 is inhibited by the switched +12V supply which blocks diode V337.

In the positive slope position of S13 the signal on collector of V332 is routed via diode V337 to the base of transistor V338 and the inverted signal on collector of V332 is inhibited as diode V336 is now blocked.

In the NORM position of S15 diodes V333 and V335 conduct and V337 and V336 are blocked.

2.3.8. T.V. SYNC Separator

The line and frame TV trigger pulses from the slope selector stage are passed via V338 to V340 to a low-pass filter for the frame pulses. The low cut-off frequency is selected in the ms position of S1, which connects capacitor C310 across the output, and triggering on TV frame pulses is possible.

In the μs position of S1, this capacitor is disconnected and triggering on TV line pulses is now possible.

The output on the collector of V343 applies a logic signal to pin 4 of NAND gate D300.

With TV selected (D300 input 12 at logic 0) the other input (pin 5) of D300 is at logic 1 and therefore output pin 6 is the inverted logic signal.

This signal is inverted again and appears at pin 6 of D301.

Time-Base and Hold-Off Circuit 2.3.9.

These two functional blocks are not described separately here because they function interdependently.

The time-base is built around the timing capacitor C402, which is always in circuit, and C401 which is switched into circuit via transistor V402 at the low sweep speeds by the +12V on switch S1 (ms position) via V400 and R402.

A constant current from current source V405 charges the capacitor(s) in order to produce a time-base voltage that is linear with respect to time; i.e. a linear sawtooth. The TIME/DIV control (S6) is incorporated in the emitter circuit of the current source transistor V405.

The TIME/DIV controls, R411 (us range) and R416 (ms range), adjust the base voltage of V405 in diodes V407 and V406 respectively. A variable time/div control is obtained with a potentiometer R7 (which can be shorted by switch S7) which is connected to the base circuit of V405 via R406 and controls the base voltage in calibrated time base mode.

The appropriate base control circuit for V405 is selected by the position of the SI switch, which provides the $\pm 12V$ to either the μs or ms position.

The time base capacitors are charged during the time-base sweep.

During this charging time, switching transistor V401, which is controlled by the sweep gating logic, is not conductive. This transistor, which starts to conduct at the end of the sweep, discharges the timing capacitor(s) and takes over the current from V405.

Switching transistor V401 cuts off when the time base is ready to start again.

The sawtooth time base voltage on the timing capacitor is picked-off by a Darlington stage (V408 and V410) and is applied to the X-deflection selector. The output signal from V410 is also applied via R421 to emitter follower V411, which feeds the hold off capacitors C403 and C405. Capacitor C405 is always in circuit and, capacitor C403 is switched into circuit by V403 in ms position of SI. The sawtooth on the hold-off capacitor(s) is applied to the input of NAND gate D300/1,2,3. This gate is effectively a schmitt-trigger with a hysteresis of approximately 0.8V. The output of the gate becomes logic 0 if the positive going slope of the input sawtooth reaches a level of IV approximately.

The two D flipflops D303 operate in parallel. The non inverting output of one flipflop (pin 5) feeds the switching transistor V401 via R400 and the Z-amplifier.

The non-inverting output of the other flip-flop (pin 9) is used as probe adjust signal.

Integrated circuit D302 is a retriggerable monostable multivibrator controlled by the trigger pulses from the trigger comparator. When a trigger pulse is received on pin 3, output pin 8 is at logic 1 for 150 ms. This time constant is determined by C312.

The trigger pulses are also routed to the clock inputs (pin 3 and pin 11 of the D303 flipflops).

Free-Run Mode (Without Trigger Pulses)

If no trigger pulses are available at retriggerable monostable input D302-3, then output pin 8 is low and a preset command is given to the D303 flipflops (pins 1 and 13).

The D flipflops now function as inverters and the clear inputs (pins 4 and 10) receive a pulse from the hold-off circuit after the time base sweep, which is inverted to give outputs on pins 5 and 9.

The output pulse on D303-5 causes the switching transistor V401 to conduct at the end of the time-base sweep and during the hold-off period the time base capacitors C401 and C402 are discharged.

In this way, the time base capacitors are alternately discharged and then charged; i.e. the timebase is free-running.

X via A mode

When X via A is selected, the +12V that is available on H performs four functions:

- Via diode V317 it inhibits the trigger source input to V318

- Via diode V215 a logic 1 is applied to input pin 2 of gate D202, which causes a logic 0 on input 13 of flipflop D200. This results in channel B being switched through for Y deflection purposes and the output of channel A being blocked. Channel A output from vertical preamplifier is routed via V418 to provide the X-deflection signal.

The +12V applied via R441 to the base of V417 switches off this transistor. The pulses from the internal time base via emitter of V410 are now blocked by diodes V412 and V413.

- The +12V applied via R442 allows switching diodes V415 and V416 to conduct so that the output on collector of transistor V418 is routed to the final X amplifier.

2.3.10. X-Deflection selector

The selection of the X via A mode has previously been described. When the internal time-base mode is selected, V417 conducts because of the bias current applied to its base via R440 and R441. The positive voltage on the collector of V417 causes switching diodes V412, V413 to conduct, which allows the time base output on emitter of V410 to be applied to the input of the X-amplifier. In this mode, switching diodes V416 and V415 are blocked by the negative potential applied via R440 and R442.

2.3.11. Final X Amplifier

The output signal from the X deflection selector is applied to the base of V421 in the series feedback stage, which consists of V421 balanced by V423. The base circuit of V423 incorporates the horizontal position control (X-POS) R4. Transistor V422 is the constant current source for this series feedback stage. In the x5 magnifier position of S2, resistor R456 shunts the emitter resistors R455, R453 to give a 5 times increase of horizontal gain.

The collectors of V423 and V421 are coupled to the output stage. This output stage consists of the shunt feedback stage V426, V428 and current source V433 that feed the X plate via R478, balanced by an identical stage comprising V427, V430 and V435.

Two transistors are employed in each of the shunt feedback stages so that the maximum current and voltage limits of the individual transistors are not exceeded, and to reduce stray capacitances.

Resistors R478 and R480 connecting the outputs to the X-plates of the CRT are inserted to increase stability.

2.4. CRT DISPLAY SECTION

2.4.1. Z Amplifier

The input to the Z amplifier is via R522 to the base of transistor V517 and receives signals from :

- the sweep gating logic in order to blank the display during the time-base hold-off period.
- the channel multivibrator in order to blank the display in the chopped mode during the switching from one channel to the other.

The Z amplifier consists of a shunt feedback stage coupled to the Wehnelt cylinder via C518. Diode V511 and resistor R512 provide D.C. restoration.

2.4.2. High voltage supply

The high voltage power supply consists of a quadruple voltage multiplier a voltage divider that produces cathode, control grid and focus potentials for the CRT and a compensation circuit (V505, V506, V504, V507) to compensate h.t voltage ripple and variations. The voltage quadrupler circuit consists of diodes V500, V501, V502, V503, and capacitors C500, C501, C502, C503. Resistors R6 and R1 control the cathode and grid g3 voltages respectively and in turn provide the intensity and focus controls.

2.4.3. Trace rotation

The emitter followers V521 and V522 and preset potentiometer R8 determine the sense and strength of the current in the trace rotation coil.

Only one emitter follower conducts at any given time, depending on the position of R8.

2.5. POWER SUPPLY

The mains voltage is applied via double-pole switch S3 to the primary winding of transformer T500, protected by a replaceable thermal fuse F500 and a replaceable cartridge fuse F501.

Provision is made to wire the primary for a nominal voltage of 120V or 220V or 240V. Two full wave bridge rectifiers V512 and V516 across the secondary winding of T500 provide the d.c. voltages for the $\pm 12V$, $\pm 12V$, $\pm 100V$ and $\pm 100V$ supplies respectively.

The low voltage supplies +12V and -12V are regulated by two integrated circuits D500 and D501 and smoothed by electrolytic capacitors C511 and C515. The +5V supply is obtained from the +12V supply with the help of a series pass transistor with Zener reference.

DISMANTLING THE INSTRUMENT

3.1. General information



3.

WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals may be live.

The instrument shall be disconnected from all voltage sources before any adjustment, replacement or maintenance and repair during which the instrument will be opened. If afterwards any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved. Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

ATTENTION: This section provides the dismantling procedures required for the removal of components during repair operations. All circuit boards removed from the oscilloscope should be adequately protected against damage, and all normal precautions regarding the use of tools must be observed.

During dismantling procedures, a careful note must be made of all disconnected leads that they may be reconnected to their correct terminals during assembly. Damage may result if the instrument is switched on when a circuit board has been removed, or if a circuit board is removed within one minute after switching off the instrument.

NOTE: All screws which have to be remounted directly in the housing-parts must be fixed with a torque of maximum 1 Nm (10 kg cm).

3.2. Removing the top and bottom covers

To adjust the instrument it is necessary to remove the top-cover.

- Remove the two carrying-handle mounting screws (Fig. 3.1.).
- Bend the handle outwards and remove it (Fig. 3.1.).
- Remove the two cabinet mounting-screws (Fig. 3.1.).
- Press the two buttons at the rear side until the click. (Fig. 3.2.)
- The top-cover will lift now about 2 mm (Fig. 3.2.).
- Now lift vertically the top-cover out of the front-and rear-cover (Fig. 3.3.).
- The bottom cover can now be removed.

NOTE: Take care of the handle gears.

3.2.1. Remounting the top-cover

- Place the top-cover between the front and rear-cover.
- Take care that the side snaps of the top and bottom-cover fix together.
- Press the upper rear side firmly down until the click (Fig. 3.2.).
- Remount cabinet mounting-screws and the handle.

3.3 ACCESS TO PARTS FOR CHECKING AND ADJUSTING PROCEDURE

The adjusting elements are accessible after removing the top-cover. To remove the top-cover see section 3.2.

NOTE: For adjustments always use an insulated adjustment tool.

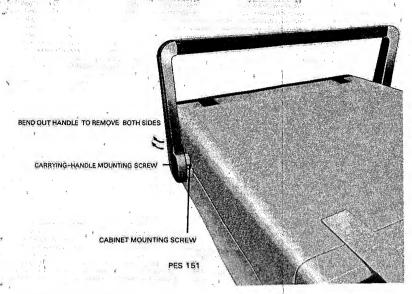


Fig. 3.1 Removing the top Cover

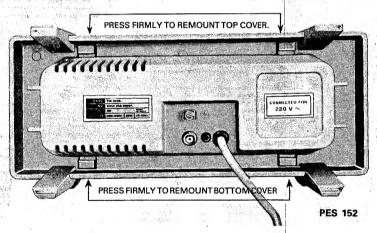


Fig. 3.2 Remounting the top cover

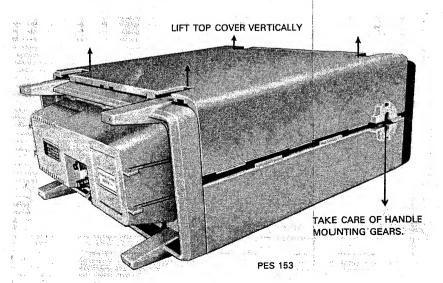


Fig. 3.3 Lifting the top cover.

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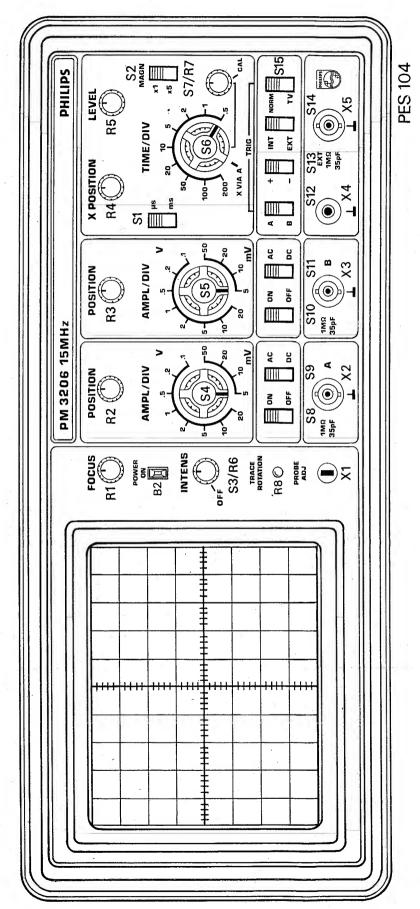


Fig. 4.1 Preliminary Settings of the controls

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4. PERFORMANCE CHECK

General information



WARNING: Before switching on, ensure that the oscilloscope has been installed in accordance with the instruction outlined in Chapter 4, Directions for use of the Operating Manual.

This procedure is intended to check the instruments specifications. It can be used for incoming inspection to determine the acceptability of newly purchased or recently recalibrated instruments, or to check the necessity of recalibration after a certain operating period. It does not check every facet of the instruments calibration; rather it is concerned primarily with those portions of the instrument which are essential to measurement accuracy and correct operation. Removing the instruments covers is not necessary to perform this procedure. All checks are made from the front panel.

If this test is started a few minutes after switching on, bear in mind that test steps may be out of specification, due to insufficient warming-up time. To avoid this situation, allow the specified warming-up time.

Numerical values without tolerances are typical and represent the characteristics of an average instrument.

The performance checks are made with a stable, well-focussed, low-intensity display. Unless otherwise noted, adjust the intensity, focus and trigger-level controls as needed.

NOTE 1: At the start of every objective, the controls always occupy the preliminary settings; unless otherwise stated.

NOTE 2: The input voltage has to be supplied to the A-input; unless otherwise stated.

NOTE 3: Set the TIME/DIV switch to a suitable position; unless otherwise stated.

4.2. Preliminary settings of the controls

- Start this check procedure with NO input signals connected.
- Set the controls as indicated in fig. 4.1.

4.3. Recommended test equipment

Type of instrument	Required specification
	a di di
Constant amplitude	Freq.: 1 Hz 15 MHz
Sine-wave generator	Constant ampl. of 10 mV p.p 30 V p.p.
Square-wave generator	Freq.: 2 Hz 1 MHz Ampl.: 10 mV 12 V Rise-time 3 nsec. Duty cycle 50%
Time-marker generator	Repetition rate: 200 msec 100 nsec.
Dummy probe 2: I	1, M Ω ± 0.1 % // 40 pF.

CHECKING PROCEDURE

STEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	MEASURING RESULTS
			et		
4.4.1	POWER ON				
4.4.1.1	Start power on		- Turn Switch S3 ON	- Starts at selected mains voltage ± 10% and mains frequency 45-66 Hz.	
				- Pilot Lamp POWER ON lights up.	
4.4.1.2	Power Consumption	y.		- 28 VA	
4.4.2	CRT SECTION	H)			
4.4.2.1	Intens		- INTENS Control R6	- Normal intensity adjustments	
4.4.2.2	Focus		- Focus Control R1	- Normal focus adjustments	
4.4.2.3	Trace Rotation		Screwdriver adjustment TRACE ROT R8.	- Trace must coincide with central horizontal graticule line; if necessary, readiust potentiometer	
			* .	TRACE ROT R8.	
4.4.3	VERTICAL AXIS				
4.4.3.1	Display modes		- Position control R2, R3.	- Traces of channel A and channel B in middle of the screen.	
		Sine-wave signal,	- Switch SI0 to OFF CH.A.AMPL/DIV. to O.IV/DIV.	- Signal of 3 div. visible on the screen.	
	-	A & B input			
		-	- Switch S8 to OFF. CH. B AMPL/DIV to 0.IV/DIV.	- Signal of 3 div. visible on the screen.	

STEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	MEASURING RESULTS
4.4.3.2	Input Coupling	Sine wave signal 1kHz +DC offset to A(B) input	- Switches S9 and S11 to DC	- Signal is visible on the screen, centre of sinewave is on DC offset level.	
4.4.3.3	Vertical deflection coefficients	Square wave signal, IkHz to A(B) input Amplitude:	- Switch S12 to A(B) - AMPL/DIV. switch position of S4(S5)	- Trace height:	
		30 mVpp 60 mVpp 120 mVpp 300 mVpp 600 mVpp 1.2 Vpp	5 mV 10 mV 20 mV 50 mV 0.1 V	6 div. ± 5%(±1.5 sub-div.)	
	*		10 V 20 V 20 V 20 V		
4.4.3.4	Input Impedance	Square wave signal, 1.2 Vpp - IkHz to A(B) input via dummy	- AMPL/DIV switch position of S4(S5) to 0.IV	- Trace height 6 div.	S.
		*			
4.4.3.5	Square Wave Response	40pF. PES121 Square wave signal, 600 mVpp,100 kHz, rise time < 5 n secs. to A(B) input.	- Switch S1 to µ secs. - Switch S2 to x5 position	- Rise time < 23 n secs. - Pulse ringing ±5% (±1.5 sub-div.)	

STEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	
				NE CONCEMENTS	MEASURING RESULTS
		- E			
4.4.3.6	Band Width	Sine wave signal to A(B) input	- Switch SI to us.		
	i i	Frequency : 100 kHz		- Adjust the sine wave amplitude for a trace height of 6 div.	, be
		10 Hz - 15 MHz	- Switch SI to ms or us.	- Trace height >4.2 div.	
	-	0 Hz - 15 MHz	- Switch S9(SII) to DC		
			- Switch SI to ms or us	- Trace height >4.2 div.	
	4		-		
4.4.4	HORIZONTAL AXIS			*	·
4.4.4.1	Time Coefficients	Marker pulse signal to A input	- Var. TIME/DIV.control S7/R7 to CAL.	·	
			- Switch SI0 OFF		
			- Switch SI to µs		
		Repetition time	- TIME/DIV. Switch position:	- Coefficient error ±5%(c.1±0.5 div.	
		0.5 µs 1 µs	0.5	over 10 div. screen width)	
-	•.		2 2 5		
			10 20		
		50 µs 100 µs 200 µs	50 100 200	-	
	**		- Switch SI to ms		
			7		
		5 ms 20 ms	10 20		

SIEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	MEASURING PESTILTS
					STED COTA
		Repetition time (cont.)	- TIME/DIV Switch position (cont.)	int.)	
*		50 ms 100 ms 200 ms	50 100 200		
4.4.4.2	X Magnifier	Marker pulse signal to A input repetition time 200 µs.	- Switch SIO to OFF - Switch SI to ms	- Coefficient error ± 7%(c.i±0.7 div) over 10 div. screen width	-
		Repetition time 100 n sec	- TIME/DIV to 1 - Switch S2 to X5. - Switch S1 to µs - TIME/DIV. to 0.5	- Coefficient error ±7%(c.i±0.7 div over 10 div. screen width)	
4.4.5	HORIZONTAL AMPLIFIER		*		
4.4.5.1	X via A	Sine wave signal, 600 mVpp, 1kHz to A and B input	- Switch S6 to X via A - Switches S9 and S11 to DC	- A line is visible with an angle of 45° with respect to the horizontal graticule line	
4.4.5.2	Phase shift	as 4.4.5.1.	- as 4.5. 1	- adjust the input voltage for a deflection of 6 div.	
-		Frequency 10kHz	-	- Phase shift 3° (c.i. 0,4 div.)	
-					So.4 div 6 div
		÷			
			:	PES	PES 105
4.4.5.3	Bandwidth	Sine wave signal	- Switch SIO to OFF	- Adjust input voltage for a trace	,

STEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	MEASURING RESULTS
9*#*#	TRIGGERING				
4.4.6.1	Trigger source A & B	Sine wave signal, 1kHz to A input and square wave signal 800 Hz to B input	- Sl0 to OFF - Switch Sl2 to A - Adjust the input signals for a trace height of 6 div	- Well triggered display of channel A	
	-	-	- Switch SI2 to B	- Well triggered display of channel B	
4.4.6.2	Trigger source EXT	Sine wave signal, 600mV, 1kHz to A input and EXT input.	- Switch SI4 to EXT.	- Well triggered display	
4.4.6.3	Slope	Sine wave signal, 600 mv, IKHz to A input	- Switch Sl3 to"+"	- Signal triggers on positive going edge	
	**		- Switch SI3 to "-"	- Signal triggers on negative going edge	
ħ*9*ħ*ħ	TV triggering	TV signal to A input, syne pulse I div.	- Switch SI5 to TV	- Well triggered display	×
4.4.6.5	Sensitivity int.	Sine wave signal 100 kHz to A input		- Signal triggers at 0.75 div.	
9.9.4.4	Sensitivity EXT	Sine wave signal, 100 kHz to A input and EXT input	- Switch SI4 to EXT	- Signal triggers at 0.75 vpp.	
4.4.6.7	Level range	Sine wave signal, 4V freq.1kHz to A input	- AMPL/DIV to IV	- Signal of 4 div. visible on the screen.	1)
			- AMPL/DIV to 50mV - LEVEL control R5	- Signal triggers in the most extreme positions of R5.	Y
4.4.7.	CALIBRATION			- Signal available for probe adjustments.	

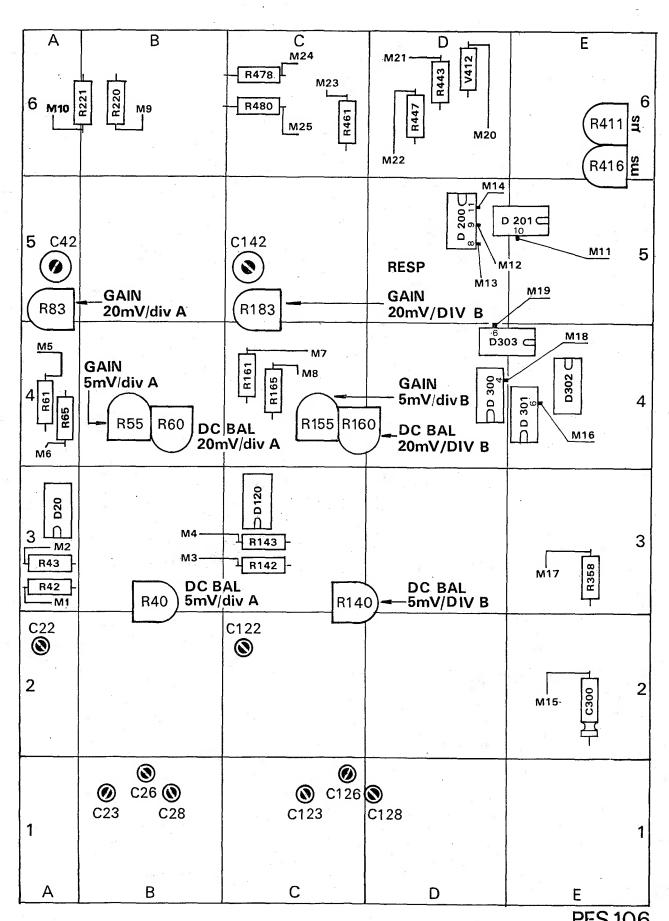


Fig. 5.1 Adjusting elements amplifier unit.

CHECKING AND ADJUSTING



5.

WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals may be live.

The instrument shall be disconnected from all voltage sources before any adjustment, replacement of maintenance and repair during which the instrument will be opened. If afterwards, any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved.

Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

5.1. General information

The following information provides the complete checking and adjusting procedure for the oscilloscope.

As various control functions are interdependent, a certain order of adjustment is often necessary. The procedure is, therefore, presented in a sequence which is best suited to this order, cross-reference being made to any circuit which may affect a particular adjustment.

- Before any check or adjustment, the instrument must attain its normal operating temperature.
 Where possible, instrument performance is checked before and adjustment is made.
- Warming-up time under average conditions is 15 minutes.
- All limits and tolerances given in this section are calibration guides and should not be interpreted as instrument specifications unless they are also published in chapter 1.6 characteristics.
- Tolerances given are for the instrument under test and do not include test equipment error.
- The most accurate display adjustments are made with a stable, well-focused, low-intensity display. Unless otherwise noted, adjust the Intensity, Focus and Trigger Level controls as needed.

5.2. Recommended test equipment

As indicated in chapter 4.3.
Additional equipment for the checking and adjusting procedure:
Digital multimeter e.g. PM
Trimming tool set e.g. Philips 800 NTX.

5.3. Preliminary settings of the controls

As indicated in chapter 4.2.

No.	Adjustments		Preparatio	on	Voltages to apply	Adjusting	g element	Adjusting Data
		Controls	Description	Position	to X2 (channel A) and X3 (channel B)	Number	Location	
1.	Intensity	\$2 \$8 \$4 \$10 \$5 \$6 R6 R3 R4	X1/X5 CH. A ON/OFF AMPL/DIV. A CH. B ON/OFF AMPL/DIV. B TIME/DIV. INTENS POSITION B] POSITION X]	X1 ON 50 mV/div. ON 50 mV/div. X via A anti-clockwise Adjust spot to centre of screen.		R501	Power Supply Unit	The spot just vanishes
2.	Trace Rotation	\$8 \$10 \$1 \$2 \$6 R2 R4	CH. A ON/OFF CH. B ON/OFF µs/ms X1/X5 TIME/DIV. POSITION A] POSITION X]	ON OFF ms XI .5 Adjust trace to centre of screen	·	R8	Front panel TRACE ROT	Trace in parallel with horizontal graticule line.
3.	DC Balance Channel A (channel B in brackets)	S9 (S11) X2 (X3) S8 (S10) S4 (S5)	AC/DC BNC Channel A(B) ON/OFF AMPL/DIV. A(B)	DC Short-circuited ON 5mV < = > 10mV alternately		R40 (R140)	B2/3 (C2/3)	Trace jump 1/2 div.
3a.	DC Balance Adjustments for 20mV/div.	S9 (S11) X2 (X3) S8 (S10) S4 (S5)	AC/DC BNC CH. A(B) ON/OFF AMPL/DIV. A(B)	Same as in 4 above 20mV < = > 50mV alternately		R60 (R160)	B4(C/4)	Trace jump 1/2 div.
4.	Gain CH. A(B)	S8 (S10)	CH. A(B) ON/OFF	ON	120 mVpp 2 kHz	R83 (R183)	A5(C5)	Amplitude 6 div.
	20 mV/div.	S10 (S8) S1 S6 S12 S14 S15 S4 (S5)	CH. B(A) ON/OFF µs/ms TIME/DIV. A/B INT/EXT NORM/TV AMPL/DIV. A(B)	OFF µs 200 A(B) INT NORM 20mV/div.	Z KIIZ			
4a.	Gain CH. A(B) 5mV/div.	S4 (S5)	All settings, same as above except S4 (S5) AMPL/DIV. A(B)	5mV/div.	30 mVpp - 2kHz	R55 (R155)	B4(C4)	Amplitude 6 div.
5.	Square wave response A (for channel B in brackets)	S1 S2 S6 S9 (S11) S12 S13 S14 S15 S10 (S8)	µs/ms X1/X5 TIME/DIV. AC/DC A/B +/- INT/EXT NORM/TV CH. B(A) ON/OFF	μs X1 200 DC A(B) + INT NORM OFF	Position 2kHz Square wave S4(S5) on X2(X3) 10 mV - 60 mV 0.1 V - 0.6 V 1 V - 6 V 10 V - 60 V	C22 (C122) C23 (C123) C26 (C126) C28 (C128)	A2(C2) B1(C1) B1(C1) B1(C1)	Topside of square wave in parallel with graticule line.
6.	HF response channel A (channel B in brackets)	S1 S4 (S5) S6 S12	μs/ms AMPL/DIV. TIME/DIV. A/B	μs 10mV 0.5 A(B)	120 mV - 1MHz square wave Rise time ≤ 5ns.	C42 (C142)	A5(C5)	Pulse drop < 3% Ringing < 5%
7.	Time Coefficient ms	S1 S2 S6 S7/R7	μs/ms X1/X5 TIME/DIV. VAR TIME/DIV.	ms XI 2 CAL	Apply pulse marks of 2 ms	R416	E6	8 pulses per 8 div.
7a.	Time Coefficient µs	SI S6 S7/R7 S2	μs/ms TIME/DIV. VAR TIME/DIV. X1/X5	μs 2 CAL. X1	Apply pulse marks of 2μs	R411	E6	8 pulses per 8 div.

to apply	Adjusting	Adjusting Data		
nannel A) channel B)	Number	Location		
*	R501	Power Supply Unit	The spot just vanishes	
,	* .	e e		
	* * * * * * * * * * * * * * * * * * * *	Front panel TRACE ROT	Trace in parallel with horizontal graticule line.	
	R8			
	R40 (R140)	B2/3 (C2/3)	Trace jump 1/2 div.	
		e	Trace jump	
	R60 (R160)	B4(C/4)	1/2 div.	
	R83 (R183)	A5(C5)	Amplitude 6 div.	
		*		
- 2kHz	R55 (R155)	B4(C4)	Amplitude 6 div.	
are wave	· · · · · · · · · · · · · · · · · · ·		Topside of square wave in parallel	
X2(X3) 60 mV .6 V /	C22 (C122) C23 (C123) C26 (C126) C28 (C128)	A2(C2) B1(C1) B1(C1) B1(C1)	with graticule line.	
- 1MHz ave e ≼ 5ns.	C42 (C142)	A5(C5)	Pulse drop ≤ 3% Ringing ≤ 5%	
lse marks of	R416	E6	8 pulses per 8 div.	
lse marks of	R411	E6	8 pulses per 8 div.	



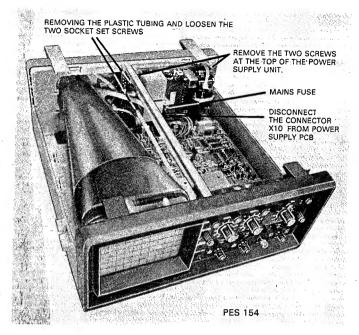


Fig. 6.1 Removing the power supply unit

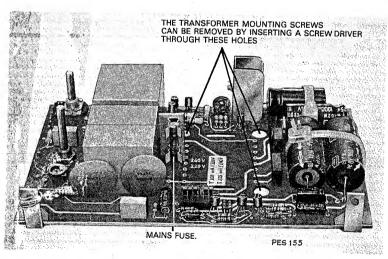


Fig. 6.2 Power supply board

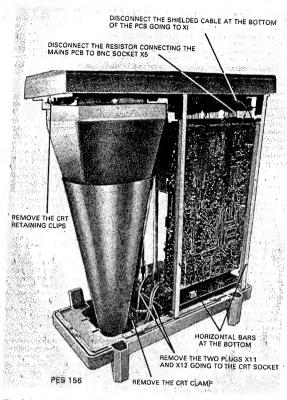


Fig. 6.3 Removing the main PCB and CRT.

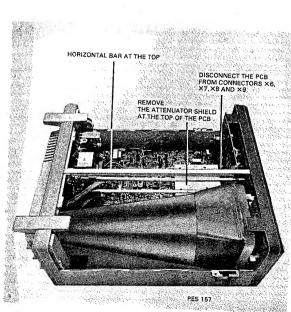


Fig. 6.4 Removing the majn PCB unit.

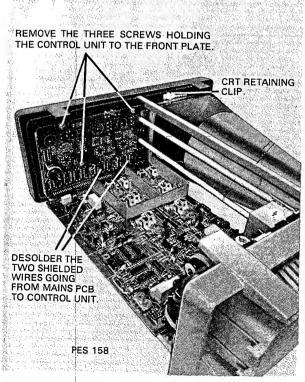


Fig. 6.5 Removing the control unit and CRT.

Fig. 6.3 Removing the main PCB and CRT.

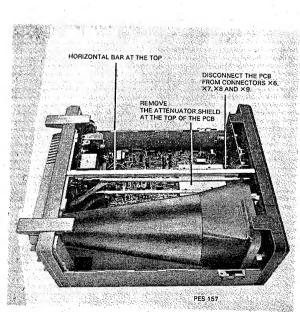


Fig. 6.4 Removing the majn PCB unit.

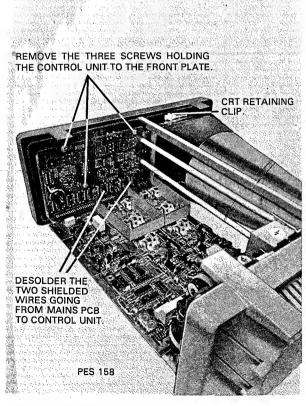


Fig. 6.5 Removing the control unit and CRT.

CORRECTIVE MAINTENANCE

6.1 REPLACEMENTS



WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals may

be live.

The instrument shall be disconnected from all voltage sources before any adjustment, replacement or maintenance and repair during which the instrument will

be open.

If afterward any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved. Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

Standard parts

Electrical and mechanical replacement parts can be obtained through your local Philips organisation or representative. However, many of the standard electronic components can be obtained from other local suppliers.

Before purchasing or ordering replacement parts, check the parts list for value tolerance, rating and description.

NOTE: Physical size and shape of a component may affect instrument performance, particularly at high frequences. Always use direct-replacement components, unless it is known that a substitute will not degrade instrument performance.

Special parts

In addition to the standard electronic components, some special components are used. These components are manufactured or selected by Philips to meet specific performance requirements.

Transistors and integrated circuits

Transistors and I.C.'s (integrated circuits) should not be replaced unless they are actually defective. If removed from their sockets during routine maintenance return them to their original sockets. Unnecessary replacement or switching of semiconductor devices may affect the calibration of the instrument. When a transistor is replaced, check the operation of the part of the instrument that may be affected.



WARNING: Handle silicone grease with care. Avoid getting silicone grease in the eyes. Wash hands thoroughly after use.

Any replacement component should be of the original type or a direct replacement. Bend the leads to fit the sockets and cut the leads to the same length as on the component being replaced.

6.1.1. Replacing the mains fuse

To replace the mains fuse F501 which is located on the power supply board, the top cover should first be removed as described under Sec. 3.2. The mains fuse which is a slow blow fuse, can be removed from its holder, and replaced, if necessary. (See fig. 6.2.).

6.1.2. Replacing Mains Cord or Transformer

- Unlock the "INTENS" and "FOCUS" extension shafts on the power supply side as follows:-

Remove the plastic tubing which covers the coupling between the shaft and the potentiometer. (See fig. 6.1.)

Loosen the two socket set screws on the coupling with an allen key (M3) of 1.5mm across flats.

WARNING: The intensity and focus potentiometers and their shafts are at 2000V with respect to earth. So a protective plastic tubing covers these shafts. While unlocking the "INTENS" and "FOCUS" extension shafts, the instrument has to be disconnected from voltage sources.

- Remove the four transformer mounting screws. (See fig. 6.2.).
- Remove the two screws at the top of the power supply unit holding it to the rear cover (See fig. 6.1.)
- Remove the two screws at the bottom of the power supply unit holding it to the horizontal bars.
- Now the power supply unit is free and the back side of this unit is now accessible. The transformer can now be desoldered and replaced if necessary. Before replacing the transformer the thermal fuse on the transformer can be checked.
- To replace the mains cord, desolder the two wires of the mains cord from switch S3/R6 and also the safety earth terminal.
- Remove the grommet at the rear.
- Solder the mains cord and fix the grommet.

6.1.3. Removing and replacing components on the main PCB

To remove and replace components on the main PCB, the top and bottom cover should first be removed as described in Sec. 3.2.

- All components on the main PCB can now be accessed.

To replace the main PCB, the following procedure should now be adopted:-

- Remove the screw holding the attenuator shield at the top of the PCB and remove the shield (See fig. 6.4.)
- Remove the attenuator shield at the bottom of the PCB by removing the screw which holds it to the PCB.
- Remove the six screws holding this PCB to the horizontal bars.
- Remove the two plugs X11 and X12 going to the CRT socket.
- Disconnect the connector X10 from the power supply PCB (Fig. 6.1.)
- Disconnect the resistor connecting the main PCB to BNC socket X5. (Fig. 6.3.)
- Disconnect the shielded cable at the bottom of the PCB going to X1 (Fig. 6.3.)
- Desolder the two shielded wires going from the main PCB to the control unit (Fig. 6.5.)
- Slide the PCB towards the rear of the instrument and thus disconnect this PCB from connectors X6, X7, X8 and X9. (See fig. 6.4.)
- The main amplifier board can now be lifted and replaced.

6.1.4. Removing and replacing components on the control unit.

- To remove or replace components on the control unit, first remove the main PCB as described in Sec. 6.1.3.
- Remove the three screws holding the control unit to the front plate. (Fig. 6.5.)
- Remove all the knobs on the front panel.

- Unlock the "INTENS" and "FOCUS" extension shafts as mentioned in Sec. 6.1.2.
- Desolder the two wires (red and yellow) which go to the trace rotation coil of the CRT.
- The control unit is now free and any component can be replaced.

6.1.5. Replacing the CRT.

To replace the CRT, the main PCB is first removed following the procedure described in Sec. 6.1.3.

- Desolder the two wires (red and yellow) which go to the trace rotation coil of the CRT from the control unit. Remove the two screws connecting the front plate assembly to the horizontal bars at the bottom. Care should be taken while removing the screws as it may damage connector X9. (Fig. 6.3.). Also remove the screw connecting the front plate assembly to the horizontal bar at the top. (Fig. 6.4.).
- Push the front plate assembly forward.
- The two CRT retaining clips holding the CRT to the front plate will come out (Fig. 6.5.).
- Loosen the CRT clamp at the rear of the instrument. (Fig. 6.3.).
- Push the CRT with its metal shield forward and disconnect the CRT from its socket. The CRT is now free.
- Remove the tape holding the trace rotation coil to the neck of the CRT. The trace rotation coil can now be removed.
- The trace rotation coil can now be inserted around the neck of the new CRT and fixed firmly with tape.
- Insert the CRT with the metal shield into its socket and put the CRT in its place.
- Now put the bottom cover.
- Bring the front plate forward, plug it to the bottom cover and tighten the screws holding it to the horizontal bars.
- Position the CRT properly with the retaining clips. Tighten the CRT clamp. Solder the trace rotation coil back.
- Fix the main PCB in its place and tighten the six screws holding it to the horizontal bars. Resolder all the wires that were removed at the time of removing the PCB.
- Fix the top cover.

NOTE: Take care of the CRT retaining clips and the CRT filter.

WARNING: Handle the CRT carefully. Rough handling or scratching can cause the CRT to implode.

6.1.6. Removing and replacing components on power supply board.
Refer section 6.1.2.

6.2 SPECIAL TOOLS

6.2.1. Trimming Tool Kit (Type 800/NTX)

This useful kit contains 3 twin-coloured holders, 2 extension holders and 21 interchangeable trimming pins. The wide variety of pin allows almost every type of trimming function to be carried out in instruments to be calibrated (e.g. measuring instruments, radio and T.V. sets). Ordering number 4822 310 50015.

(A spare set containing the 8 most commonly used pins is available under the ordering number 4822 310 50016).



Fig. 6.6 Trimming Tool kit.

6.3. RECALIBRATION AFTER REPAIR

After any electrical component has been replaced the calibration of that particular circuit should be checked, as well as the calibration of other closely related circuit. Since the power supply affects all circuits, calibration of the entire instrument should be checked if work has been done in the power supply or if the transformer has been replaced.

6.4. INSTRUMENT REPACKAGING

If the instrument is to be shipped to a Service Centre for service or repair, attach a tag showing owner (with address) and the name of an individual at your firm that can be contacted. The Service Centre needs the complete instrument serial number and a fault description.

Save and re-use the packing in which your instrument was shipped. If the original packing is unfit for use or not available, repack the instrument in such a way that no damage during transport occurs.

6.5. TROUBLE-SHOOTING

6.5.1. Introduction

The following information is provided to facilite trouble shooting. Information contained in other sections of this manual should be used along with the following information to aid in locating the defective component. An understanding of the circuit operation is helpful in locating troubles, particularly where integrated circuits are used. Refer to the Circuit Description section for this information.

6.5.2. Trouble-Shooting hints

If a fault appears, the following test sequences can be used to find the defective circuit part:

- Check if the settings of the controls of the oscilloscope are correct. Consult the operating instructions in the Operating manual.
- Check the equipment to which the oscilloscope is connected and the interconnection cables.
- Check if the oscilloscope is well-calibrated. If not refer to section 5 (checking and adjusting).
- Visually check the part of the oscilloscope in which the fault is suspected. In this way, it is possible to find faults such as bad soldering connections, bad interconnection plugs and wires, damaged components or transistors and IC's that are not correctly plugged into their sockets.
- Location of the circuit part in which the fault is suspected: the symptom often indicates this part of the circuit. If the power supply is defective the symptom will appear in several circuit parts.

After having carried out the previous steps, individual components in the suspected circuit parts must be examined:

- Transistors and diodes. Check the voltage between base and emitter (0.7 Volt approx. in conductive state) and the voltage between collector and emitter (0.2 Volt approx. in saturation) with a voltmeter or oscilloscope. When removed from the p.c.b. it is possible to test the transistor with an ohmmeter since the base/emitter and base/collector junctions can be regarded as diodes. Like a normal diode, the resistance is very high in one direction and low in the other direction. When measuring take care that the current from the ohmmeter does not damage the component under test.

 Replace the suspected component by a new one if you are sure that the circuit is not in such a condition that the new one will be damaged.
- Integrated circuit. In circuit testing can be done with an oscilloscope or voltmeter. A good knowledge of the circuit part under-test is essential. Therefore first read the circuit description in section 2.
- Capacitors. Leakage can be traced with an ohmmeter adjusted to the highest resistance range. When testing take care of polarity and maximum allowed voltage. An open capacitor can be checked if the response for AC signals is observed. Also a capacitance meter can be used: compare the measured value with value and tolerance indicated in the parts list.

- Resistors. Can be checked with an ohmmeter after having unsoldered one side of the resistor from the p.c.b. Compare the measured value with value and tolerance indicated in the parts list.
- Coils and transformers. An ohmmeter can be used for tracing an open circuit. Shorted or partially shorted windings can be found by checking the wave-form response when HF signals are passed through the circuit. Also an inductance meter can be used.

NOTE: If a component must be replaced always use a direct-replacement. If not available use an equivalent after carefully checking that it does not degrade the instrument's performance. See also section 6.1. (replacement).

After replacement of a component the calibration of the instrument may be affected due to component tolerances. If necessary do the required adjustments.

6.6. ADAPTION OF MAINS (LINE) VOLTAGE

Before opening any part of the instrument it must be disconnected from all voltage sources. Access to the power supply unit:

- Remove the two handle mounting screws.

- Bend the handle brackets outwards and remove it.

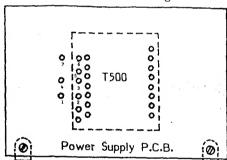
- Remove the two cabinet mounting screws which become visible now.

- Press firmly the two buttons of the rear cover until the click (the top cover will lift for approx. 2 mm).

- Now lift vertically the top cover out of the front and rear-cover.

- The power supply board is accessible now to adapt the mains voltage.

- Connections should be changed as follows:



NOTE: FOR 240V CONNECT 3-4 & 6-7

FOR 220V CONNECT 3-4 & 5-6

FOR 120V CONNECT 2-3, 6-7 & 1-4

6.7. SAFETY INSPECTION AND TESTS AFTER REPAIR AND MAINTENANCE IN THE PRIMARY CIRCUIT.

6.7.1 General directives

- Take care that the creepage distances and clearances have not been reduced.
- Before soldering, the wires should be bent through the holes of solder tags, or wrapped around the tag in the form of an open U, or, wiring shall be rigidly maintained by cable clamps or cable lacing.
- Replacing all insulating guards and plates.

6.7.2 Safety components

Components in the primary circuit may only be renewed by components selected by Philips.

6.7.3 Checking the protective earth connection

The correct connection and condition is checked by visual control and by measuring the resistance between the protective lead connection at the plug and the cabinet/frame. The resistance shall not be more than 0.1Ω . During measurement the mains cable should be removed. Resistance variations indicate a defect.

6.7.4 Checking the insulation resistance

Measure the insulation resistance at U = 500 V dc between the mains connections and the protective lead connections. For this purpose set the mains switch to ON. The insulation resistance shall not be less than 2 M Ω .

NOTE: 2 M Ω a minimum requirement at 40°C and 95% Relative Humidity. Under normal conditions the insulation resistance should be much higher (10...20 M Ω).

6.7.5 Checking the leakage current

The leakage current shall be measured between each pole of the mains supply in turn, and all accessible conductive parts connected together (including the measuring earth terminal). The leakage current is not excessive if the measured currents from the mentioned parts is $\leqslant 3.5$ mA rms.

6.7.6 Voltage test

The instrument shall withstand, without electrical breakdown, the application of a test voltage between the supply sircuit and accessible conductive parts that are likely to become energized. The test potential shall be 1500 V rms at supply-circuit frequency, applied for one second. The test shall be conducted when the instrument is fully assembled, and with the primary switch in the ON position.

During the test, both sides of the primary circuit of the instrument are connected together and to one terminal of the voltage test equipment; the other voltage test equipment terminal is to be connected to the accessible conductive parts.

6.8. SURVEY OF MEASURING POINTS

- To make fault finding easy, test points M1 to M26 are given below. These test points can be located on the PCB with the help of fig. 6.8 and fig. 6.9.
- Apply a sine wave signal of 120 mV peak to peak to YA(YB) input.
- Set AMPL/DIV in 20 mV position.

Measuring point	Location	Values to be measured	Remarks
M1	A3	100 mV p.p. superimposed on 1.5 V DC approx.	Signal on Channel A.
M2	A3	1.5 V DC approx.	
М3	C3	100 mV p.p. superimposed on 1.5 V DC approx.	Signal on Channel B.
M4	C3	1.5 V DC approx.	
M5/M6	A4	1 V p.p. superimposed on 8.1 V DC approx.	Signal on Channel A.
M7/M8	C4	1 V p.p. superimposed on 8.1 V DC approx.	Signal on Channel B.
М9	В6	36 V p.p. on 0 VDC with trace at the centre	Signal on Channel A.
M10	A/B6	36 V p.p. on 0 VDC with trace at the centre	Signal on Channel B.
M11	E5	4 V DC (TTL high voltage) when CH A and CH B is ON and switch S1 in us position. In ms position of S1, the Chopper frequency of 120 KHZ will be visible.	
M12	.D5	TTL high voltage when CH.A is ON and CH.B is OFF	
M13	D5 , .	TTL high voltage when CH.A is OFF and CH.B is ON	
M14	D5	Blanking pulse varying with TIME/DIV position.	
M15	E2	1.2 V p.p. square wave	
M16	E4	4.5 V p.p. square wave	
M17	E3	1.2 V p.p. square wave with NORM/TV switch in TV position	
M18	D4	With switch S1 in us position needle like pulses and in ms position, no signal	
M19	D/E4	No trigger signal: square wave depending on TIME/DIV position	
		Trigger Signal: Square wave depending on TIME/DIV position and trigger signal.	
M20	D6	Sweep voltage (sawtooth) - 1 V up to + 4.2 V	

Measuring	Location	Values to be measured	Remarks
point			
M21	D6	Sawtooth	
M22	D6	2.5 V p.p. square wave	
M23	C6	Position control voltage - 1.5 V DC up to + 4.5 V DC	
M24/M25	C6	Sawtooth 100 V p.p. (in X via A position - 60 V up to + 40 V)	* * .
M26	Power Supply (R 528)	Blanking pulse to CRT - 32 V p.p. square wave superimposed on 35 V DC approx.	

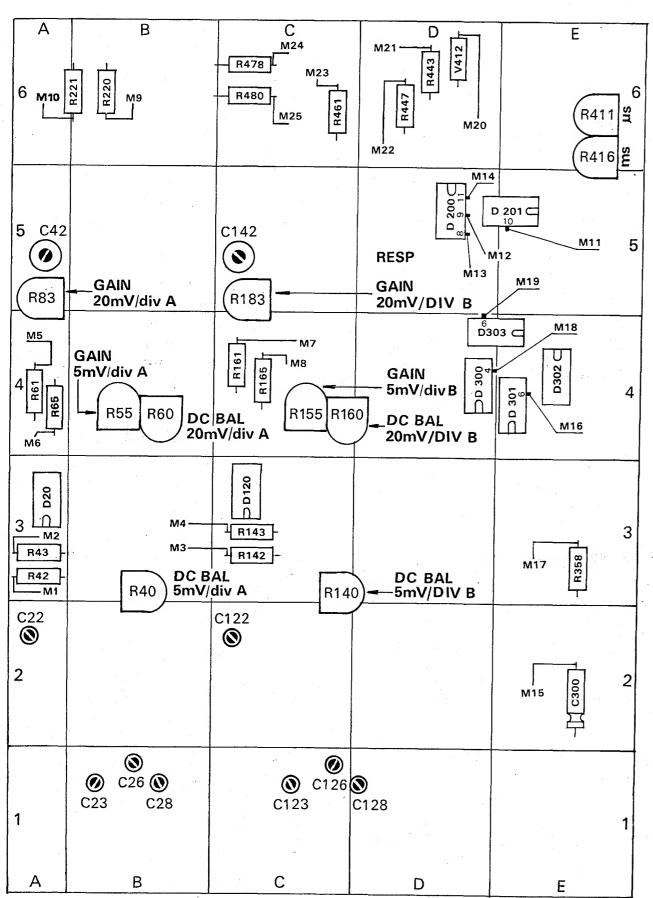
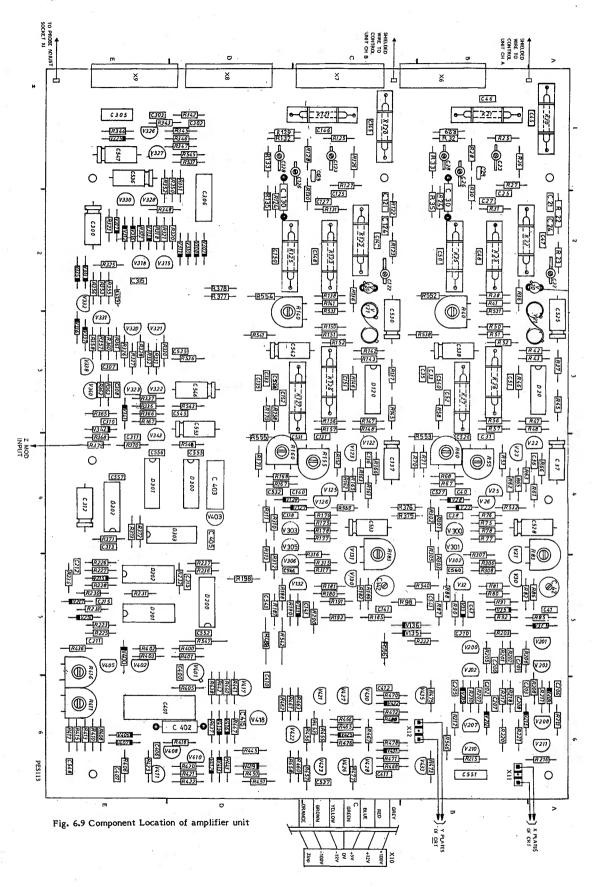


Fig. 6.8 Survey of measuring points

PES 106



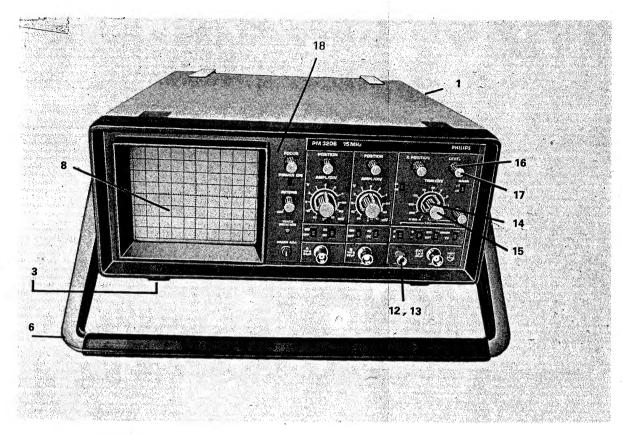
6.9 PIN CONNECTIONS (Refer to Fig. 8.1.)

6.9	PIN CONNECTIONS (Refer to	o Fig. 8.1.)	
	1. AMPLIFIER UNIT	CONNECTOR NO./ PIN NO.	CONTROL UNIT
6.9.1	Coil K20, C45 Coil K21, C46 Coil K22, C47 Coil K25, C50 Coil K23, C48 Coil K27, C52 R70, C33 R71, C35 Coil K26, C51 +12V -12V	X6-1 X6-2 X6-3 X6-4 X6-5 X6-6 X6-7 X6-8 X6-9 X6-10 X6-11	V37,V40,S4-3 V57,V58,V60 V42,V45,V60,S4-6 V52,V55,V57,S4-12 V47,V50,V58,S4-9 V38,V43,V48,V53 R2 R2 V41,V46,V51,V56 +12V -12V
	Coil K120, C145 Coil K121, C146 Coil K122, C147 Coil K125, C150 Coil K123, C148 Coil K127, C132 C133 R171, C135 Coil K126, C151 D301-9 D300-2.12	X7-1 X7-2 X7-3 X7-4 X7-5 X7-6 X7-7 X7-8 X7-9 X7-10 X7-11	V137,V140,S5-3 V157,V158,V160 V142,V145,V160, S5-6 V152,V155,V157,S5-12 V147,V150,V158, S5-9 V138,V143,V148,V153 R3 R3 V141,V143,V151,V156 S13-13 S15-2
	V321, R329 V307, V310 V300, V307, V308 V303, V310, V311 R463 V421, R453 R456 R441,R440,R442,V215 V330	X8-1 X8-2 X8-3 X8-4 X8-5 X8-6 X8-7 X8-8 S8-9 X8-10 X8-11	R5 S14-10 S12-10 S12-12 R4 S2-8 S2-7 S6-10, K28 S14-12
	V331, V333, V337 V332, V335, V336 V342, R367 OV V333, V335 R410, R411 R226, C212 R415, R416, R417 R436 R223, C211	X9-1 X9-2 X9-3 X9-4 X9-5 X9-6 X9-7 X9-8 X9-9 X9-10 X9-11	S13-1 S13-3 S1-10 OV S15-5 S1-1 S8-6 S1-3 R425-R435 S10-6
5.9.2 5.9.3	POWER SUPPLY UNIT (Refer +100V +12V +5V OV -12V -100V Z input (D202-13) CRT SOCKET		+100V +12V +5V OV -12V -100V R522
. .	R220 R221 R478 R480	X11-1 X11-2 X12-1 X12-2	CRT Pin 7 CRT Pin 9 CRT Pin 11 CRT Pin 12

7. PARTS LISTS (SUBJECT TO ALTERATION WITHOUT NOTICE).

7.1. Mechanical Parts (See fig. 7.1 & 7.2)

 ccanicai	. di 10 (000 11)			
Item	Qty.	Ordering Co	ode	Description
1.	2	5322 447	90604	Top and bottom cover.
2.	1	5322 462	50314	CRT shield.
3.	4	5322 462	40643	Foot.
4.	2	5322 535	91665	Extension shaft
5.	1	5322 255	40541	CRT socket.
6.	1	5322 498	50157	Handle Assembly.
7.	2	5322 522	31739	Handle gear.
8.	1	5322 466	70513	CRT filter.
9.	2	5322 462	44398	Retaining clip.
10.	1	5322 290	40191	CRT clamp assembly.
11.	1	5322 325	50101	Grommet.
12.	1	5322 535	80692	Earth stud with threaded end.
13.	1	5322 506	41004	Knurled nut.
14.	3	5322 414	30063	Switch knob. Dia. 18.7 mm.
15.	3	5322 447	90385	Brown cover for the knob. Dia. 18.7 mm.
16.	5	5322 414	30064	Control knob. Dia. 10 mm.
17.	5	5322 414	70036	Brown cover with line for control knob of dia. 10 mm.
18 •	1 .	5322 447	90602	Front Plate Assembly
19•	1 .	5322 447	90384	Rear Plate Assembly



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Fig. 7.1 Mechanical parts, front view.

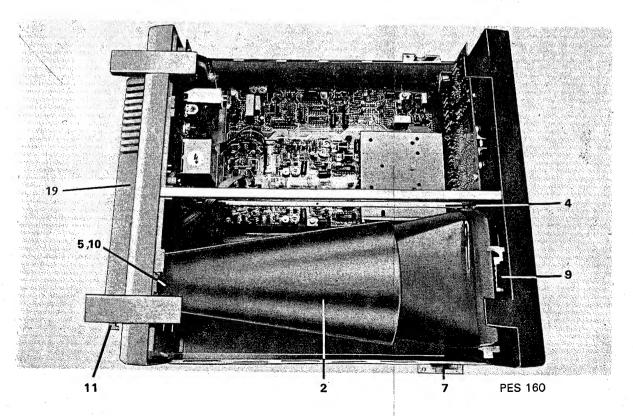


Fig. 7.2 Internal view, top cover removed

7.2 ELECTRICAL PARTS

F	POSNR	DESCRIPTION				ORE	ERI	NG CO	DE
ì		CAPACITORS			. ((-				-
С	20	CAPACITOR, POLYCAP	27N	10%	400V	4822	121	42066	
C	21	CAPACITOR, CERCAP	56P	2%	500V	5322	122	32791	
Ę	22	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
C	23	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
C	24	CAPACITOR, CERCAP	3P9		500V	5322	122	32789	
С	25	CAPACITOR, CERCAP	18P	2%	100V	4822	122	31061	
C	26	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
С	27	CAPACITOR, CERCAP	330PF	2%	100V	4822	122	31353	
C	28	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
С	29	CAPACITOR, CERCAP	1P5		100V	5322	122	32101	
C	30	CAPACITOR,,POLYCAP	3N3	2%	63V	5322	121	54049	
C		CAPACITOR, CERCAP	56P	2%	100V	4822	122	32027	
С		CAPACITOR, CERCAP	22N		63V	4822	122	30103	
С	35	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
С	36	CAPACITOR, CERCAP	22N -		63V	4822	122	30103	
С		CAPACITOR, ELCAP	33UF		16V	322	124	21431	
С		CAPACITOR, CERCAP	22N ·		63V	4822	122	30103	
С		CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C		CAPACITOR, CERCAP	680P	10%	100V	4822	122	30053	
С	42	CAPACITOR, TRIMCAP	22P		250V	4822	125	50045	
C	43	CAPACITOR, CERCAP	22P	2%	100V	5322	122	32242	
C	45	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
, C	46	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
С		CAPACITOR, CERCAP	22N		63V	4822	122	30103	
С	48	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	50	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	. 51	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	53	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	120	CAPACITOR, POLYCAP	27N	10%	400V	4822	121	42066	
C	121	CAPACITOR CERCAP	56P	2%	500V	5322	122	32791	
, or C	122	CAPACITOR TRIMCAP	5P5		400V	5322	125	54027	
С	123	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
С	124	CAPACITOR, CERCAP	3P9		500V	5322	122	32789	
, C	125	CAPACITOR, CERCAP	18P	2%	100V	4822	122	31061	
С	126	CAPACITOR TRIMCAP	5P5		400V	5322	125	54027	
С	127	CAPACITOR, CERCAP	330PF	2%	100V	4822	122	31353	
С	128	CAPACITOR, TRIMCAP	5P5		400V	5322	125	54027	
C	129	CAPACITOR, CERCAP	1P5		100V	5322	122	32101	
C	130	CAPACITOR, POLYCAP	3N3	2%	63V	5322	121	54049	
С	131	CAPACITOR, CERCAP	56P	2%	100V	4822	122	32027	
C	133	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	135	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	136	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	137	CAPACITOR, ELCAP	33UF		16V	5322	124	21431	
С	138	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
C	140	CAPACITOR, CERCAP	22N		63V	4822	122	30103	
С	141	CAPACITOR, CERCAP	6 80PF		100V	4822	122	30053	
С	142	CAPACITOR, TRIMCAP	22PF		250V	4822	125	50045	
. C	143	CAPACITOR, CERCAP	22PF	2%	100V	5322	122	32242	
C	145	CAPACITOR, CERCAP	22N		63V	4822	122	30103	

POSNR	DESC	RIPT	ION
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ORDERING CODE

C	146	CAPACITOR CERCAP	22N		63V	4822	122	30103	
C	147	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	148	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	150	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	151	CAPACITOR CERCAP	22N		63V	4822	122	30103	
_	450	CARACITOR GERCAR	001		(7)/	4022	122	30103	
С	152	CAPACITOR CERCAP	22N		63V	4822			
С	200	CAPACITOR CERCAP	1P5		100V	5322	122	32101	
С	201	CAPACITOR CERCAP	1P5		100V	5322	122	32101	
Ç	202	CAPACITOR CERCAP	1P5		100V	5322	122	32101	
C	203	CAPACITOR CERCAP	1P5		100V	5322	122	32101	
								70015	
С	205	CAPACITOR CERCAP	100P	2%	500V	5322	122	32265	
С	206	CAPACITOR CERCAP	100P	2%	500V	5322	122	32265	
С	207	CAPACITOR CERCAP	10N		100V	4822	122	31414	
С	208,	CAPACITOR CERCAP	10N		100V	4822	122	31414	
C	210	CAPACITOR CERCAP	22N		63V	4822	122	30103	
	2,0	on north orthogram	2-2-1		0,7				
С	211	CAPACITOR CERCAP	22N		- 63V	4822	122	30103	
С	212	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	213	CAPACITOR CERCAP	01N	10%	100V	4822	122	30027	
C	215	CAPACITOR CERCAP	3N3	10%	1007	4822	122	30099	
			22NF	1070	16V	5322	124	21429	
С	300	CAPACITOR ELCAP	ZZINF		104	222	124	41447	
C ·	301	CAPACITOR CERCAP	22N		63V	4822	122	30103	
c	302	CAPACITOR CERCAP	39PF	2%	500V	5322	122	32788	
				2%	500V	5322	122	32792	
С	303	CAPACITOR CERCAP	180P	20,					
С	304	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	305	CAPACITOR POLYCAP	220N	10%	100V	4822	121	40232	
C	306	CAPACITOR POLYCAP	1UF	10%	100V	5322	121	40197	
				1070					
С	307	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	308	CAPACITOR CERCAP	4N7	10%	100V	4822	122	30128	
С.	310	CAPACITOR CERCAP	4N7	10%	100V	4822	122	30128	
С	311	CAPACITOR CERCAP	3N9	10%	100V	4822	122	30098	-
С	<i>3</i> 12	CAPACITOR ELCAP	33U		16V	5322	124	21431	
				400/		4822			
С	313	CAPACITOR CERCAP	100P	10%	100V		122	32024	
С	315	CAPACITOR CERCAP	100P	10%	100V	4822	•	32024	
C	400	CAPACITOR CERCAP	150P	2%	100V	4822	122	31413	
С	401	CAPACITOR POLYCAP	2U2	2%	100V	5322	121	44246	
_	4.00	CARACITOR POLYCAR	01.10	4.0/	(7)/	4822	121	50415	
Ċ	402	CAPACITOR POLYCAP	2U2	1%	63V				
Ċ	403	CAPACITOR POLYCAP	330N	5%	100V	5322	121	50905	
С	405	CAPACITOR CERCAP	1NF	10%	100V	4822	122	30027	
C	406	CAPACITOR CERCAP	1NF	. 10%	1טטע	4822	122	30027	
С	407	CAPACITOR CERCAP	22N		63V	4822	122	30103	
					50014		400	74047	
С	411	CAPACITOR CERCAP	OP68		500V	4822	122	31213	
С	412	CAPACITOR CERCAP	OP68		500V	4822	122	31213	
С	500	CAPACITOR POLYCAP	150N	10%	1.5KV	5322	121	44329	
С	501	CAPACITOR POLYCAP	150N	10%	1.5KV	5322	121	44329	
C	502	CAPACITOR POLYCAP	150N	10	1.5KV	5322	121	44329	
				. 5					
С	503	CAPACITOR POLYCAP	300N		630V	4822	121	40344	
C	505	CAPACITOR CERCAP	22N		63V	4822	122	30103	
С	506	CAPACITOR POLYCAP	10N		2KV	5322	121	41603	
C	507	CAPACITOR POLYCAP	0.22U	10%	100V	4822	121	40232	
С			100N	10%	100V	5322	121	40323	
Ų	508	CAPACITOR POLYCAP	10014	1 U /0	1007	1166	141	70747	

POSNR DESCRIPTION ORDERING CODE C 510 CAPACITOR ELCAP 1000U 25V 5322 124 21428 C 511 CAPACITOR ELCAP 33U 16V 5322 124 21428

С	510	CAPACITOR	ELCAP	1000U		25V	5322	124	21428	
С	511	CAPACITOR	ELCAP	33U		16V	5322	124	21428	
С	512	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	513	CAPACITOR	ELCAP	1000U		25V	5322	124	21428	
С	515	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	516	CAPACITOR	ELCAP	100U		160V	5322	124	24221	
С	517	CAPACITOR	ELCAP	100U		160V	5322	124	24221	
С	518	CAPACITOR	LCCCAP	10N		3KV	5322	122	54009	
С	520	CAPACITOR	METCAP	047U		250V	4822	121	40239	
C	521	CAPACITOR	CERCAP	22PF		100V	4822	122	31063	
С	550	CAPACITOR	POLYCAP	68N		250V	4822	121	41156	
С	535	CAPACITOR	ELCAP	33U		-16V	5322	124	21431	
С	526	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	527	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	528	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	530	CAPACITOR		33U		16V	5322	124	21431	
C	531	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	532	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	533	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	535	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	536	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	537	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	538	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	540	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	541	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	542	CAPACITOR		33U	÷	16V	5322	124	21431	
С	543	CAPACITOR		22N		63V	4822	122	30103	
С	545	CAPACITOR		22N		63V	4822	122	30103	
С	546	CAPACITOR		33U		16V	5322	124	21431	
С	547	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	548	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	551	CAPACITOR	POLYCAP	68N	10%	250V	4822	121	41156	,
С	552	CAPACITOR	POLYCAP	0.1U	20%	50V	4822	121	42456	
С	553	CAPACITOR	ELCAP	33U		16V	5322	124	21431	
С	555	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	556	CAPACITOR		22N		63V	4822	122	30103	
C .	557	CAPACITOR		22N		63V	4822	122	30103	
С	558	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	560	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	561	CAPACITOR	CERCAP	22N		63V	4822	122	30103	
С	410	CAPACITOR	CERCAP	22N		63V	4822	122	30103	

F	2051	AID.	DESCRIPTION			ORDERING CODE
_	03	1477				
			RESISTORS			5700 404 00775
R		1	CARB.TRACK POTMETER	2M2		5322 101 20735 5322 101 24159
R		2	CARB. TRACK POTMETER	LIN 220E		
R		3	CARB. TRACK POTMETER	LIN 220E		
R		4	CARB. TRACK POTMETER	LIN 10K		•
		5	CARB. TRACK POTMETER	LIN 10K		
R		6	CARB, 220K POT + DPST SW			5322 101 40132 5322 101 40126
٠R		7	CARB, TRACK LIN 10K + SPSTSW			5322 101 24122
R	:	8	CARB. TRACK POTMETER	LIN 10K	4.0/	4822 116 51235
R		20	RESISTOR METAL FILM	1K	1%	5322 116 55549
R	:	21	RESISTOR METAL FILM	100E	1.%	•
R	2	22	RESISTOR METAL FILM	196K	1%	5322 116 50576
R		23	RESISTOR METAL FILM	806K	1%	5322 116 51369
R		24	RESISTOR METAL FILM	2E7A	1%	5322 116 52571
F		25	RESISTOR STD. FILM	2M43	1%	5322 116 53467
F		26	RESISTOR METAL FILM	348K	1%	5322 116 55499
F	,	27	RESISTOR METAL FILM	348K	1%	5322 116 55499
F		28	RESISTOR STD. FILM	2M43	1%	5322 116 53467
F		29	RESISTOR METAL FILM	1K	1%	4822 116 51235
	₹	30	RESISTOR METAL FILM	412K	1%	5322 116 55424
	` ?	31	RESISTOR METAL FILM	23K7	1%	5322 116 54646
-		32	RESISTOR STD. FILM	2M43	1%	5322 116 53467
	₹	33	RESISTOR METAL FILM	576K	1%	5322 116 53466
	٦ -	35	RESISTOR METAL FILM	2K43	1%	5322 116 54004
	₹	36	RESISTOR STD. FILM	51E	5%	4822 116 52369
	२ २	37	RESISTOR STD. FILM	51E	5%	4822 116 52369
	R.	38	RESISTOR METAL FILM	12K1	1%	5322 116 50572
	R	40	POTMETER TRIMMER	1K		4822 100 10037
	R	41	RESISTOR METAL FILM	12K1	1%	5322 116 50572
	R	42	RESISTOR STD. FILM	10E	5%	4822 116 52332
	R ·	43	RESISTOR STD. FILM	10E	5%	4822 116 52332
	R	45	RESISTOR STD. FILM	1K5	5%	4822 116 52399
	R	47	RESISTOR METAL FILM	191E	1%	5322 116 54495
	R	48	RESISTOR METAL FILM	191E	1%	5322 116 54495
	R	50	RESISTOR METAL FILM	6K49	1%	5322 116 54603
	R	51	RESISTOR METAL FILM	5K11	1%	5322 116 54595
	R	52	RESISTOR METAL FILM	1K21	1%	5322 116 54557
	R	53	RESISTOR METAL FILM	51E1	1%	5322 116 54442
	R	55	POTMETER TRIMMER	100E		5322 101 14011
	R	56	RESISTOR METAL FILM	274E	1%	5322 116 54504
	R	57	RESISTOR METAL FILM	51E1	1%	5322 116 54442
	-	58	RESISTOR STD. FILM	110K	5%	4822 116 52455
	R	60	POTMETER TRIMMER	100K		4822 100 10052
	R R	61	RESISTOR METAL FILM	3K74	1%	5322 116 54588
	R	62	RESISTOR METAL FILM	3K74	1%	5322 116 54588
	r R	63	RESISTOR METAL FILM	750E	1%	4822 116 51234
	R	65	RESISTOR METAL FILM	750E	1%	4822 116 51234
	R	66	RESISTOR STD. FILM	560E	5%	4822 110 73101
	R	67		4K3	5%	4822 116 52424
	R	68		3K6	5%	4822 116 52419
	R	70		430E	5%	4822 116 52423

POSNR	DESCRIPTION

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R	71	RESISTOR STD. FILM		430E		5%	4822	116	52423	
R	73	RESISTOR METAL FILM		2K61		1%	5322	116	50671	
R	75	RESISTOR STD. FILM		1K2		5%	4822	110	73109	
R	76	RESISTOR STD. FILM		1K2			4822	110	73109	
R	77	RESISTOR STD. FILM		10E		5%	4822	116	52332	
R	78	RESISTOR STD. FILM		10E		5%	4822	116	52332	
R	80	RESISTOR METAL FILM		453E		1%	5322	116	54523	
R	81	RESISTOR METAL FILM		453E	**	1%	5322	116	54523	
R	82	RESISTOR METAL FILM		634E		1%	5322	116	54531	
R	83	POTMETER TRIMMER		470E			5322	101	14047	
R	85	RESISTOR METAL FILM		33K2		1%	4822	116	51259	
R	86	RESISTOR METAL FILM		15E		1%	4822	116	51221	
R	87	RESISTOR METAL FILM		90E9		1%	5322	116	54466	
R	- 88	RESISTOR METAL FILM		8K25		1%	5322	116	51498	
R	90	RESISTOR METAL FILM		110K		1%	5322	116	54701	
R	91	RESISTOR STD. FILM		51E		5%	4822	116	52369	
R	92	RESISTOR STD. FILM		51E		5%	4822	116	52369	
R	98	RESISTOR STD. FILM		1E0		5%	4822	116	52385	
R	120	RESISTOR METAL FILM		1K		1%	4822	116	51235	
R	121	RESISTOR METAL FILM		100E		1%	5322	115	55549	
R	122	RESISTOR METAL FILM		196K		1%	5322	116	50576	
R	123	RESISTOR METAL FILM		806K		1%	5322	116	51369	
R	124	RESISTOR METAL FILM		2E74		1%	5322	116	52571	
R	125	RESISTOR STD. FILM		2M43		1%	5322	116	53467	
R	126	RESISTOR METAL FILM		348K		1%	5322	116	55499	
R	1:27	RESISTOR METAL FILM		348K		1%	5322	116	55499	
R	128	RESISTOR STD. FILM		2M43		1%	5322	116	53467	
R	129	RESISTOR METAL FILM		1K0		1%	4822	116	51235	
R	130	RESISTOR METAL FILM		412K		1%	5322	116	55424	
R	131	RESISTOR METAL FILM		23K7		1%	5322	116	54646	
R.	132	RESISTOR METAL FILM		2M43		1%	5322	116	53467	
R	133	RESISTOR METAL FILM		576K		1%	5322	166	53466	
R	135	RESISTOR METAL FILM		2K43		1%	5322	116	54004	
R	136	RESISTOR STD. FILM		51E		5%	4822	116	52369	
R	137	RESISTOR STD. FILM		51E		5%	4822	116	52369	
R	138	RESISTOR METAL FILM		12K1		1%	5322	116	50572	
R	140	POTMETER TRIMMER		1K			4822	100	10037	
R	141	RESISTOR METAL FILM	•	12K1		1%	5322	116	50572	
R	142	RESISTOR STD. FILM		10E		5%	4822	116	52332	
R	143	RESISTOR STD. FILM		10E		5%	4822	116	52332	
R	145	RESISTOR STD. FILM		1K5		5%	4822	116	52399	
R	146	RESISTOR STD. FILM		1K5		5%	4822	116	52399	
R	147	RESISTOR METAL FILM		191E		1%	5322	116	54495	
R	148	RESISTOR METAL FILM		191E		1%	5322	116	54495	
R	150	RESISTOR METAL FILM		6K49		1%	5322	116	54603	
R	151	RESISTOR METAL FILM		5K11		1%	5322	116	54595	
R	152	RESISTOR METAL FILM		1K21		1%	5322	116	54557	
R	153	RESISTOR METAL FILM		51E1		1-%	5322	116	54442	
R	155	POTMETER TRIMMER		100E			5322	101	14011	
R	156	RESISTOR METAL FILM		274E		1%	5322	116	54504	

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			*.				
R	157	RESISTOR METAL FILM	51E1	1%	5322	116	54442
R	158	RESISTOR STD. FILM	110K	5%	4822	116	52455
R	160	POTMETER TRIMMER	100K	2,0	4822	100	10052
R	161	RESISTOR METAL FILM	3K74	1%	5322	116	54588
R	162	RESISTOR METAL FILM					
K	102		3K74	1%	5322	116	54588
R	163	RESISTOR METAL FILM	750E	1%	4822	116	51234
R	165	RESISTOR METAL FILM	750E	1%	4822	116	51234
R	166	RESISTOR STD. FILM	560E	5%	4822	110	73101
R	167	RESISTOR STD. FILM	4K3	5%	4822	116	52424
R	168	RESISTOR STD. FILM	3K6	5%	4822	116	52419
R	170	RESISTOR STD. FILM	430E	5%	4822	116	52423
R	171	RESISTOR STD. FILM	430E	5%	4822	116	52423
R	173	RESISTOR METAL FILM	2K61	1%	5322	116	50671
R	175	RESISTOR STD. FILM	1K2	5%	4822	110	73109
R	176	RESISTOR STD. FILM			4822	110	73109
			1K2	5%			
R	177	RESISTOR STD. FILM,	10E	5%	4822	116	52332
R	178	RESISTOR STD. FILM	10E	5%	4822	116	52332
R	180	RESISTOR METAL FILM	453E	1%	5322	116	54523
R	181	RESISTOR METAL FILM	453E	1%	5322	116	54523
R	182	RESISTOR METAL FILM	643F.	1%	5322	116	54531
R	183	POTMETER TRIMMER	470E		5322	101	14047
R	185	RESISTOR METAL FILM	33K2	1%	4822	116	512 <i>59</i>
R	186	RESISTOR METAL FILM	15E	1%	4822	116	51221
R	187	RESISTOR METAL FILM	90E9	1%	5322	116	54466
R	188	RESISTOR METAL FILM	8K25	1%	5322	116	51498
R	190	RESISTOR METAL FILM	110K	1%	5322	116	54701
R	191	RESISTOR STD. FILM	51E	5%	4822	116	52369
R	192	RESISTOR STD. FILM	51E	5%	4822	116	52369
R	198	RESISTOR STD. FILM	1E0	5%	4822	116	52385
R	200	RESISTOR METAL FILM	1K40	1%	5322	116	5556 9
R	201	RESISTOR METAL FILM	61K9	1%	4822	116	51265
R	202	RESISTOR METAL FILM	61K9	1%	4822	116	51265
R	203	RESISTOR STD. FILM	1K	5%	4822	116	52391
R	205	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	206	RESISTOR METAL FILM	5K62		4822	116	51281
R	207	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	208	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	210	RESISTANCE METAL FILM	121E	1%	5322	116	54426
R	211	RESISTOR METAL FILM	121E	1%	5322	116	54426
R	212	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	213				4822	116	51281
		RESISTOR METAL FILM	5K62	1%			
R	215	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	216	RESISTOR METAL FILM	5K62	1%	4822	116	51281
R	218	RESISTOR STD. FILM	110K	5%	4822	116	52455
R	220	RESISTOR STD. FILM	120E	5%	4822	110	73083
R	221	RESISTOR STD. FILM	120E	5%	4822	110	73083
R	222	RESISTOR STD. FILM	200E	5%	4822	116	52405
R	223	RESISTOR STD. FILM	3K6 .	5%	4822	116	52419
R	225	RESISTOR STD. FILM	1K .	5%	4822	116	52391
R	226	RESISTOR STD. FILM	3K6	5%	4822	116	52419

PC	OSNR	DESCRIPTION		 ···	ORI	DERI	NG CODE
		Λ					
R	228	RESISTOR STD. FILM	3K6	5%	4822	116	52419
R	230	RESISTOR STD. FILM	1K	5%	4822	116	52391
R.	231	RESISTOR STD. FILM	3K6	. 5%	4822	116	52419
R	232	RESISTOR STD. FILM	1K	5%	4822	116	52391
R	233	RESISTOR METAL FILM	7K15	1%	5322	116	54606
R	235	RESISTOR METAL FILM	7K15	1%	5322	116	54606
R	236	RESISTOR METAL FILM	2K43	1%	5322	116	54004
R	237	RESISTOR METAL FILM	2K43	1%	5322	116	54004
R	301 300	RESISTOR STD. FILM	10K	5%	4822	116	52452
R	302	RESISTOR STD. FILM	10K	5%	4822	116	52452
R	303	RESISTOR METAL FILM	634E	1%	5322	116	54531
R	305	RESISTOR METAL FILM	634E	1%	5322	116	54531
R	306	RESISTOR METAL FILM	909E	1%	5322	116	55278
R	307	RESISTOR METAL FILM	4K64	1%	5322	116	50484
R	308	RESISTOR METAL FILM	7K32	1%	5322	116	55372
R	310	RESISTOR STD. FILM	10K	5%	4822	116	52452
R	311	RESISTOR STD. FILM	10K	5%	4822	116	52452
R	312	RESISTOR METAL FILM	634E	1%	5322	116	54531
R	313	RESISTOR METAL FILM	634E	1%	5322	116	54531
R	315	RESISTOR METAL FILM	909E	1%	5322	116	55278
R	316	RESISTOR METAL FILM	4K64	1%	5211	116	50484
R	317	RESISTOR METAL FILM	7K32	1%	5322	116	55372
R	318	RESISTOR METAL FILM	825E	1%	5322	116	5 4541
R	319	RESISTOR METAL FILM	825E	1%	5322	116	54541
R	320	RESISTOR METAL FILM	1K05	1%	4822	116	52898
R	321	RESISTOR METAL FILM	909E	1%	5322	116	55278
R	322	RESISTOR STD. FILM	2K	5%	4822	116	52406
R	323′	RESISTOR STD. FILM	1K	5%	4822	116	52391
R	324	RESISTOR METAL FILM	825E	1%	5322	116	54541
R R	325	RESISTOR METAL FILM	3K65	1%	5322	116	54587
R	326	RESISTOR METAL FILM	36E5	1% .	5322	116	50409
R	327	RESISTOR METAL FILM	36E5	1%	5322	116	50409
R	328	RESISTOR METAL FILM	100E	1%	5322	116	50581
R	329	RESISTOR STD. FILM	100E	5%	4822	116	52389
R	330	RESISTOR STD. FILM	5K1	5%	4822	110	70126
R	331	RESISTOR STD. FILM	5K1	5%	4822	110	70126
R	332	RESISTOR METAL FILM	2K15	1%	5322	116	50767
R	333	RESISTOR METAL FILM	2K49	1%	5322	116	50581
R	335	RESISTOR METAL FILM	2K49	1%	5322	116	50581
R	336	RESISTOR METAL FILM	1K54	1%	5322	116	50586
R	337	RESISTOR METAL FILM	1K3	1%	4822	116	51238
R	338	RESISTOR STD. FILM	6K8	5%	4822	110	73129
R	340	RESISTOR STD. FILM	4K7	5%	4822	116	52426
R	342	RESISTOR METAL FILM	845K	1%	5322	116	52107
R.	343	RESISTOR METAL FILM	402K	1%	5322	116	55283
R	344	RESISTOR METAL FILM	301K	1%	5322	116	55743
R	345	RESISTOR METAL FILM	5K6	5%	4822	110	73127
R	346	RESISTOR METAL FILM	26E1	1%	5322	116	50876
R	347	RESISTOR STD. FILM	5K6	5%	4822		73127
R	348	RESISTOR STD. FILM	30K	5%	4822	116	52466

5K6 30K

RESISTOR STD. FILM RESISTOR STD. FILM

348

5%

4822 116 52466

F	POSNR	DESCRIPTION			ORDER	ING CODE
R	350	RESISTOR METAL FILM	523E	1%	5322 116	5 54526
R	351	RESISTOR METAL FILM	523E	1%	5322 116	5 54526
R	352	RESISTOR METAL FILM	2K32	1%	5322 116	54575
R	353	RESISTOR METAL FILM	1K	1%	4822 116	5 51235
R	355	RESISTOR STD. FILM	10K	5%	4822 116	52452
R	356	RESISTOR METAL FILM	1K	1%	4822 116	5 51235
R	357	RESISTOR METAL FILM	3K83	1%	5322 11 <i>6</i>	5 54589
Ŕ		RESISTOR METAL FILM	2K87	1%	5322 116	5 55279
R	360	RESISTOR METAL FILM	4K02	1%	5322 116	5 55448
R	361	RESISTOR STD. FILM	. 1M	5%	4822 116	52493
R	362	RESISTOR STD. FILM	10K	5%	4822 116	52452
R	363	RESISTOR METAL FILM	3K16	1%	5322 116	50579
R	365	RESISTOR METAL FILM	10K	1%	4822 116	51253
R	366	RESISTOR STD. FILM	1M	5%	4822 116	52493
R	367	RESISTOR METAL FILM	196K	1%	5322 116	55364
R	368	RESISTOR METAL FILM	10K	1%	4822 116	51253
R	370	RESISTOR STD. FILM	2K	5%	4822 116	52406
R	371	RESISTOR STD. FILM	5K1	5%	4822 110	70126
R	272	RESISTOR STD. FILM	1K	5%	4822 116	52391
R	373	RESISTOR STD. FILM	10K	5%	4822 116	52452
R	375	RESISTOR STD. FILM	1E0	5%	4822 116	52385
R	376	RESISTOR STD. FILM	1EO	5%	4822 11 <i>6</i>	52385
- R	377	RESISTOR STD. FILM	1E0	5%	4822 11 <i>6</i>	52385
R	378	RESISTOR STD. FILM	1E0	5%	4822 11 <i>6</i>	52385
R	379	RESISTOR STD. FILM	200E	5%	4822 116	52405
R	380	RESISTOR STD. FILM	2K	5%	4822 116	52406
R	400	RESISTOR METAL FILM	1K	1%	4822 116	51235
R	401	RESISTOR METAL FILM	10K	1%	4822 116	51253
R	402	RESISTOR STD. FILM	5K6	5%	4822 116	73127
R	403	RESISTOR METAL FILM	11K	1%	5322 116	54623
R	404	RESISTOR STD. FILM	10K	5%	4822 116	52452
R	405	RESISTOR STD. FILM	51E	5%	4822` 116	52369
R	406	RESISTOR STD. FILM	51K	5%	4822 116	52473
R	408	RESISTOR METAL FILM	16K9	1%	5322 116	54635
R	410	RESISTOR METAL FILM	3K16	1%	5322 116	50579
R	411	POTMETER TRIMMER	22K		5322 101	14069
R	413	RESISTOR METAL FILM	16K9	1%	5322 116	54635
R	415	RESISTOR METAL FILM	3K16	1%	5322 116	50579
R	416	POTMETER TRIMMER	22K		5322 101	14069
R	417	RESISTOR METAL FILM	1K15	1%	5322 116	54606
R	418	RESISTOR STD. FILM	51E	5%	4822 116	52369
Ŕ	420	RESISTOR METAL FILM	30K1	1%	5322 116	54655
R	421	RESISTOR METAL FILM	316E	1%	5322 116	54511
R	422	RESISTOR METAL FILM	2K05	1%	5322 11 <i>6</i>	50664
R	423	RESISTOR METAL FILM	51K1	1%	5322 116	50672
R	425	RESISTOR METAL FILM	768K	1%	5322 116	52106
R	426	RESISTOR METAL FILM	383K	1%	5322 116	55335
R	427	RESISTOR METAL FILM	191K	1%	5322 116	55363
R	428	RESISTOR METAL FILM	76K8	1%	5322 116	
R	430	RESISTOR METAL FILM	38K3	1%	5322 1:16	55369

POSNR		DESCRIPTION		ORDERING CODE				
R	431	RESISTOR METAL FILM	18K7	1%	5322	116	55362	
R	432	RESISTOR METAL FILM	7K32	1%		116	55372	
R	433	RESISTOR METAL FILM	3K48	1%		116	55367	
R	435	RESISTOR METAL FILM	1K58	1%		116	50622	
R	436	RESISTOR METAL FILM	274E	1%		116	54504	
							70126	
R	437	RESISTOR STD. FILM RESISTOR METAL FILM	5K1 100K	5% 1%		110 116	51268	
R R	438 440	RESISTOR METAL FILM	20K5	1%	,	116	55419	
R	441	RESISTOR METAL FILM	51K1	1%		116	50672	
R	442	RESISTOR METAL FILM	5K11	1%		116	54595	
R	443	RESISTOR METAL FILM	20K5	1%		116	55419	
R	445	RESISTOR METAL FILM	1K27	1%		116	50555	
R	447	RESISTOR METAL FILM	7K5	1%		116	54608	
R	450	RESISTOR METAL FILM	6K65	1%		116	54604	
R`	451	RESISTOR METAL FILM	4K99	1%	5322	116	50523	
R	453	RESISTOR METAL FILM	1K78	1%		116	50515	
·R	455	RESISTOR METAL FILM	1K78	1%		116	50515	
R	456	RESISTOR METAL FILM	806E	1%		116	54539	
R	457	RESISTOR STD. FILM	10K	5%		116	52452	
R	458	RESISTOR - STD. FILM	2K	5%	4822	116	52406	
R	460	RESISTOR STD. FILM	750E	5%	4822	116	52432	
R	461	RESISTOR METAL FILM	2K87	1%	5322	116	55279	
R	462	RESISTOR METAL FILM	2K	1%	5322	116	54572.	
R	463	RESISTOR METAL FILM	4K64	1%		116	50484	
R	465	RESISTOR METAL FILM	3K16	1%	5322	116	50579	
R	466	RESISTOR METAL FILM	3K16	1%		116	50579	
R	467	BESISTOR METAL FILM	3K16	1%		116	50579	
R	468	RESISTOR METAL FILM	82K5	1%		116	55374	
R	470	RESISTOR METAL FILM	82K5	1%		116	55374	
R	471	RESISTOR METAL FILM	348E	1%	5322	116	54515	
R	472	RESISTOR METAL FILM	348E	1%	5322	116	54515	
R	473	RESISTOR METAL FILM	82K5	1%		116	55374	
R	475	RESISTOR METAL FILM	82K5	1%		116	55374	
R	476	RESISTOR STD. FILM	16K	- 5%		116	52459	
R	477	RESISTOR STD. FILM	100K	.5%	4822	116	52543	
R	478	RESISTOR STD. FILM	510E	5%		110	70099	
R	480	RESISTOR STD. FILM	510E	5% .	4822	110	70099	
R	488	RESISTOR STD. FILM	1E0	5%		116	52385	
R	500	RESISTOR STD. FILM	10K	5%		116	52452	
R	501	POTMETER TRIMMER	220K		4822	100	10088	
R	502	RES. HI-TENSION	1M2	5%	4822	110	72189	
R	503	RES. HI-TENSION	8M2	5%	4822	110	42212	
R	504	RES. HI-TENSION	2M7	5%	4822	110	72198	
R	505	RESISTOR STD. FILM	82K	5%	4822	116	52478	
R	506	RESISTOR STD. FILM	120K	5%	4822	116	52496	
R	508	RESISTOR STD. FILM	27K	5%	4822	116	52465	
R	509	RESISTOR METAL FILM	3M	5%	5322	116	518 3 6	
R	510	RESISTOR HI-TENSION	2M7	5%	4822	110	721 9 8	
R	511	RESISTÖR HI-TENSION	2M7	5%		110	72198	
R	512	RESISTOR HI-TENSION	22M	5%	4822	110	42223	

PC	SNR	DESCRIPTION			ORDERING CODE
R	513	RESISTOR STD. FILM	10K	5%	4822 116 52452
R	515	RESISTOR STD. FILM	75E	5%	4822 116 52377
R	516	RESISTOR STD. FILM	110	≺ 5%	4822 116 52455
R	517	RESISTOR STD. FILM	1K	5%	4822 116 52391
R	519	RESISTOR STD. FILM	5K1	5%	4822 110 70126
R	520	RESISTOR STD. FILM	110	K 5%	4822 116 52455
R	521	RESISTOR STD. FILM	1K5	5%	4822 116 52399
R	522	RESISTOR METAL FILM	1K7	8 1%	5322 116 50515
R	523	RESISTOR METAL FILM	4K6	4 1%	5322 116 50484
R	525	RESISTOR METAL FILM	23K	7 1%	5322 1 54646
R	526	RESISTOR METAL FILM	20K	5 1%	5322 116 55419
R	527	RESISTOR METAL FILM	274	E · 1%	5322 116 54504
R	528	RESISTOR METAL FILM	511	E 1%	4822 116 51282
R	530	RESISTOR METAL FILM	1K5	0 1%	4822 116 51239
R	531	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	532	RESISTOR STD. FILM	. 22E	5%	4822 116 52349
R	533	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	535	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	536	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	537	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	538	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	540	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	541	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	542	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	543	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	545	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	546	RESISTOR STD. FILM	22E	5%	4822 116 52349
R	547	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	548	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	552	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	553	RESISTOR STD: FILM	1E	5%	4822 116 52385
R	554	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	555	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	556	RESISTOR STD. FILM	1E	5%	4822 116 52385
R	341	RESISTOR STD. FILM	360	E 5%	4822 116 52418

P	OSNR	DESCRIPTION		ORDERI	NG CODE
		SEMI CONDUCTORS			
٧	20	DIODE	BAV45	5322 130	34037
٧	21	TRANSISTOR, FET	BFS21A	5322 130	40709
٧	22	TRANSISTOR	BF199	4822 130	44154
٧	23	TRANSISTOR	BF199	4822 130	44154
V	25 '	TRANSISTOR	BC558B	4822 130	44197
¥	26	TRANSISTOR			
٧	27	DIODE, REFERENCE ZENER	07775071/	4822 130	44197
٧	28	DIODE, REFERENCE ZENER	BZX75C3V6	4822 130	30765
٧	30	TRANSISTOR	BZX75C3V6	4822 130	30765
٧	31		BF199	4822 130	44154
		TRANSISTOR	BF199	4822 130	44154
٧	32	TRANSISTOR	BF199	4822 130	44154
٧	33	DIODE	1N4148	4822 130	30621
V	35	DIODE	1N4148	4822 130	30621
V	36	DIODE	1N4148	4822 130	30621
٧	37	DIODE	1N4148	4822 130	30621
٧	38.	DIODE	1N4148	4822 130	30621
٧	40	DIODE	1N4148	4822 130	30621
٧	· 41	DIODE	1N4148	4822 130	30621
٧	42	DIODE	1N4148	4822 130	30621
٧	43	DIODE	1N4148	4822 130	30621
٧	45	DIODE	1N4148	4822 130	30621
v	46	DIODE	1N4148	4822 130	30621
v	47	DIODE	1N4148	4822 130	30621
v	48	DIODE	1N4148	4822 130	30621
v	50	DIODE	1N4148	4822 130	30621
٧	51	DIODE	1N4148	4822 130	30621
v	52	DIODE	1N4148	4822 130	30621
v	53	DIODE	1N4148		30621
V	55	DIODE	1N4148	4822 130	30621
v	56	DIODE	1N4148	4822 130	30621
٧	57	DIODE	1N4148	4822 130	30621
٧	58	DIODE	1N4148	4822 130	30621
V	60	DIODE	1N4148	4822 130	30621
٧	120	DIODE	BAV45	5322 - 130	34037
٧	121	TRANSISTOR, FET	BFS21A	5322 130	40709
٧	122	TRANSISTOR	BF199 .	4822 130	44154
٧	123	TRANSISTOR	BF 199	4822 130	44154
٧	126	TRANSISTOR	BC558B	4822 130	44197
٧	127	DIODE, REFERENCE ZENER	BZX75C3V6	4822 130	30765
٧	128	DIODE, REFERENCE ZENER	BZX75C3V6	4822 130	30765
٧	130	TRANSISTOR	BF199	4822 130	44154
٧	131	TRANSISTOR	BF199	4822 130	44154
٧	132	TRANSISTOR	BF199	4822 130	44154
٧	133	DIODE	1N4148	4822 130	30621
٧	135	DIODE	1N4148	4822 130	30621
٧	136	DIODE		4822 130	30621
	. 125	TRANSISTOR	1N4148	4822 130	44197
V	138	DIODE	BC558B	4822 130	
V V	140	DIODE	1N4148 1N4148	4822 130	30621 30621
. V	140	DIODE	1N4148 1N4148	4822 130	30621
٧	144	DIODE	1144 140	4022 170	70041

pα	OSNR	DESCRIPTION				ORD	ERIN	NG CODE
					· · · · · · · · · · · · · · · · · · ·			
								4
٧	143	DIODE	1	N4148		4822	130	30621
v	145	DIODE	1	N4148		4822	130	30621
v	146	DIODE	1	N4148		4822	130	30621
V -	147	DIODE	* 1	N4148		4822	130	30621
V	148	DIODE		N4148	•	4822	130	30621
				N4148		4822	130	30621
V	150	DIODE				4822	130	30621
, V	151	DIODE		N4148		4822	130	30621
٧	152	DIODE		N4148		4822	130	30621
٧	153	DIODE		N4148		4822	130	30621
٧	155	DIODE	x - *	N4148		4022	170	70021
V	156	DIODE	1	N4148		4822	130	30621
٧-	157	DIODE	1	N4148		4822	130	30621
٧	158	DIODE	1	N4148		4822	130	30621
V	160	DIODE	1	N4148		4822	130	30621
V	200	TRANSISTOR	E	3F 199		4822	130	44154
				05400		4822	130	44154
٧	201	TRANSISTOR		3F199		4822	130	44154
٧	202	TRANSISTOR		3F199		4822	130	44154
٧	203	TRANSISTOR		3F199		4822	130	34047
С	205	DIODE, REFERENCE ZENER		3ZX75C1V4				
٧	206	DIODE, REFERENCE ZENER	E	3ZX75C1V4		4822	130	34047
V	207	TRANSISTOR		3F199		4822	130	44154
·v	208	TRANSISTOR		3F199		4822	130	44154
v	210	TRANSISTOR	E	3F 199		4822	130	44154
v	211	TRANSISTOR	Ε	3F199		4822	130	44154
v	212	DIODE, REFERENCE ZENER		3ZX79C56		4822	130	34258
			_			4000	170	43258
· V	213	DIODE, REFERENCE ZENER		3ZX79C56		4822	130	
٧	215	DIODE		N4148		4822	130	30621
٧	216	DIODE		N4148		4822	130	30621
٧	217	DIODE		N4148		4822	130	30621
٧	300	TRANSISTOR	E	3F450		4822	130	44237
٧	301	TRANSISTOR	· E	3F450		4822	130	44237
v	302	TRANSISTOR	. Е	BC558B		4822	130	44197
v	303	TRANSISTOR		3F450		4822	130	44237
v	305	TRANSISTOR		3F450		4822	130	44237
v	306	TRANSISTOR		BC558B		4822	130	44197
						4000	170	70/21
٧	307	DIODE		IN4148		4822	130	30621
V	308	DIODE		IN4148		4822	130	30621
٧	310	DIODE		IN4148		4822	130	30621
V	311	DIODE		IN4148		4822	130	30621
٧	312	DIODE	•	1N4148		4822	130	30621
V	313	DIODE, REFERENCE ZENER		3ZX79C5V6		4822	130	34173
v	315	TRANSISTOR		3F199		4822	130	44154
v	316	DIODE, REFERENCE ZENER		BZX79C6V2		4822	130	34167
V	317	DIODE		1N4148		4822	130	30621
v	318	TRANSISTOR		BC548C		4822	130	44196
V	320	TRANSISTOR		BF199		4822	130	44154
٧	321	TRANSISTOR		BF199		4822	130	44154
٧	322	TRANSISTOR	•	BF 199		4822	130	44154
٧	323	TRANSISTOR		BC558B		4822	130	44197
٧	325	DIODE		1N4148		4822	130	30621

P	OSNR	DESCRIPTION		ORDI	ERIN	NG COD
٧	326	TRANSISTOR	BC558B	4822	130	44197
٧	327	TRANSISTOR	BC558B	4822	130	44197
٧	328	TRANSISTOR	BC558B	4822	130	44197
٧	330	TRANSISTOR	BC558B	4822	130	44197
٧	331	TRANSISTOR	BC548C	4822	130	44196
٧	332	TRANSISTOR	BC548C	4822	130	44196
٧	333	DIODE	1N4148	4822	130	30621
٧	335	DIODE	1N4148	4822	130	30621
V	336	DIODE	1N4148		130	30621
٧	337	DIODE	1N4148		130	30621
V	338	TRANSISTOR	BC558B	4822	130	44197
٧	340	TRANSISTOR	BC548C	4822	130	44196
٧	341	DIODE	1N4148	4822	130	30621
٧	342	DIODE	1N4148	4822	130	30621
V	343	TRANSISTOR	BC558B	4822	130	44197
٧	400	DIODE	1N4148	4422	130	30621
٧	401	TRANSISTOR	BSX20	4822	130	41705
٧	402	TRANSISTOR	BC548C	4822	130	44196
٧	403	TRANSISTOR	BC548C	4822	130	44196
٧	405	TRANSISTOR	BC558B	4822	130	44197
٧	406	DIODE	1N4148	4822	130	30621
٧	407	DIODE	1N4148	4822	130	30621
٧	408	TRANSISTOR	BC548C	4822	130	44196
٧	410	TRANSISTOR	BC548C	4822	130	44196
٧	411	TRANSISTOR	BC548C	4822	130	44196
٧	412	DIODE	1N4148		130	30621
٧	413	DIODE	1N4148		130	30621
٧	415	DIODE	1N4148		130	30621
٧	416	DIODE	1N4148		130	30621
٧	417	TRANSISTOR	BC558B	4822	130	44197
٧	418	TRANSISTOR	BC548C	4822	130	44196
٧	419	DIODE	1N4148	4822	130	30621
٧	421	TRANSISTOR	BF423	4822	130	41646
٧	422	TRANSISTOR	BC558B	4822	130	44197
٧	423	TRANSISTOR	BF423	4822	130	41646
٧	425	DIODE, REFERENCE ZENER	BZX79C6V2	4822	130	34167
٧	426	TRANSISTOR	BC548C	4822	130	44196
٧	427	TRANSISTOR	BC548C	4822	130	44196
٧	428	TRANSISTOR	BF422	4822	130	41782
٧	430	TRANSISTOR	BF422	4822	130	41782
٧	431	DIODE, REFERENCE ZENER	BZX75C2V1	4822	130	34049
٧	432	DIODE, REFERENCE ZENER	BZX75C2V1	4822	130	34049
٧	433	TRANSISTOR	BF422	4822	130	41782
٧	435	TRANSISTOR	BF422	4822	130	41782
٧	500	DIODE	BY548	5322	130	32274
٧	501	DIODE	BY548	5322	130	32274
٧	502	DIODE	BY548	5322	130	32274
٧	503	DIODE	BY548	5322	130	32274
٧	504	TRANSISTOR	BF422	4822	130	41782
٧	505	TRANSISTOR	BF422	4822	130	41782

POSNR DESCRIPTION

ORDERING CODE

٧	506	TRANSISTOR	BF422	4822	130	41782
٧	507	TRANSISTOR	BC548C	4822	130	44196
٧.	508	DIODE	1N4148	4822 4822	130 130	30621 41485
٧	510	DIQDE	BY509	4022	170	41407
٧	511	DIODE	BAV21			
٧	512	BRIDGE RECT.	SKB2-08/L5A	4822	130	32031
٧	513	TRANSISTOR	BC338/16	4822	130	44121
٧	515	DIODE, REFERENCE ZENER	BZX79B5V6	4822	130	34173
٧	516	BRIDGE RECT.	SKB2-08/L5A	5322	130	32031
٧	517	TRANSISTOR	BF422	4822	130	41782
V	518	DIODE, REFERENCE ZENER	BZX75C1V4	4822	130	34047
v	520	TRANSISTOR	BF422	4822	130	41782
V	521	TRANSISTOR	BC548C	4822	130	44196
٧	522	TRANSISTOR	BC558B	4822	130	44197

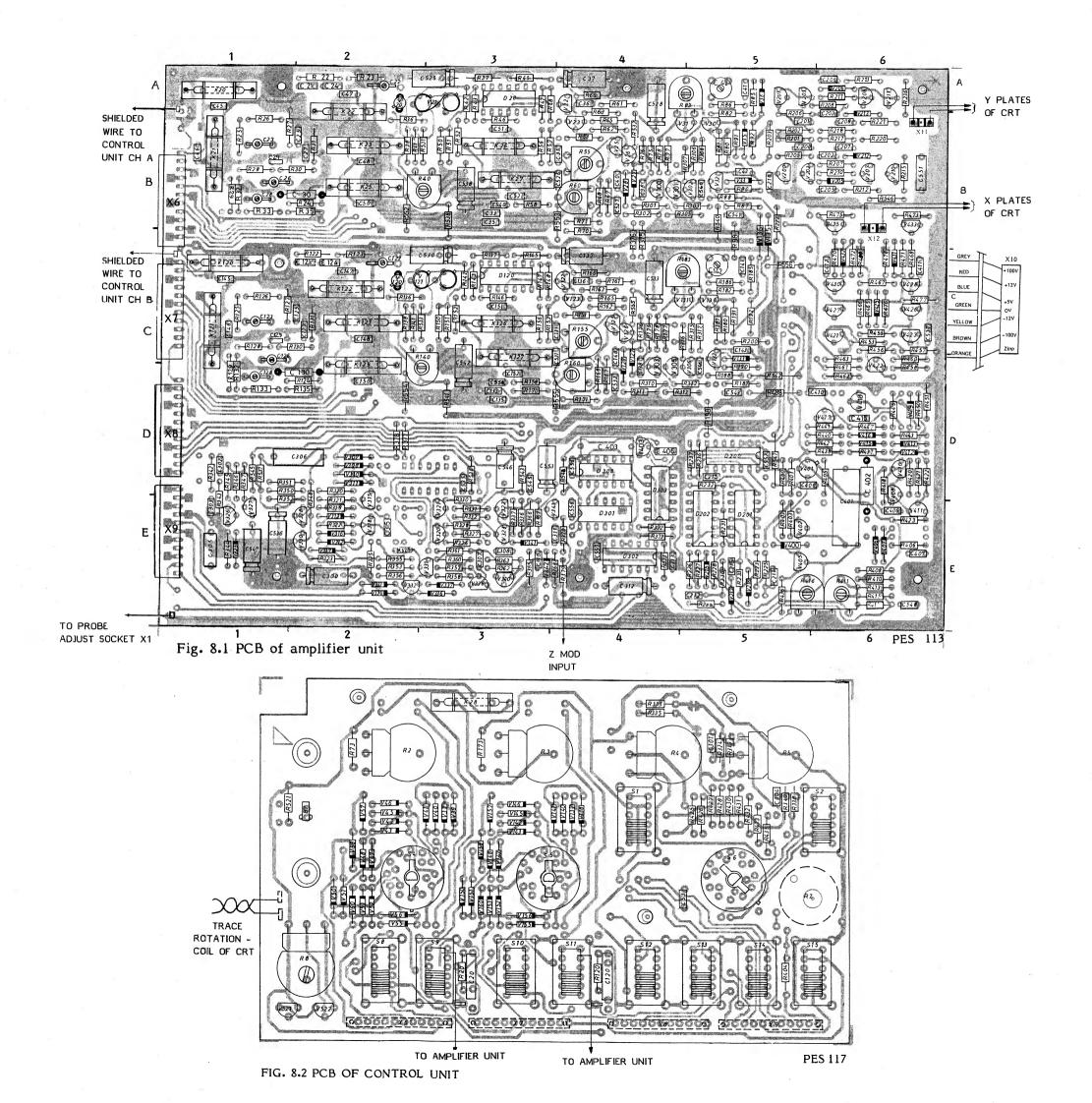
INTEGRATED CIRCUITS

	20	INTEGR. CIRCUIT	SG3823	532	2 209	84862
D	20			532	2 209	84862
D	120	INTEGR. CIRCUIT	SG3823			
D	200	INTEGR. CIRCUIT	74LS74	482	2 209	80782
D	201	INTEGR. CIRCUIT	74LS132N	532	2 209	85201
D	202	INTEGR. CIRCUIT	74LS02N	532	2 209	85312
D	300	INTEGR. CIRCUIT	74LS132N	532	2 209	85201
D	301	INTEGR. CIRCUIT	74LS86N	532	2. 209	8.4997
D	302	INTEGR. CIRCUIT	74122	532	2 209	84231
D	303	INTEGR. CIRCUIT	74LS74	482	2 209	80782
D	500	INTEGR. CIRCUIT	7812	482	2 209	81016
n	501	INTEGR. CIRCUIT	7912T	532	2 209	81856

MISCELLANEOUS

PO	SNR	DESCRIPTION	ORDERII	NG CODE		
		BNC CONNECTOR	5322 26	7 10004		
		SOLDER TAG	5322 29	0 30204		
		HEX NUT (M4)	5322 50	6 14005		
		CONTROL UNIT	5322 21	6 51112		
			5322 210			
		TEXT STRIP	5322 45	5 81015		
		SCREW 3, 9x13-BLACK	5322 502	2 14159		
		MAINS CORD (EUR-TYPE)	5322 32	1 14071		
		MOLEX CONNECTOR	5322 265	5 40506		
		MOLEX CONNECTOR 11 PINS FEM	5322 267	7 50638		
		MOLEX CONNECTOR 7 PINS MA	5322 265	5 40505		
		MOLEX CONNECTOR 7 PINS FEM	5322 267	50439		
		REED RELAY CONTACT	5322 280	24126		
		CATHODE RAY TUBE		20106		
		POWER ON LED	5322 131	•		
F	500	THERMAL FUSE	4822 130 5322 252			
F	501	MAINS FUSE 125 mA SLOW BLOW	4822 253		*	
F	501	MAINS FUSE	"4000 DE0	20012		
K	20	250 mA SLOW BLOW REED RELAY ASSEMBLY	4822 253 5322 280			
K	21	REED RELAY ASSY REED RELAY ASSY	5322 280	20126		
K K	22 23	REED RELAY ASSY	5322 280 5322 280			
K	25	REED RELAY ASSY	5322 280			
K K	26 27	REED RELAY ASSY REED RELAY ASSY	5322 280 5322 280			
, K	28	REED RELAY ASSY	5327 280			
K	120	REED RELAY ASSY	5322 280			
K K	121 122	REED RELAY ASSY REED RELAY ASSY	5322 280 5322 280	20126 20126		
K	123	FEED RELAY ASSY	5322 280			
K	125	REED RELAY ASSY	5322 280	20126		
K K	126 127	REED RELAY ASSY REED RELAY ASSY	5322 280 5322 280			
		• •	J322 28U	20126		
S S	1 2	SLIDE SWITCH SLIDE SWITCH	5322 277 5322 277			
S S S	4 5	ROTARY AMPL/DIV SWITCH ROTARY AMPL/DIV SWITCH				
S	6	ROTARY TIME/DIV SWITCH	5322 273			
S	8	SLIDE SWITCH	5322 277	24077		
S	9 10	SLIDE SWITCH	5322 277			
S	11	SLIDE SWITCH SLIDE SWITCH	5322 277 5322 277			
S	12 13	SLIDE SWITCH SLIDE SWITCH	5322 277	24077		
\$ \$ \$ \$ \$ \$ \$	14	SLIDE SWITCH	5322 277 5322 277			
S	15	SLIDE SWITCH	5322 277			
T	500	MAINS TRANSFORMER	5322 146	40385		

8. CIRCUIT DIAGRAMS and PCB LAY-OUTS.



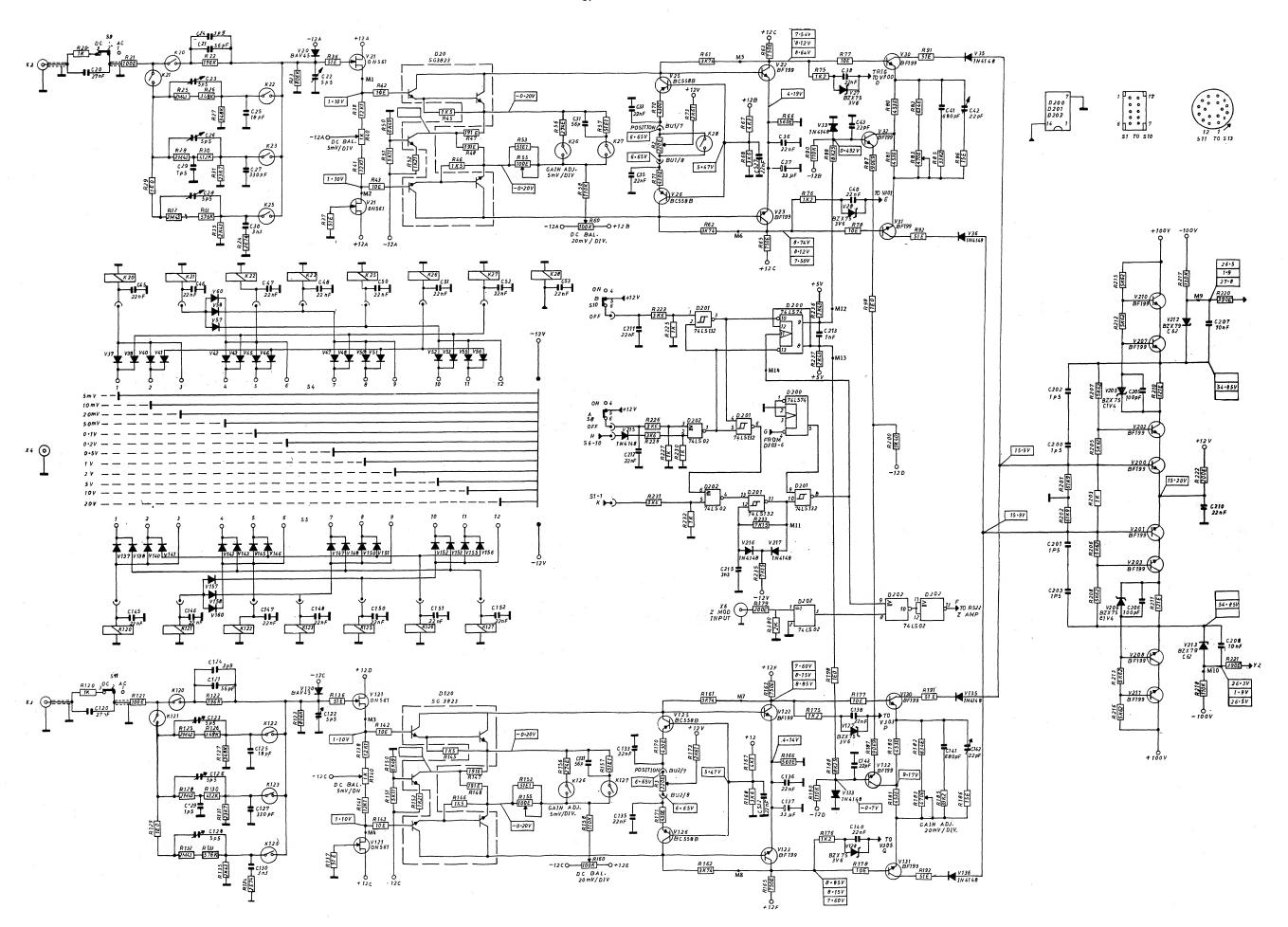
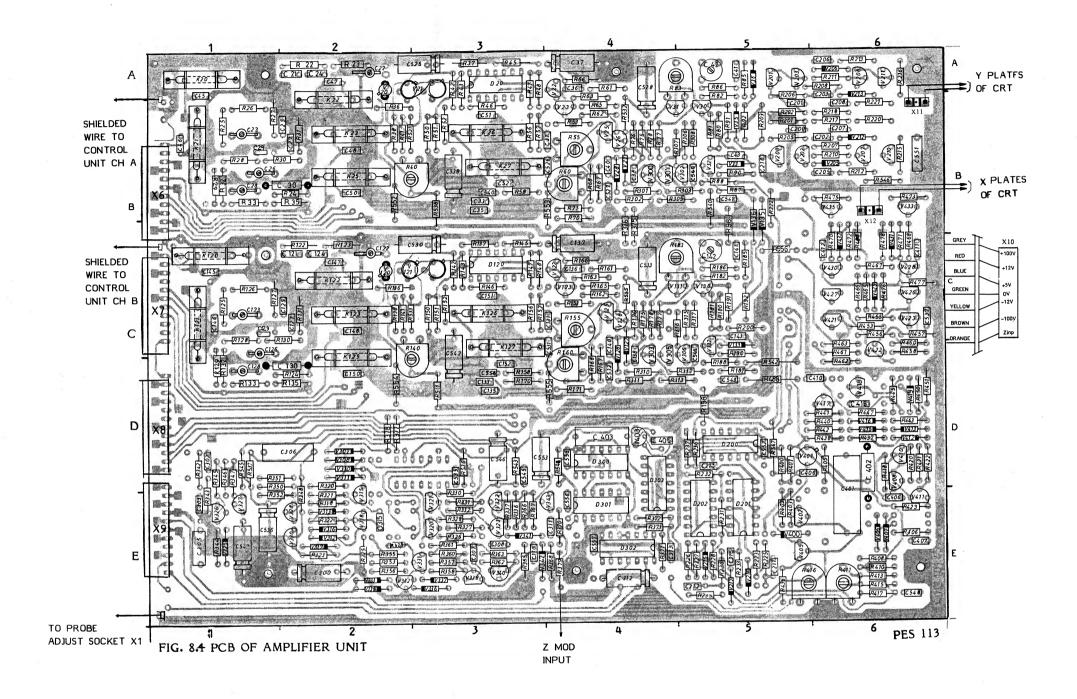


Fig. 8.3 Circuit diagram vertical channel.



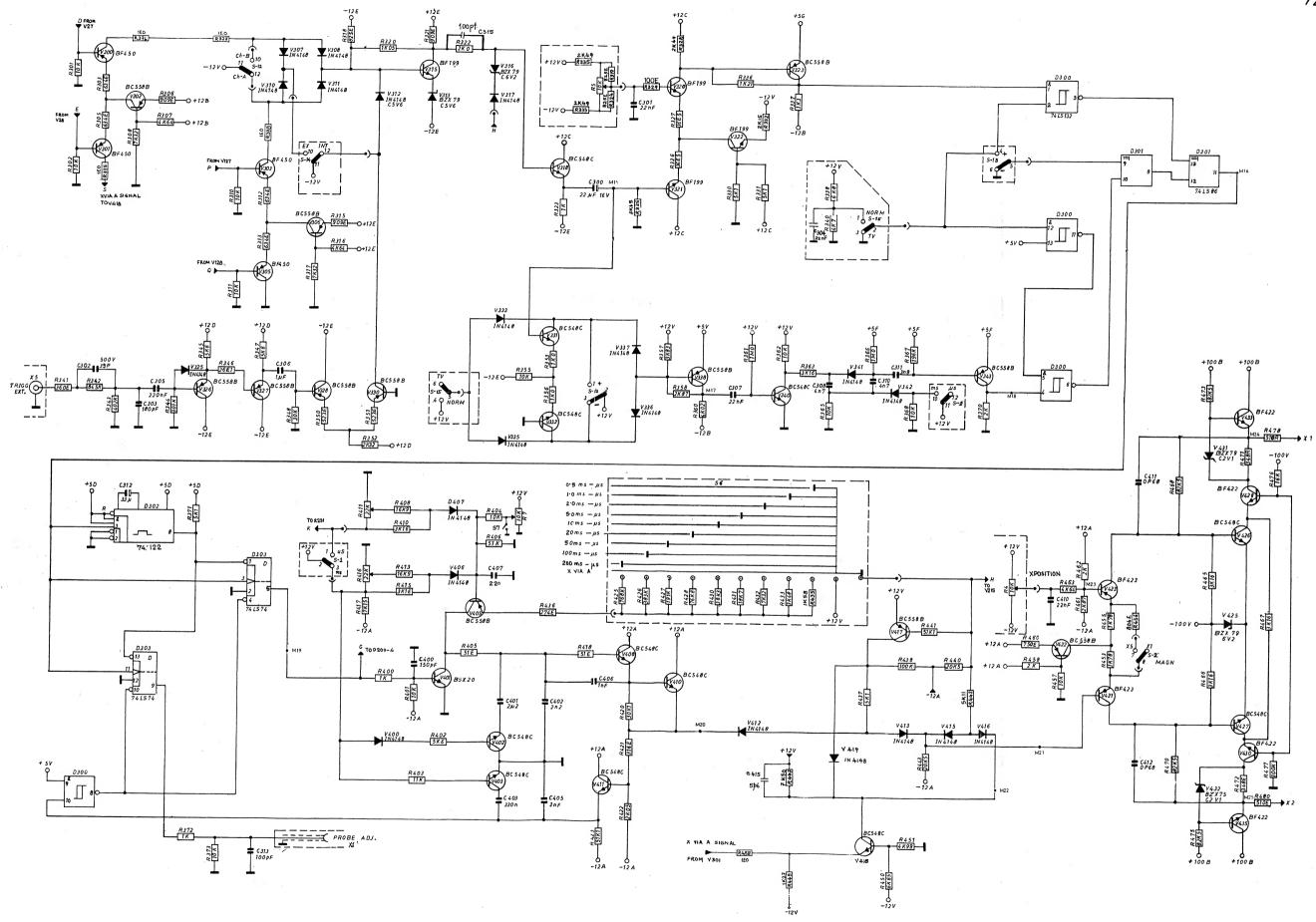
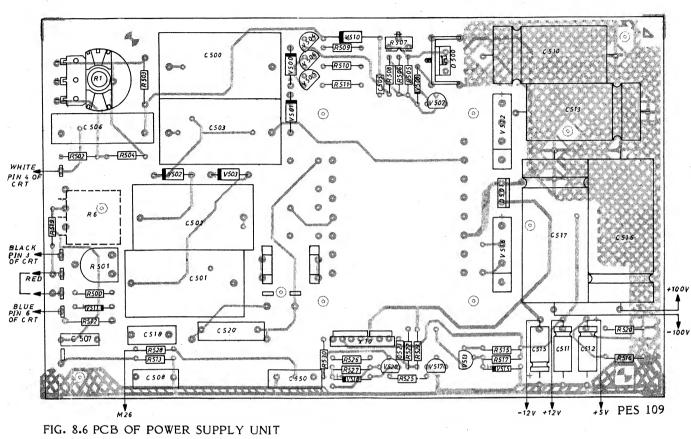


Fig. 8.5 Circuit diagram x-channel and time base.



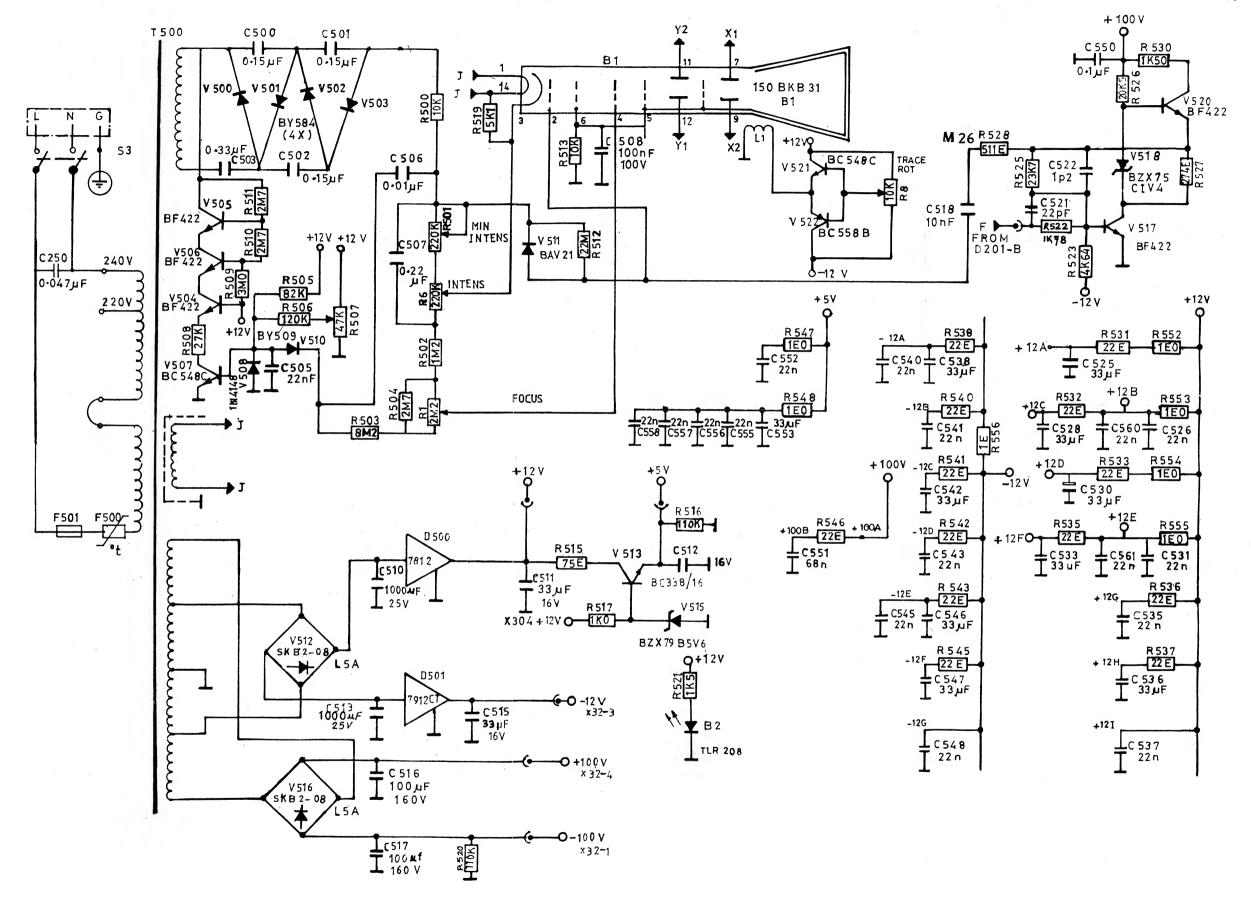


Fig. 8.7 Circuit diagram of power supply unit

9. MODIFICATIONS

- (1) V213, V212 has been changed to BZX79 C62 (4822 130 34384)
- (2) R29, R129 has been changed to 1E (4822 116 52385)
- (3) R220, R221 has been changed to 390E (5322 116 54518)
- (4) Read V21 & V121 as BFS21A in Fig. 8.3

All these changes are incorporated from the very first version of PM3206/02. Hence no change needs to be done in the field.

CODING SYSTEM OF FAILURE REPORTING FOR QUALITY ASSESSMENT OF T & M INSTRUMENTS (excl. potentiometric recorders)

The information contents of the coded failure description is necessary for our computerized processing of quality data.

Since the reporting of repair and maintenance routines must be complete and exact, we give you an example of a correctly filled-out PHILIPS SERVICE Job sheet.

(1)	<u> </u>	•		•	
Country	2) Day Month Year	③ Typenumber	Version	Factory Serial	l no
3 2	1 5 0 4 7 5	O P M 3 2	6002	D 0 0 0 7	8 3 1
ا النات	لندلنداندان		لسلسلسا	1—1—1—1—1—1—1—1—1—1—1—1—1—1—1—1—1—1—1—	<u> </u>
	COL	PED FAILURE DESC	KIPITON	6	
(5)					
Nature of	call Location	Component s	equence no. C	ategory	
Installatio Pre sale re Preventive maintenan Corrective maintenan Other	pair 0 0 2 1 ce	T S O 6 R O O 6 9 9 0 O	0 7 3 1 0 1	- C-3 '	ne (8)
	tion of the information	to be entered in the	various boxes:		
①Country: 3	2 = Switzerland			Frein	
②Day Month Yo	ear 1 5 0 4 7	5 =15 April 1975		· ·	
③Type number/	Version 0 P M 3	2 6 0 0 2		PM 3260, version 0: s this number is place)	
4Factory/Serial	number D 0 0	7 8 3 = DO 7	83 These data a	re mentioned on the	type plate of
5Nature of call:	Enter a cross in the red	elevant box	the instrumer	π	
Location		Component sequ	ience no.	Category	
no or mechanica of this part (refe LISTS' in the ma Example: 0001 f 000A f 0075 f If units are not r	oblem area. of the part alt occurs, e.g. unit al item no er to 'PARTS anual).	graticule 990002 Knob (in etc.) 990003 Probe (c to instru 990004 Leads an 990005 Holder (fuse, bo 990006 Complete board, h 990007 Accesso without	ty component. Inponent sed in the circuit esignation is I letters must be from the left) and boxes and be written (in the last digit nt-most box) in nd boxes. Itified in the In/Not applicable or rack (text nblem, grip, rail, e, etc.) ncl. dial knob, ca canly if attached ment) id associated plu valve, transistor, ard, etc.) e unit (p.w. it. unit, etc.) ry (only those type number)	not press disappea 1 Software 2 Readjust 3 Electrical joint, etc 4 Mechanic filing, rer 5 Replacem resistor, 6 Cleaning 7 Operator 8 Missing 9 Environm not met	error ment repair (wiring, solder) cal repair (polishing, machining, etc.) nent (of transistor, etc.) and/or lubrication
D.		990008 Documer supplements 990008 Foreign 990099 Miscella	ent, etc.) object		

- ①Job completed: Enter a cross when the job has been completed.
- (3) Working time: Enter the total number of working hours spent in connection with the job (excluding travelling, waiting time, etc.), using the last box for tenths of hours.

	1	2	=1.2	working	hours	(1	h	12	min	

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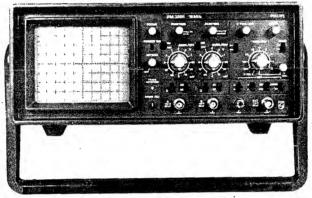
MODIFICATIONS

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15 MHz Dual trace oscilloscope PM3206

Service Manual

4822 872 05316 870601/1/03



SCS 001



PHILIPS

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1. GENERAL INFORMATION

Read these pages carefully before installation and use of the instrument.

The following clauses contain information, cautions and warnings which must be followed to ensure safe operation and to retain the instrument in a safe condition. Adjustment, maintenance and repair of the instrument shall be carried out only by qualified personnel.

1.1. Safety Precautions

For the correct and safe use of this instrument it is essential that both operating and servicing personnel follow generally-accepted safety procedures in addition to the safety precautions specified in this manual.

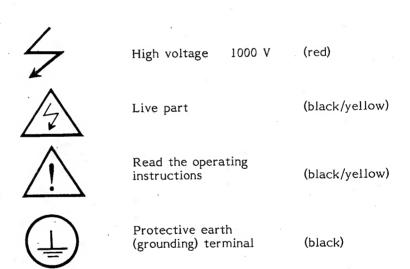
Specific warning and caution statements, where they apply, will be found throughout the manual. Where necessary, the warning and caution statements and/or symbols are marked on the apparatus.

1.2. Caution and Warning Statements

is used to indicate correct operating or maintenance procedures in order to prevent damage to or destruction of the equipment or other property.

warning: calls attention to a potential danger that requires correct procedures or practices in order to prevent personal injury.

1.3. Symbols



1.4. Impaired Safety-Protection

Whenever it is likely that safety-protection has been impaired, the instrument <u>must</u> be made inoperative and be secured against any unintended operation. The matter should then be referred to qualified technicians. Safety protection is likely to be impaired if, for example, the instrument fails to perform the intended measurements or shows visible damage.

1.5. General Clauses

1.5.1. WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts and accessible terminals which can be dangerous to life.

- 1.5.2. The instrument shall be disconnected from all voltage sources before it is opened.
- 1.5.3. Bear in mind that capacitors inside the instrument can hold their charge even if the instrument has been separated from all voltage sources.
- 1.5.4. WARNING: Any interruption of the protective earth conductor inside or outside the instrument, or disconnection of the protective earth terminal, is likely to make the instrument dangerous. Intentional interruption is prohibited.

1.6. CHARACTERISTICS

A. Performance Characteristics

- Properties expressed in numerical values with stated tolerance are guaranteed by PHILIPS. Specified non-tolerance numerical values indicate those that could be nominally expected from the mean of a range of identical instruments.

This specification is valid after the instrument has warmed up for 15 minutes (reference temperature 23°C).

B. Safety Characteristics

This apparatus has been designed and tested in accordance with Safety Class I requirements of IEC Publication 348, and has been supplied in a safe condition.

C. Initial Characteristics

- Overall dimensions :

Height
Width (including handle)
Depth

Maximum weight (Mass) : 5 Kg

Operation position a) horizontally on bottom feet

b) vertically on rear feet

142 mm

378 mm

348 mm

c) one fixed tilted position between a) and b)

1.6.1 C.R.T.

Type : 150 BTB 31 rectangular tube with 2KV accelerating voltage

Screen type : P 31 phosphor standard

Useful screen area : 8 x 10 div of 1 cm

Graticule : Internal graticule with Centimeter division and 2 mm divisions along the central axes.

1.6.2 VERTICAL OR Y-AXIS

Response : DC: 0 Hz ... 15 MHz (-3dB) AC: 10 Hz ... 15 MHz (-3dB)

Risetime : 23 ns approximately

Deflection coefficient : 5 mV ... 20V/div. calibrated steps, 1-2-5 sequence.

Accuracy : $\pm 5\%$ Display modes : A

B A & B in chopped or alternate mode

A ± B

Input impedance : $1 M\Omega // 35 pF$

Input coupling : AC,DC,O: In 'Zero' input signal is disrupted and amplifier is grounded.

Maximum rated input voltage

400 V (dc + ac peak) test voltage 570V (rms)

50 Hz during 1 min according to IEC 348

1.6.3 HORIZONTAL OR X-AXIS

Horizontal display : - Time base

modes - X-Y operation with X deflection via A-input

1.6.4 HORIZONTAL AMPLIFIER

Response : DC : 0 Hz ... 1 MHz (-3dB)

Deflection coefficients : See Y-axes

Phase error : 3° at 10 kHz

1.6.5 TIME BASE

Time coefficients : 0.2s/div. 0.5us/div. in 2 x 9 calibrated steps in 1-2-5 sequence.

Variable sweep rate facility at any time/div setting.

X 5 magnifier extends max. sweep rate to 100 ns/div

Accuracy : ± 5%

Additional error for magnifier: ± 2%

1.6.6 TRIGERRING

Trigger source : Internal : A or B

External

Trigger coupling : Normal (AC Coupled)

T٧

Slope : : + or -

Trigger sensitivity : Internal: 1 div. at 100 kHz

External: 0.75V at 100 kHz

Trigger level range : ± 8 div.

External trigger

input impedance : $1 \text{ M } \Omega // 35 \text{ pF}$

Max. rated input voltage: 400 V (dc + ac peak)

test voltage 570 V(rms)

50 Hz during 1 min according to IEC348

1.6.7 Z-MOD INPUT

Trace blanking : TTL High blanks trace.

OV or not connected no trace blanking

Max. rated input voltage: +25V and -10V.

1.6.8 CALIBRATION

Signal available for probe adjustment. Probe to be adjusted in 0.5 ms/div setting of TIME/DIV and 0.2V/DIV setting of AMPL/DIV switches.

1.6.9 **POWER**

Nominal Line voltage : $120V \pm 10\%$, 50 - 60 Hz (ac) and freq $220V \pm 10\%$, 50 - 60 Hz

220V±10%, 50 - 60 Hz 240V±10%, 50 - 60 Hz

Power consumption : 30W maximum. The insulation between PM 3206 and line

fulfills the safety requirements of IEC 348 for Class 1

instruments.

1.6.10. ENVIRONMENTAL CAPABILITIES

The environmental data are valid only if the instrument is checked in accordance with official checking procedure. Details on these procedures and failures criteria are supplied on request by PHILIPS Organisation in your country or by PHILIPS, Industrial & Electroacoustic Systems Division, Eindhoven, The Netherlands.

Ambient temperature.

Rated range of use: 5°C...+40°C Limits for operating: -10°C...+55°C Storage and transport: -40°C...+70°C

Altitude

: Operating: to 5000m (15000 ft)
Non-operating: to 15000m (45000 ft)

Humidity

: In accordance with IEC 68 Db

Shock

300 m/s²(30g): half sine wave shock of 11ms. duration: 3 shocks per direction for a total of 12 shocks.

Vibration.

30 m/s²(3g) vibrations in three directions with a maximum of 15 min. per direction; 10 mins. with a frequency of 15-25 Hz and a peak-peak altitude of 1 mm. Unit mounted on vibration table without shock absorbing material.

Recovery time

Operates within 60 min. coming from - 10°C soak, going into 60% relative humidity at +20°C room conditions.

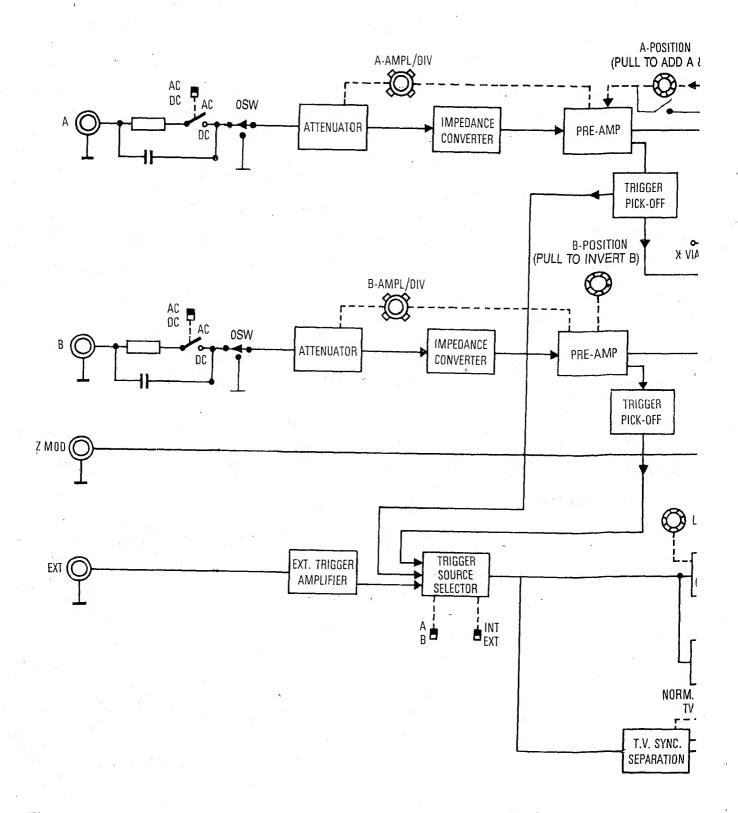


Fig. 2.1. Block diagram

- The channel A or B TRIGGER PICK-OFF amplifiers which are activated by the A and B switches.
- The EXTERNAL TRIGGER AMPLIFIER in order to trigger on a signal applied to the external input X5 activated by the EXT switch.

The TRIGGER SOURCE SELECTOR output is applied to the TRIGGER COMPARATOR, where it is compared with the input from the LEVEL control. This enables the selected level to be compared with the incoming trigger signal to determine the level at which the time-base starts. With switch NORM/TV in TV position the TRIGGER SOURCE SELECTOR output is also fed to the TV SYNC SEPARATION stage. In case of positive video signal the +/- switch should be in + position.

The TV SYNC SEPARATOR stage allows triggering on TV frame pulses (ms position of ms/us switch) or TV line pulses (us position of ms/us switch). The output of the TRIGGER COMPARATOR feeds the SWEEP GATING LOGIC. Depending on the position of the +/- switch the SWEEP GENERATOR starts the positive or negative going slope of the input signal. With the NORM/TV switch in TV position the TV SYNC SEPARATOR output is applied to the SWEEP GATING LOGIC and the TRIGGER COMPARATOR output is blocked. The HOLD-OFF circuit prevents the SWEEP GENERATOR from responding to a trigger command before the time base capacitors are fully discharged.

The SWEEP GATING logic, in addition to feeding the SWEEP GENERATOR, also feeds the CHANNEL SELECTOR in order to control the ALTERNATE vertical display mode switching. An output also controls the Z AMPLIFIER in order to blank the trace during the flyback of the time-base.

The SWEEP GENERATOR produces the sawtooth waveform that is used for horizontal deflection. The time-base sweep period can be adjusted by the step control TIME/DIV. The VAR. TIME/DIV control provides the variable sweep facility. The X DEFLECTION SELECTOR enables the input to the FINAL X AMPLIFIER to be selected by diode switching networks. Normally, the internal time-base produced by the SWEEP GENERATOR is routed to the FINAL X AMPLIFIER, but in the X via A mode, an output signal from the X via A AMPLIFIER is selected.

The FINAL X AMPLIFIER, which drives the X1 and X2 horizontal deflection plates of the C.R.T. has a X POSITION control potentiometer. In addition the stage has X5 magnifier switch facility, which increases the horizontal gain by a factor of 5.

2.1.3. CRT Display Section

The Z AMPLIFIER receives an input blanking pulse which originates in the channel selector multivibrator and the sweep gating logic. There is also another input to this amplifier from the Z-MOD input socket. Normally, blanking of the trace takes place during flyback of the time-base and also in the chopped mode during switching from one channel to the other. The INTENS control determines the d.c. level fed to the cathode of the CRT. The output of the Z AMPLIFIER is capacitively coupled to the CRT control electrodes.

A HIGH VOLTAGE multiplier provides the final anode potential (-2kV) of the CRT. The TRACE ROTATION circuit, operated by a front panel control, enables adjustment of the sense and strength of current through the trace rotation coil of the CRT. This allows alignment of the trace with the horizontal graticule lines.

2.1.4. Power Supply

The mains supply is transformed by means of a TRANSFORMER and rectified to give d.c. supplies of +100V and -100V.

VOLTAGE REGULATOR stages provides low voltage d.c. outputs of +5V, +12V and feed the various circuits of the oscilloscope.

A 6.3V a.c. secondary winding of the mains transformer supplies the filament of the CRT.

Channel B is almost identical to channel A except that the signal before being applied to the final Y amplifier can be inverted in channel B, if required.

2.2 DESCRIPTION OF THE VERTICAL SECTION

As channels A and B are almost identical, only channel A is described.

2.2.1. Input coupling stage

In the DC position (switch S13 closed) and in the other than 'O' position of the 'O' switch (S12) the signal applied to input socket XI is fed to the high impedance attenuator via R20 and R21.

In the AC position (switch S13 open) and in the other than 'O' position of the 'O' switch (S12), the signal applied to input socket XI is fed to the high impedance attenuator via R21 and d.c. blocking capacitor C20.

Resistor R20 discharges C20 when switch S13 is changed from the AC to the DC position.

In 'O' position of the 'O' switch the input of high impedance attenuator is grounded and signal applied at $X\mathbf{I}$ to the high impedance attenuator is disconnected.

2.2.2 High impedance attenuator

This section of the circuit comprises of the 1,10,100 and 1000 times attenuator.

The 1000 times attenuator is active in the 5.10 and 20V/div attenuator switch (S8) positions; i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising of R32, R33, R35 and parallel capacitors. The signal, reduced by an attenuation factor of 1000 is fed via K25 contact to the PREAMPLIFIER.

The 100 times attenuator is active in the 0.5,1 and 2V/div attenuator switch (S8) positions; i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising R28, R30, R31 and parallel capacitors. The signal, reduced by an attenuation factor of 100 is fed via K23 contact to the PREAMPLIFIER.

The 10 times attenuator is active in the 50mv, 0.1V and 0.2V/div attenuator switch (S8) positions; i.e. the output signal from the coupling stage is applied via K21 to the attenuator section comprising R25, R26, R27 and parallel capacitors. The signal, reduced by an attenuation factor of 10 is fed via K22 contact to the PREAMPLIFIER.

The 1 x attenuator is active in the 5mV, 10mV and 20mV/div positions of S8 i.e. the output signal from the coupling stage is applied via K20 to the attenuator section comprising of R22, R23 and parallel capacitors. In conjunction with these four basic attenuator coefficients switched by reed relays K20 (5mV-20mV), K22 (50mV-0.2V), K23 (0.5V-2V) and K25 (5V-20V), the 1-2-5 attenuator sequence of adjacent ranges is provided by gain switching of PREAMPLIFIER by relays K26 and K27.

2.2.3. Preamplifier

The output of the HIGH IMPEDANCE ATTENUATOR is connected via resistor R38 to the input of a symmetrical impedance converter consisting of two matched FET's V21 in source follower configuration.

Diode V20 protects the FET input against excessive negative voltages. The output of the Impedance Converter is fed to a transistor array D20 which uses series feedback in the emitter for gain control. The current source for this circuit is obtained with a transistor biased by resistors R45, R46 and R48. Resistors R50 and R51, with K26 and K27 released, determine the gain of the stage in the attenuator switch (S8) positions 20mV, 0.2V, 2V and 20V/div.

- When contact K26 closes, R55 is switched into circuit and the gain of the stage increases 2x.

This occurs in the attenuator switch positions 10mV, 0.1V, 1V and 10V/div.

- When contact K27 closes, R56 is switched into circuit and the gain of the stage increases 4x.

This occurs in the attenuator switch positions $5 \, \text{mV}$, $50 \, \text{mV}$, $0.5 \, \text{V}$ and $5 \, \text{V/div}$.

Resistor R52 in this preamplifier stage is used to calibrate the gain in 5mV/diy range. D.C.

balance control potentiometer R57 adjusts the D.C. balance for the 20mV/div switch position. Potentiometer R41 is used to adjust the D.C. balance for 5mV/div switch position.

The final pre-amplifier output stage uses two transistors which are in-built in D20 having their emitters common and uses R62 and R63 as shunt feedback resistors. The Y position control circuit is implemented here using transistors V22 and V23.

Resistor R2 is used for Y position control. The collector current drawn by the transistor array D20 is fixed and is dependent on the current source feeding the emitters of D20. The collectors of V22 and V23 are also connected to the collectors of the series feedback stage in D20. The change in the collector currents of V22 and V23 due to variation of R2 will now be reflected in the collector voltages of shunt feedback stage transistors of D20, which are the output voltages of the preamplifier circuit. Relay K28 (operated in X via A mode) shorts the Y position control R2 in the X via A mode only. The outputs of the preamplifier stage are d.c. coupled to the trigger amplifier circuit via resistors R76 and R77.

2.2.4. Switched amplifier

The output of the preamplifier stage goes to the series stage of the output amplifier consisting of transistors V32, V31 with series feedback resistors R86, R85, R87, R88 is used for gain adjustment in the 20mV/div attenuator switch position. In this stage itself channel switching is incorporated with the help of a switching current source (V30). When the output Q of the flipflop D252 in the CHANNEL SWITCHING circuit goes high, the current source (V32) conducts and switches channel Å 'ON', while output Q of the same flipflop switches channel B 'OFF'.

The output of this stage is passed through a common base stage V35, V33 only to ensure symmetry of Channels A and B, since Channel B uses similar common base stage to invert the input signal before it is applied to final Y amplifier in Normal/Invert mode.

2.2.5. Channel flipflop and logic circuits

The logic circuits used in this unit can have two logic output levels:

A low level or logic 0 between 0V and 0.8V and a high level or logic 1 between 2V and 5V. The unit has two outputs (pins 9 & 8 of D252) which are the Q and \overline{Q} outputs of a D flipflop. A logic 1 in the Q and \overline{Q} of this flipflop switches channel A and channel B respectively. The various switching modes that control these outputs are CH. A ON/OFF CH. B ON/OFF, ADD, the ms/us (for selecting ALT/CHOP mode of operation) and X via A modes.

CH. A ON

When channel A switch S3 is ON, input 3 of NOR gate D251 is at logic 0. Unless the X via A mode is used, the other input 2 of this gate is also at logic 0. Hence the output of this gate is at logic 1.

If channel B switch S4 is OFF, then input 5 of NAND gate D250 is at logic 1 and hence the output 6 of D250 is at logic 0. This causes the D flipflop D252 to be set and its Q output 9 goes high switching ON channel A.

CH. B ON

When channel B switch S4 is ON, input 5 of NAND gate D250 is at logic 0. If channel A switch S3 is OFF, then input 3 of NOR gate D251 is at logic 1 and its output 1 is at logic 0. This means that reset input 13 of flipflop D252 is at logic 0 and hence its \overline{Q} output goes high, switching channel B ON.

CH. A ON, CH. B ON and switch S5 in us position (ALT mode). Here the +12V on S5 is fed via resistors R256 and R258 to provide a logic 1 at input 6 of NOR gate D251. This makes the output of this gate go to logic 0 which blocks the CHOP oscillator D250/10,9,8 and the output D250-8 is at logic 1.

Output 9 of flipflop D303, connected to input 4 of flipflop D252 is at logic 0 during the time base sweep and at logic 1 during the hold-off period. Thus input 13 of NAND

gate D250 is at logic 1 during the time base sweep and at logic 0 during the hold-off period. Therefore, the CLOCK input (pin 11) of flipflop D252 goes from logic 0 to logic 1 at the end of every sweep and changes the state of the flipflop. In this way the display switches alternately between the channels.

CH. A ON, CH. B ON and switch S5 in ms position (CHOP mode). Here the input 6 of NOR gate D251 is at logic 0 since switch S5 is in 'ms' position. Since both inputs 2 and 3 of NOR gate D251 is at logic 0, its output 1 is at logic 1. Since channel B is ON, input 5 of NAND gate D250 is at logic 0. Hence output 6 of the same gate is at logic 1. Since inputs 1 and 2 of NAND gate D250 is at logic 1, therefore its output 3 is at logic 0. Hence output 4 of NOR gate D251 is at logic 1. This enables the CHOP oscillator D250/10,9,8. This oscillator is NAND - schmitt trigger with a RC feedback loop, which produces a 120 KHZ square wave signal on its output pin 8. This is at logic 1 if the oscillator is switched off in single channel or ALT mode.

The oscillator output is fed to input 12 of NAND gate D250.

During the time base sweep the other input (pin 13) of D250 is at logic 1; therefore the inverted chopper pulses are fed to the CLOCK input (pin 11) of flipflop D252. As both the Clear and Preset inputs of the flipflop are at logic 1 (switches S4 and S3 ON) they are inactive. Therefore, due to the feedback connection between output pin 8 and pin 12, the flipflop changes state at every clock pulse. In this way the display switches between the A and B channels at a frequency of 60 KHZ.

The CLOCK input 11 of flipflop D250, which gets the inverted chopper pulses, is also taken to the Z amplifier to blank the display when switching over between the A and B channels.

Add A & B (Channel A & B ON and switch S1 in Add position i.e. pulled). In this mode of operation input 4 of D301 is at logic 1 unless S10 is in X via A mode. (When switch S10 is in X via A position then V421 conducts and input 4 of D301 is at logic 0, so that 'Add' is not selected). The other input 5 of D301 is at logic 1. Therefore, output 6 of D301 is at logic 0. This will take input 10 & 13 of D252 to logic 0 through diodes V252 & V253. Thus output 9 & 8 of D252 are at logic 1. This will switch ON channel A & B both at the same time. Hence the 2 signals in channel A & B respectively are added and then applied to the final Y amplifier.

In add mode since both the channels are switched on and selected there is a demand for double the current as compared to when either is selected. This extra current is supplied by R210, R211, R212 and V210.

	D250/6	D251/1	D250/8	D252/10	D252/13	D252/9	D252/8
	,	×	-	PRESET	CLEAR	Q	Q
AON/BOFF BON/AOFF A & B ON	0 1	0	- -	0	1 0	1 0	0 1
- in ALT mode	1	1	1	1	1	1/0	0/1
(us)				(state o	hanges at e	nd of ever	y sweep)
- in CHOP	1	1	1/0	1	1	1/0	0/1
mode (ms)				(state o	hanges at c	hopper fre	guency)
ADD	-	-	-	0	0	1	1

2.2.6 Final Amplifier

The final Y amplifier consists of V217, V216, V215, V213 which drive the Y1 deflection plate and are balanced by transistors V218, V220, V221, V222 which drive deflection plate Y2. In order to increase stability, the deflection plates are driven via resistors R233 and R235.

In the Y1 plate driven section, V215 & V213 function as a current source. Transistors V217 and V216 form a shunt feedback stage. Two transistors are used in each case so as not to exceed the maximum permissible current and voltage limits of the transistors. In the Y2 plate drive section, V221 and V222 are the current source and V218 and V220 for the shunt feedback stage.

2.3 HORIZONTAL DEFLECTION AND TIME BASE

2.3.1 CH.A trigger pick off

The trigger signal picked off from the output stage of the CH. A PRE-AMPLIFIER is fed via resistors R77 and R76 to the trigger amplifier consisting of V300 and V301 for which V302 forms a current source. When channel A triggering is selected (S16 to A and S18 to INT), then the collector of V300 is applied via diode V310 in the trigger source select to the base of transistor V313. The channel B pick off and EXT trigger input are inhibited by the -12V switched supply that switches off diodes V311 and V307.

The channel A trigger amplifier has another output from the collector of V301 that is used to drive the horizontal amplifier, via the shunt feedback stage using V415, in the X via A mode.

2.3.2 CH. B trigger pick off

The trigger signal picked off from the channel B pre-amplifier is fed to the trigger amplifier consisting of V303 and V305 for which V308 forms the current source. The collector signal from V303 is routed via diode V311 in the trigger source selector to the base of transistor V313. The A channel pick off and the EXT trigger input are inhibited by the -12V switched supply that switches off diodes V310 and V306.

2.3.3 External trigger amplifier

The signal applied to the external trigger input socket (X5) is attenuated by a voltage divider network R351/C340, R352/C341 in the base circuit of emitter followers V351 and V352 connected in cascade. Capacitor C342 serves for d.c. blocking and diode V350 protects transistor V351 against excessive positive input voltage swings.

The emitter of V352 is coupled via C343 to the series feedback stage V353/V355. The collector current of V353 is connected via diode V312 to the base of transistor V313. The A and B internal pick offs are inhibited in the EXT position of S18 by the -12V switched supply via diodes V307 and V306 that blocks V311 and V310 respectively.

2.3.4. Trigger source selector amplifier

The diode networks referred to in the foregoing descriptions of the trigger pick offs and the external trigger amplifier are all associated with the trigger source selector switches S16 and S18, and the method of selecting each trigger source has been described under these headings. The selected trigger signal to the base of transistor V313 is amplified and fed to emitter follower V318. Electrolytic capacitor C302 connects this signal to the comparator circuit which follows.

2.3.5 Level control

The level voltage control R5 permits variation of the trigger level of the signal.

2.3.6. Comparator and sweep gating logic

The trigger level selected by the LEVEL control is applied to the base of V320, which together with V321 forms a differential amplifier comparator circuit, for which V322 acts as a current source. The trigger signal from the emitter follower V318 is fed directly to the base of V321. When this trigger signal exceeds the reference set by the LEVEL control, V320 conducts less and the collector current decreases so the voltage of the shunt feedback stage V323 increases.

This signal is applied to input 13 of NAND schmitt - trigger D300 in the sweep gating logic, and if S19 is in NORM position the inverted signal appears at output 11 of D300. With the +/- switch S17 in "+" position, output 11 of D301 is logic 1 and the EX-OR D301 (output 8) inverts the signal applied to input 10, so triggering is effected on the positive slope. If switch S17 is in "-" position the EX-OR D301 input 9 is at logic 0, so the signal applied to input 10 appears at pin 8. Now triggering is effected on the negative slope.

2.3.7. T.V. Slope selector

—> The output signal from the trigger source selector is fed to the base of transistor V358 which is balanced by transistor V360.

In the positive slope position of S17 the signal on collector of V360 is routed via diode V362 to the base of transistor V363 and the inverted signal on collector of V358 is inhibited by the switched +12V supply which blocks diode V361.

In the negative slope position of S17 the inverted signal on collector of V358 is routed via diode V361 to the base of transistor V363 and the signal on collector of V360 is inhibited as diode V362 is now blocked.

In the NORM position of S19 diodes V356 and V357 conduct and V361 and V362 are blocked.

2.3.8. T.V. SYNC Separator

The line and frame TV trigger pulses from the slope selector stage are passed via V363 to V365 to a low-pass filter for the frame pulses. The low cut-off frequency is selected in the ms position of S5, which connects capacitor C347 across the output, and triggering on TV frame pulses is possible.

In the us position of S5, this capacitor is disconnected and triggering on TV line pulses is now possible.

The output on the collector of V368 applies a logic signal to pin 4 of NAND gate D300.

With TV selected (D300 input 9 at logic 0) the other input (pin 10) of D300 is at logic 1 and therefore output pin 8 is the inverted logic signal.

This signal is inverted again and appears at pin 8 of D301.

2.3.9. Time-Base and Hold-Off Circuit

These two functional blocks are not described separately here because they function inter-dependently.

The time-base is built around the timing capacitor C405, which is always in circuit, and C402 which is switched into circuit via transistor V406 at the low sweep speeds by the +12V on switch S5 (ms position) via V400 and R411.

A constant current from current source V405 charges the capacitor(s) in order to produce a time-base voltage that is linear with respect to time; i.e. a linear sawtooth. The TIME/DIV control (S10) is incorporated in the emitter circuit of the current source transistor V405.

The TIME/DIV controls, R400 (us range) and R401 (ms range), adjust the base voltage of V405 in diodes V401 and V402 respectively. A variable time/div control is obtained with a potentiometer R7 (which can be shorted by switch S11) which is connected to the base circuit of V405 via R415 and controls the base voltage in calibrated time base mode.

The appropriate base control circuit for V405 is selected by the position of the S5 switch, which provides the +12V to either the us or ms position.

The time base capacitors are charged during the time-base sweep.

During this charging time, switching transistor V403, which is controlled by the sweep gating logic, is not conductive. This transistor, which starts to conduct at the end of the sweep, discharges the timing capacitor(s) and takes over the current from V405.

Switching transistor V403 cuts off when the time base is ready to start again.

The sawtooth time base voltage on the timing capacitor is picked-off by a Darlington stage (V410 and V411) and is applied to the X-deflection selector. The output signal from V411 is also applied via R422 to emitter follower V408, which feeds the hold off capacitors C406 and C403. Capacitor C406 is always in circuit and capacitor C403 is switched into circuit by V407 in ms position of S5. The sawtooth on the hold-off capacitor(s) is applied to the input of NAND gate D300/1,2,3. This gate is effectively a schmitt-trigger with a hysteresis of approximately 0.8V. The output of the gate becomes logic 0 if the positive going slope of the input sawtooth reaches a level of IV approximately.

The two D flipflops D303 operate in parallel. The non inverting output of one flipflop (pin 9) feeds the switching transistor V403 via R403 and the Z-amplifier.

The non-inverting output of the other flip-flop (pin 5) is used as probe adjust signal.

Integrated circuit D302 is a retriggerable monostable multivibrator controlled by the trigger pulses from the trigger comparator. When a trigger pulse is received on pin 3, output pin 8 is at logic 1 for 150 ms. This time constant is determined by C350.

The trigger pulses are also routed to the clock inputs (pin 3 & pin 11 of the D303 flipflops).

Free-Run Mode (Without Trigger Pulses)

If no trigger pulses are available at retriggerable monostable input D302-3, then output pin 8 is low and a preset command is given to the D303 flipflops (pins 1 and 13).

The D flipflops now function as inverters and the clear inputs (pins 4 and 10) receive a pulse from the hold-off circuit after the time base sweep, which is inverted to give outputs on pins 5 and 9.

The output pulse on D303-9 causes the switching transistor V403 to conduct at the end of the time-base sweep and during the hold-off period the time base capacitors C402 and C405 are discharged.

In this way, the time base capacitors are alternately discharged and then charged; i.e. the time-base is free-running.

X via A mode

When X via A is selected, the +12V that is available via S10 performs four functions:

- Via diode V317 it inhibits the trigger source input to V318

Via diode V250 a logic 1 is applied to input pin 2 of gate D251, which causes a logic 0 on input 13 of flipflop D252. This results in channel B being switched through for Y deflection purposes and the output of channel A being blocked. Channel A output from vertical preamplifier is routed via V415 to provide the X-deflection signal.

The +12V applied via R452 to the base of V416 switches off this transistor.

The pulses from the internal time base via emitter of V411 are now blocked by diodes V412 and V417.

- The +12V applied via R457 allows switching diodes V418 and V420 to conduct so that the output on collector of transistor V415 is routed to the final X amplifier.

2.3.10. X-Deflection Selector

The selection of the X via A mode has previously been described. When the internal time-base mode is selected, V416 conducts because of the bias current applied to its base via R455 and R452. The positive voltage on the collector of V416 causes switching diodes V412, V417 to conduct, which allows the time base output on emitter of V411 to be applied to the input of the X-amplifier. In this mode, switching diodes V418 and V420 are blocked by the negative potential applied via R455 and R457.

2.3.11. Final X Amplifier

The output signal from the X deflection selector is applied to the base of V462 in the series feedback stage, which consists of V462 balanced by V461. The base circuit of V461 incorporates the horizontal position control (X-POS) R4. The Transistor V460 is the constant current source for this series feedback stage. In the x5 magnifier position

of S6, resistor R470 shunts the emitter resistors R467, R468 to give a 5 times increase of horizontal gain.

The collectors of V461 and V462 are coupled to the output stage. This output stage consists of the shunt feedback stage V470, V468 and current source V467 that feed the X plate via R481, balanced by an identical stage comprising V471, V472 and V473.

Two transistors are employed in each of the shunt feedback stages so that the maximum current and voltage limits of the individual transistors are not exceeded, and to reduce stray capacitances.

Resistors R481 and R486 connecting the outputs to the X-plates of the CRT are inserted to increase stability.

2.4. CRT DISPLAY SECTION

2.4.1. Z Amplifier

The input to the Z amplifier is via R530 to the base of transistor V526 and receives signals from:

the sweep gating logic in order to blank the display during the time-base hold-off period.
 the channel multivibrator in order to blank the display in the chopped mode during the switching from one channel to the other.

The Z amplifier consists of a shunt feedback stage coupled to the Wehnelt cylinder via C520. Diode V522 and resistor R526 provide D.C. restoration.

2.4.2. High Voltage Supply

The high voltage power supply consists of a quadruple voltage multiplier, a voltage divider that produces cathode, control grid and focus potentials for the CRT and a compensation circuit (V515, V516, V517, V518) to compensate h.t. voltage ripple and variations. The voltage quadrupler circuit consists of diodes V510, V511, V512, V513 and capacitors C510, C511, C512, C513. Resistors R6 and R1 control the cathode and grid g3 voltages respectively and in turn provide the intensity and focus controls.

2.4.3. Trace Rotation

The emitter followers V540 and V541 and preset potentiometer R8 determine the sense and strength of the current in the trace rotation coil.

Only one emitter follower conducts at any given time, depending on the position of R8.

2.5. POWER SUPPLY

The mains voltage is applied via double-pole switch S3 to the primary winding of transformer T500, protected by a replaceable thermal fuse F500 and a replaceable cartridge fuse F501.

Provision is made to wire the primary for a nominal voltage of 120V or 220V or 240V. Two full wave bridge rectifiers V501 and V502 across the secondary winding of T500 provide the d.c. voltages for the +12V, -12V, +100V and -100V supplies respectively.

The low voltage supplies +12V and -12V are regulated by two integrated circuits D500 and D501 and smoothed by electrolytic capacitors C502 and C506. The +5V supply is obtained from the +12V supply with the help of a series pass transistor with Zener reference.

Channel A and Channel B are almost identical except Normal/Invert function is applicable to Channel B. When switch S2 is in Normal position R203 one end is open and +12V is applied to base of V133 and V135 through R196 and R195. The signal from switching amplifier is then applied to the final Y amplifier without inversion.

When S2 is in 'Invert' position R202 one end is open and +12V is applied to base of V137 and V136 through R200 & R198. The inverted signal output from collectors of V137 & V136 is then applied to the final Y amplifier.



DISMANTLING THE INSTRUMENT 3.

General Information 3.1.

WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals

> The instrument shall be disconnected from all voltage sources before any adjustment, replacement or maintenance and repair during which the instrument will be opened. If afterwards any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved. Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

ATTENTION This section provides the dismantling procedures required for the removal of components during repair operations. All circuit boards removed from the oscilloscope should be adequately protected against damage, and all normal precautions regarding the use of tools must be observed. During dismantling procedures, a careful note must be made of all disconnected leads that they may be reconnected to their correct terminals during assembly.

Damage may result if the instrument is switched on when a circuit board has been removed, or if a circuit board is removed within one minute after switching off the instrument.

All screws which have to be remounted directly in the housing-parts must NOTE: be fixed with a torque of maximum 1 Nm (10 kg cm).

Removing the top and bottom covers 3.2.

To adjust the instrument it is necessary to remove the top-cover.

- Remove the two carrying-handle mounting screws (Fig. 3.1.).

- Bend the handle outwards and remove it (Fig. 3.1.).

- Remove the two cabinet mounting-screws (Fig. 3.1.). - Press the two buttons at the rear side until the click. (Fig. 3.2.).

- The top-cover will lift now about 2 mm (Fig. 3.2.).

- Now lift vertically the top-cover out of the front and rear-cover (Fig. 3.3.).

- The bottom cover can now be removed.

NOTE: Take care of the handle gears.

Remounting the top-cover 3.2.1.

- Place the top-cover between the front and rear-cover.

- Take care that the side snaps of the top and bottom-cover fix together.

- Press the upper rear side firmly down until the click (Fig. 3.2.).

- Remount cabinet mounting-screws and the handle.

ACCESS TO PARTS FOR CHECKING AND ADJUSTING PROCEDURE 3.3.

The adjusting elements are accessible after removing the top-cover. To remove the top-cover see section 3.2.

NOTE: For adjustments always use an insulated adjustment tool.

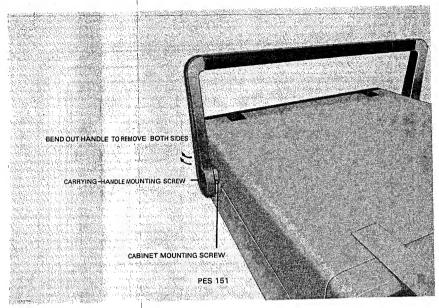


Fig. 3.1 Removing the top Cove.

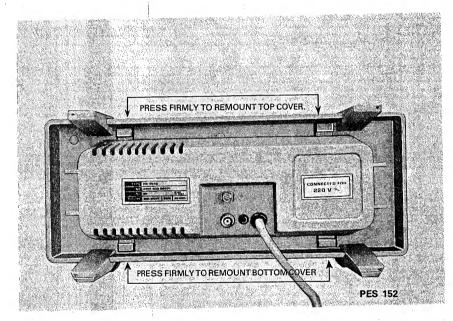


Fig. 3.2 Remounting the top cover

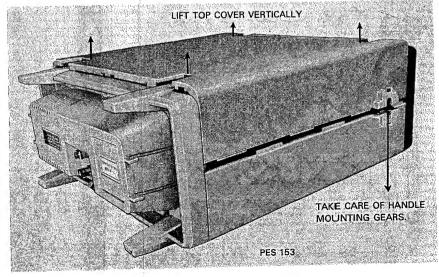


Fig. 3.3 Lifting the top cover

21

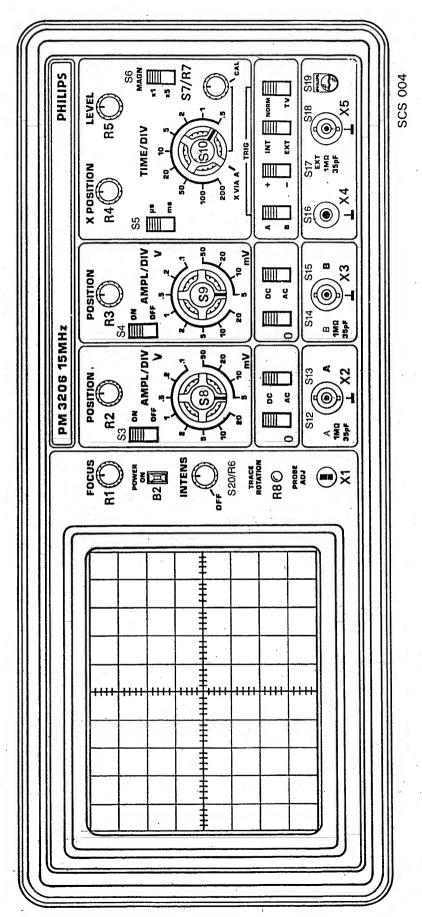


Fig. 4.1 Preliminary Settings of the controls

4. PERFORMANCE CHECK

4.1. General Information



WARNING: Before switching on, ensure that the oscilloscope has been installed in accordance with the instruction outlined in Chapter 4, Directions for use of the Operating Manual.

This procedure is intended to check the instruments specifications. It can be used for incoming inspection to determine the acceptability of newly purchased or recently recalibrated instruments, or to check the necessity of recalibration after a certain operating period. It does not check every facet of the instruments calibration; rather it is concerned primarily with those portions of the instrument which are essential to measurement accuracy and correct operation. Removing the instruments covers is not necessary to perform this procedure. All checks are made from the front panel.

If this test is started a few minutes after switching on, bear in mind that test steps may be out of specification, due to insufficient warming-up time. To avoid this situation, allow the specified warming-up time.

Numerical values without tolerances are typical and represent the characteristics of an average instrument.

The performance checks are made with a stable, well-focussed low-intensity display. Unless otherwise noted, adjust the intensity, focus and trigger-level controls as needed.

NOTE 1: At the start of every objective, the controls always occupy the preliminary settings; unless otherwise stated.

NOTE 2: The input voltage has to be supplied to the A-input; unless otherwise stated.

NOTE 3: Set the TIME/DIV switch to a suitable position; unless otherwise stated.

4.2. Preliminary settings of the controls

- Start this check procedure with NO input signals connected.
- Set the controls as indicated in Fig. 4.1.

4.3. Recommended test equipment

Type of Instrument	Required Specification
Constant amplitude	Freq.: 1 Hz 15 MHz
Sine-wave generator	Constant ampl. of 10 mV p.p 30 V p.p.
Square-wave generator	Freq.: 2 Hz 1 MHz Ampl.: 10 mV 12 V Rise-time 3 nsec. Duty cycle 50%
Time-marker generator	Repetition rate: 200 msec 100 nsec.
Dummy probe 2:1	1 M Ω ± 0.1% // 40 pF.

CHECKING PROCEDURE

STEP	OBJEÇTIVE	INPUT VOLTAGE	SETTINGS	REQUIRÉMENTS	MEASURING RESULTS
4.4.1	POWER ON		T		
4.4.1.1	Start power on	*	- Turn Switch S20 ON	 Starts at selected mains voltage ±10% & mains frequency 45-66 Hz. Pilot Lamp POWER ON lights up. 	
4.4.1.2	Power Consumption			- 28 VA	
4.4.2	CRT SECTION				
4.4.2.1	Intens		- INTENS Control R6	- Normal intensity adjustments	
4.4.2.2	Focus		- Focus Control R1	- Normal focus adjustments	
4.4.2.3	Trace Rotation	×	- Screwdriver adjustment TRACE ROT R8.	- Trace must coincide with central horizontal graticule line; if necessary, readjust potentiometer TRACE ROT R8.	
4.4.3	VERTICAL AXIS				,
4.4.3.1	Display modes		- Position control R2, R3.	- Traces of channel A and channel B in middle of the screen.	
		Sine-wave signal, 300 mVpp, 1 kHz to	- Switch S4 to OFF - Switch S16 to 'A' position CH.A.AMPL/DIV. (S8) to O.IV/DIV.	- Signal of 3 div. visible on the screen.	
	-	A & B input	- Switch S12 to 'O' position - Switch S3 to OFF & S4 to ON.	- Trace in middle of the screen.	
-			-Switch S16 to 'B' position. CH. BAMPL/DIV (S9) to	- Signal of 3 div. visible on the	
·	, et	-	- Pull R3 to invert 'B'.	- Inverted signal of 3 div. visible on the screen.	
:		'	- Press R3 to get Normal B - Switch S3 to ON	on the second	
			- Pull R2 to Add A & B.	- Signal of 6 div. visible on the screen	

		20 1118		
5	00 01 7	2 ms		
) 1			
	- Switch S5 to ms			
	200	100 µs		
	20 50			
	10			
	2 -	1 µs 2 µs		
	0.5	0.5 µs		
- Coefficient error ±5%(c.i.±0.5 div. over 10 div. screen width)	- TIME/DIV. Switch position:	Repetition time		
	- Switch S5 to us.			
	- Switch S4 OFF	*		
	- Var. TIME/DIV. control S7/R7 to CAL.	Marker pulse signal to A input	Time Coefficients	4.4.4.1
			HORIZONTAL AXIS	4.4.4
- Trace height > 4.2 div.	- Switch S5 to ms or us.			
	- Switch S13(S15) to DC.	0 Hz - 15 MHz	•	
- Trace height > 4.2 div.	- Switch S5 to ms or µs.	10 Hz - 15 MHz		
 Adjust the sine wave amplitude for a trace height of 6 div. 		Frequency: 100 kHz		
	- Switch S5 to us.	Sine wave signal to A(B) input	Band Width	4.4.3.6
REQUIREMENTS	SETTINGS	INPUT VOLTAGE	OBJECTIVE	STEP

	4.4.5.3 Bandwidth		3-		4.4.5.2 Phase shift	4.4.5.1 X via A	4.4.5 HORI		4.4.4.2 X Ma			STEP OBJECTIVE	
-	width				shift	>	HORIZONTAL AMPLIFIER		X Magnifier		-	ĊTIVE	
	Sine wave signal I kHz to A input			Frequency 10 kHz	As 4.4.5.1.	Sine wave signal, 600 mVpp, 1 kHz to A and B input.	4.7	Repetition time 100 nsec.	Marker pulse signal to A input repetition time	50 ms 100 ms 200 ms	Repetition time (cont.)	INPUT VOLTAGE	
- Switch S10 to X via A - Switch S13 to DC	- Switch S4 to OFF	,			- As 4.5.1.	- Switch S10 to X via A - Switches S13 & S15 to DC		- Switch So to ins - TIME/DIV to 1 - Switch S6 to X5 Switch S5 to µs - TIME/DIV. to 0.5	Switch S4 to	50 100 200	- TIME/DIV Switch position (cont.)	SETTINGS	
ē	- Adjust input voltage for a PE trace width of 8 div.			- Phase shift 3° (c.i. 0.4 div.)	- Adjust the input voltage for a deflection of 6 div.	 A line is visible with an angle of 45° with respect to the horizontal graticule line 	0.6	- Coefficient error ±7%(c.i±0.7 div) over 10 div. screen width)	- Coefficient error ±7%(c.i±0.7 div) over 10 div. screen width			REQUIREMENTS	
 	PES 105	*	SO:4 div			,						MEASURING RESULTS	

STEP	OBJECTIVE	INPUT VOLTAGE	SETTINGS	REQUIREMENTS	MEASURING RESULTS
9.4.4	TRIGGERING				
4.4.6.1	Trigger source	Sine wave signal, 1kHz	- S4 to OFF		
	a 8	wave signal 800 Hz to	- Switch S16 to A	- Well triggered display of	
\$* + \$* *			- Adjust the input signals for a trace height of 6 div.	Cidille! A	
	-		- Switch S16 to B	- Well triggered display of channel B	
4.4.6.2	Trigger source EXT	Sine wave signal, 600mV, IkHz to A input and EXT input.	- Switch SI8 to EXT.	- Well triggered display	
4.4.6.3	Slope	Sine wave signal, 600mV, I kHz to A input.	- Switch S17 to "+"	- Signal triggers on positive going edge.	,
			- Switch S17 to "-"	- Signal triggers on negative going edge.	
h*9*h*h	TV triggering	TV signal to A input, sync. pulse I div.	- Switch S19 to TV.	- Well triggerd display.	
4.4.6.5	Sensitivity int.	Sine wave signal 100 kHz to A input		- Signal triggers at 0.75 div.	
9.9.4.4	Sensitivity EXT	Sine wave signal, 100kHz to A input and EXT input	- Switch S18 to EXT	- Signal triggers at 0.75 vpp.	
4.4.6.7	Level range	Sine wave signal, 4V freq. 1kHz to A input	- AMPL/DIV to IV	- Signal of 4 div. visible on the screen.	
			- AMPL/DIV to 50 mV - LEVEL control R5	- Signal triggers in the most extreme positions of R5.	
4.4.7	CALIBRATION			- Signal available for probe adjustments.	

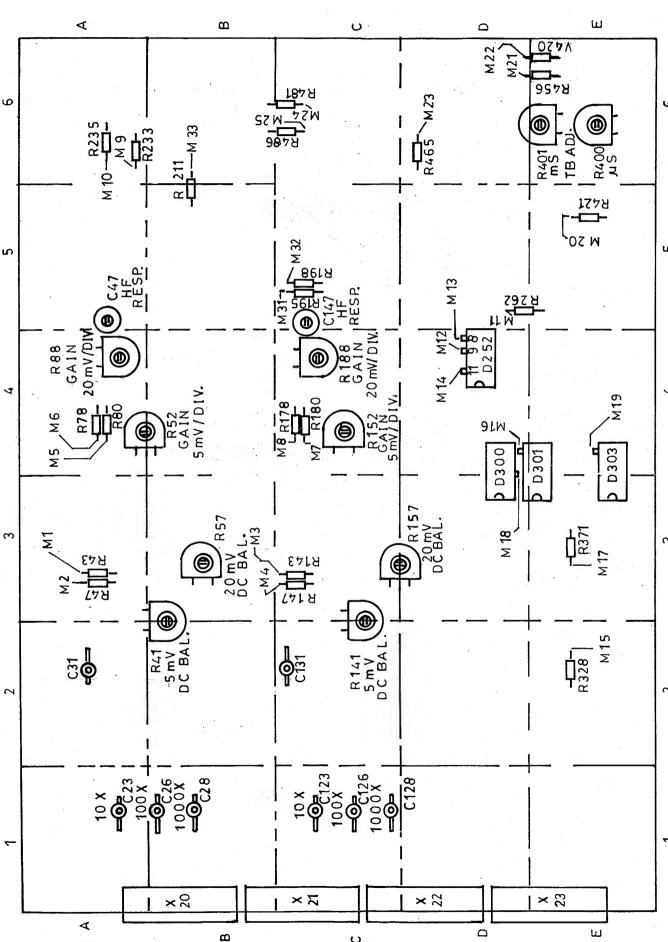


Fig. 5.1 Adjusting elements amplifier unit.

CHECKING AND ADJUSTING 5.



SCS 005

WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals

The instrument shall be disconnected from all voltage sources before any adjustment, replacement of maintenance and repair during which the instrument will be opened. If afterwards, any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved. Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

5.1. General Information

The following information provides the complete checking and adjusting procedure for the

As various control functions are interdependent, a certain order of adjustment is often necessary. The procedure is, therefore, presented in a sequence which is best suited to this order, cross-reference being made to any circuit which may affect a particular

Before any check or adjustment, the instrument must attain its normal operating tempera-

- Where possible, instrument performance is checked before and adjustment is made.
- Warming-up time under average conditions is 15 minutes.
- All limits and tolerances given in this section are calibration guides and should not be interpreted as instrument specifications unless they are also published in chapter 1.6
- Tolerances given are for the instrument under test and do not include test equipment
- The most accurate display adjustments are made with a stable, well-focussed, lowintensity display. Unless otherwise noted, adjust the Intensity, Focus and Trigger Level controls as needed.

Recommended test equipment 5.2.

As indicated in chapter 4.3. Additional equipment for the checking and adjusting procedure: Digital multimeter e.g. PM 2518X. Trimming tool set e.g. Philips 800 NTX.

5.3. Preliminary settings of the controls

As indicated in chapter 4.2.

No.	Adjustments		Preparation		Voltages to apply	Adjustir	ng Element	Adjusting Data
		Controls	Description	Position	to X2 (channel A) and X3 (channel B)	Number	Location	
			X1/X5	X1				
1.	Intensity	56 53	CH. A ON/OFF	ON				The spot just
		58	AMPL/DIV. A	50 mV/div.		R522	Power	vanishes
- 1		54	CH. B ON/OFF	ON			Supply	
ĺ	-	S9	AMPL/DIV. B	50 mV/div.			Unit	
1		S10	TIME/DIV.	X via A				
	l	R6	INTENS	anti-clockwise	*	1	}	
		R3 R4	POSITION B] POSITION X]	Adjust spot to centre of screen.	÷			
	*	K4	FOSITION X J	to contro or serecin		†	*	
2.	Trace	S3	CH. A ON/OFF	ON			Front panel	Trace in para-
	Rotation	S4	CH. B ON/OFF	OFF	•		TRACE	llel with hori-
- 22		S5	us/ms	ms	`	0.00	ROT	zontal grati-
		56	X1/X5 TIME/DIV:	X1 .5	·	R\text{R}		cule line.
		S10 R2	POSITION A]	Adjust		1		
-		R4	POSITION X]	trace to centre of	. •			
	*	*		screen		-		
	*		12/22	20	*	D64 (D464)	50/7	
3,	DC Balance	S13 (S15)	AC/DC BNC	DC Short-circuited		R41 (R141)	B2/3 (C2/3)	Trace jump 1/2 div.
- 6	Channel A	X2 (X3) S3 (S4)	Channel A(B)	Siloit-circuited	0		(02/3)	1/2 div.
	(channel B in	53 (54)	ON/OFF	ON				
	brackets)							
*		S8 (S9)	AMPL/DIV. A(B)	5mV < = > 10 mV	χ.		· `.	
				alternately				
7.0	DC Balance	S13 (S15)	AC/DC		*			
3a.	Adjustments	X2 (X3)	BNC	Same as in 3 above				Trace jump
	for 20m V/div.	S3 (S4)	CH. A(B) ON/OFF	20mV < = > 50mV	*	R57 (R157)	B3 (C3/D3)	1/2 div.
		S8 (S9)	AMPL/DIV. A(B)	alternately			9 *	
		*.	92			Pag (5.400)	1	
4.	Gain CH. A(B)	S3 (S4)	CH, A(B) ON/OFF	ON	120 mVpp	R88 (R188)	A4 (C4)	Amplitude 6
-1-	30 1/-#	S4 (S3)	CH. B(A) ON/OFF	OFF	2 kHz	0		div.,
	20 mV/div	S5	us/ms	us	* * *		* 1	
4		S10	TIME/DIV.	200	*			
-	ė.	S16	A/B	A(B)				•
		S18 ·	INT/EXT	INT				
	•	519	NORM/TV	NORM	. *	-		
	7.	S8 (S9)	- AMPL/DIV. A(B)	200 mV/div.	** **	1		
4a.	Gain CH. A(B)		All settings, same as	* -	30 mVpp - 2 kHz	R52 (R152)	A4/B4 (C4)	Amplitude 6
,	5 mV/div.		above except S8 (S9)					div.
		S8 (S9)	AMPL/DIV. A(B)	5mV/div.				
_			us/ms	us				
5.	Square wave response A	S5 S6	X1/X5	X1/X5	Position			Topside of
	(for channel B	S10	TIME/DIV.	208	2 kHz Square wave	**		square wave
-	in brackets)	S13 (S15)	AC/DC	DC	S8 (S9)			in parallel
	٠٠.	516	A/B	A(B)	58 (S9) on X2(X3)			with graticule
	*	917	+/-	+	10 mV - 60 mV	C31 (C131)	A2 (C2)	line.
		S18	INT/EXT NORM/TV	INT NORM	0.1 V - 0.6 V	C23 (C123)	A1 (C1)	
	·	S19 S4 (S3)	CH. B(A) ON/OFF	OFF	1 V - 6 V 10 V - 60 V	C26 (C126) C28 (C128)	B1 (C1) B1 (C1)	
0		J4 (32)	2 25.7 3.7 3.7			025 (0120)		•
6.	HF response	S5	us/ms	us	120 mV - 1 MHz	1 *		Pulse drop
	channel A	S8 (S9)	AMPL/DIV.	10mV	square wave	C47 (C147)	A5 (C5)	≤ 3%
	(channel B in	S10	TIME/DIV.	0.5 A(B)	Rise time	-	-	Ringing ≤5%
	brackets)	S16	A/B	א(ס)				
7.	Time	S 5	us/ms	ms	Apply pulse marks of			8 puises per
••	Coefficient	56	X1/X5	X1	2 ms	R401	E6	8 div.
	ms	S10 ·	TIME/DIV.	2				
		\$7/R7	VAR TIME/DIV.	CAL			1	
7 _a .	Time	S 5	ns/wis	us	Apply pulse marks of	R400	E6	8 pulses per
	Coefficient	S10	TIME/DIV.	2	2 us			8 div. *
	us	57/R7	VAR TIME/DIV	CAL	I			

	Adjusting	Element	Adjusting Data
	Number	Location	
	R522	Power	The spot just vanishes
· (C		Supply Unit	
		Front panel TRACE ROT	Trace in para- llel with hori- zontal grati- cule line.
	Ř8		cue ine.
	R41 (R141)	B2/3 (C2/3)	Trace jump 1/2 div.
141		\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \	Trace jump
	R57 (R157) R88 (R188)	B3 (C3/D3)	1/2 div. Amplitude 6
	*		div. ₁
	R52 (R152)	A4/B4 (C4)	Amplitude 6
	-		div. Topside of
	C31 (C131)	A2 (C2)	square wave in parallel with graticule line.
	C23 (C123) C26 (C126) C28 (C128)	A1 (C1) B1 (C1) B1 (C1)	Pulse drop
	C47 (C147)	A5 (C5)	≤ 3% Ringing ≤5%
	R401	E6	8 pulses per 8 div.
	R4QD	E6	8 pulses per 8 div.

Fig. 6.1 Removing the power supply unit

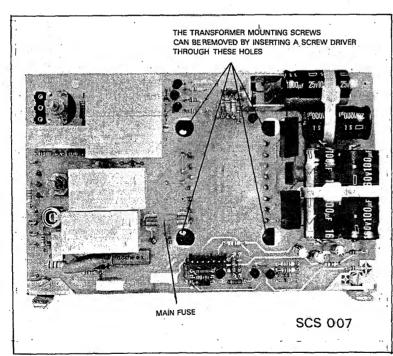


Fig. 6.2 Power supply board

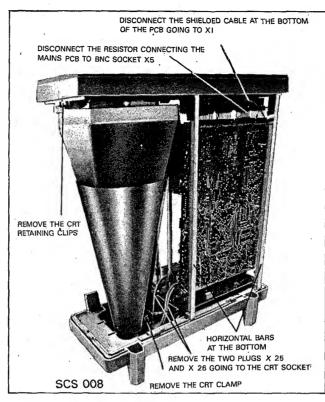


Fig. 6.3 Removing the main PCB and CRT.

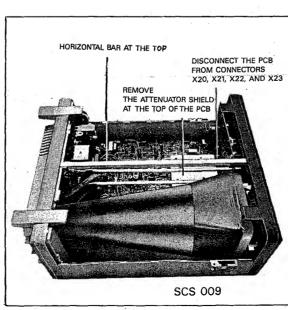


Fig. 6.4 Removing the main PCB unit.

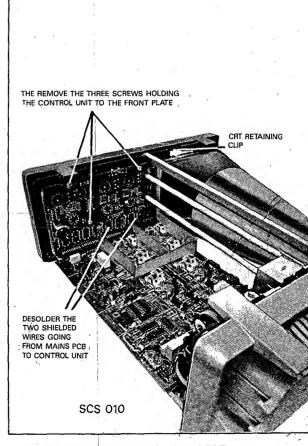


Fig. 6.5 Removing the control unit and CRT.

34

6.1.



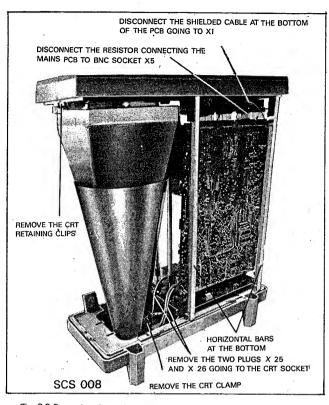


Fig. 6.3 Removing the main PCB and CRT.

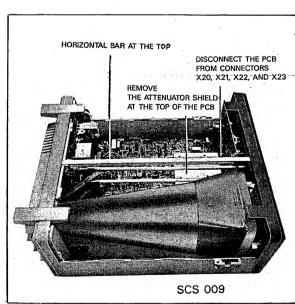


Fig. 6.4 Removing the main PCB unit.

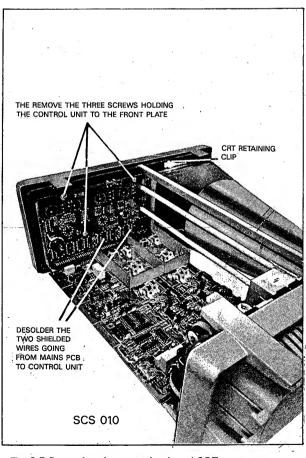


Fig. 6.5 Removing the control unit and CRT.

CORRECTIVE MAINTENANCE

6.1. REPLACEMENTS



WARNING: The opening of covers or removal of parts, except those to which access can be gained by hand, is likely to expose live parts, and also accessible terminals may be live.

The instrument shall be disconnected from all voltage sources before any adjustment, replacement or maintenance and repair during which the instrument will be open.

If afterwards any adjustment, maintenance or repair of the opened instrument under voltage is inevitable, it shall be carried out only by a qualified person who is aware of the hazard involved. Bear in mind that capacitors inside the instrument may still be charged even if the instrument has been separated from all voltage sources.

Standard parts

Electrical and mechanical replacement parts can be obtained through your local Philips organisation or representative. However, many of the standard electronic components can be obtained from other local suppliers.

Before purchasing or ordering replacement parts, check the parts list for value tolerance, rating and description.

Physical size and shape of a component may affect instrument performance, particularly at high frequences. Always use direct-replacement components, unless it is known that a substitute will not degrade instrument performance,

Special parts

In addition to the standard electronic components, some special components are used. These components are manufactured or selected by Philips to meet specific performance requirements.

Transistors and integrated circuits

Transistors and I.C.'s (integrated circuits) should not be replaced unless they are actually defective. If removed from their sockets during routine maintenance return them to their original sockets. Unnecessary replacement or switching of semiconductor devices may affect the calibration of the instrument. When a transistor is replaced, check the operation of the part of the instrument that may be affected.



WARNING: Handle silicone grease with care. Avoid getting silicone grease in the eyes. . Wash hands thoroughly after use.

Any replacement component should be of the original type or a direct replacement. Bend the leads to fit the sockets and cut the leads to the same length as on the component being replaced.

6.1.1. Replacing the mains fuse

To replace the mains fuse F501 which is located on the power supply board, the top cover should first be removed as described under Sec. 3.2. The mains fuse which is a slow blow fuse, can be removed from its holder, and replaced, if necessary. (See fig. 6.2).

6.1.2. Replacing Mains Cord or Transformer

- Unlock the "INTENS" and "FOCUS" extension shafts on the power supply side as follows:

Remove the plastic tubing which covers the coupling between the shaft and the potentiometer. (See fig. 6.1.)

Loosen the two socket set screws on the coupling with an allen key (M3) of 1.5 mm across flats.

WARNING: The intensity and focus potentiometers and their shafts are at 2000V with respect to earth. So a protective plastic tubing covers these shafts. While unlocking the "INTENS" and "FOCUS" extension shafts, the instrument has to be disconnected from voltage sources.

- Remove the four transformer mounting screws. (See fig. 6.2.).
- Remove the two screws at the top of the power supply unit holding it to the rear cover (See fig. 6.1.).
- Remove the two screws at the bottom of the power supply unit holding it to the horizontal bars.
- Now the power supply unit is free and the back side of this unit is now accessible. The transformer can now be desoldered and replaced if necessary. Before replacing the transformer the thermal fuse on the transformer can be checked.
- To replace the mains cord, desolder the two wires of the mains cord from switch \$20/R6 and also the safety earth terminal.
- Remove the grommet at the rear.
- Solder the mains cord and fix the grommet.

6.1.3. Removing and replacing components on the main PCB

To remove and replace components on the main PCB, the top and bottom cover should first be removed as described in Sec. 3.2.

- All components on the main PCB can now be accessed.

To replace the main PCB, the following procedure should now be adopted:

- Remove the screw holding the attenuator shield at the top of the PCB and remove the shield (See fig. 6.4.).
- Remove the six screws holding this PCB to the horizontal bars.
- Remove the attenuator shield at the bottom of the PCB by removing the screw which holds it to the PCB.
- Remove the two plugs X25 and X26 going to the CRT socket.
- Disconnect the connector X501 from the power supply PCB (Fig. 6.1.).
- Disconnect the wire connecting the main PCB to BNC socket X5. (Fig. 6.3.).
- Disconnect the shielded cable at the bottom of the PCB going to X1 (Fig. 6.3.).
- Desolder the two shielded wires going from the main PCB to the control unit (Fig. 6.5.).
- Slide the PCB towards the rear of the instrument and thus disconnect this PCB from connectors X20, X21, X22 and X23 (See fig. 6.4.).
- The main amplifier board can now be lifted and replaced.

6.1.4. Removing and replacing components on the control unit

- To remove or replace components on the control unit, first remove the main PCB as described in Sec. 6.1.3.
- Remove the three screws holding the control unit to the front plate. (Fig. 6.5.).
- Remove all the knobs on the front panel.

6.3. RECALIBRATION AFTER REPAIR

After any electrical component has been replaced the calibration of that particular circuit should be checked, as well as the calibration of other closely related circuit. Since the power supply affects all circuits, calibration of the entire instrument should be checked if work has been done in the power supply or if the transformer has been replaced.

6.4. INSTRUMENT REPACKAGING

If the instrument is to be shipped to a Service Centre for service or repair, attach a tag showing owner (with address) and the name of an individual at your firm that can be contacted. The Service Centre needs the complete instrument serial number and a fault description.

Save and re-use the packing in which your instrument was shipped. If the original packing is unfit for use or not available, repack the instrument in such a way that no damage during transport occurs.

6.5. TROUBLE-SHOOTING

6.5.1. Introduction

The following information is provided to facilitate trouble shooting. Information contained in other sections of this manual should be used along with the following information to aid in locating the defective component. An understanding of the circuit operation is helpful in locating troubles, particularly where integrated circuits are used. Refer to the Circuit Description section for this information.

6.5.2. Trouble-Shooting Hints

If a fault appears, the following test sequences can be used to find the defective circuit part:

- Check if the settings of the controls of the oscilloscope are correct. Consult the operating instructions in the Operating Manual.
- Check the equipment to which the oscilloscope is connected and the interconnection cables.
- Check if the oscilloscope is well-calibrated. If not refer to section 5 (checking and adjusting).
- Visually check the part of the oscilloscope in which the fault is suspected. In this way, it is possible to find faults such as bad soldering connections, bad interconnection plugs and wires, damaged components or transistors and IC's that are not correctly plugged into their sockets.
- Location of the circuit part in which the fault is suspected: the symptom often indicates this part of the circuit. If the power supply is defective the symptom will appear in several circuit parts.

After having carried out the previous steps, individual components in the suspected circuit parts must be examined:

- Transistors and diodes. Check the voltage between base and emitter (0.7 Volt approxin conductive state) and the voltage between collector and emitter (0.2 Volt approxin saturation) with a voltmeter or oscilloscope. When removed from the p.c.b. it is possible to test the transistor with an ohmmeter since the base/emitter and base/collector junctions can be regarded as diodes. Like a normal diode, the resistance is very high in one direction and low in the other direction. When measuring take care that the current from the ohmmeter does not damage the component under test. Replace the suspected component by a new one if you are sure that the circuit is not in such a condition that the new one will be damaged.
- Integrated circuit. In circuit testing can be done with an oscilloscope or voltmeter. A good knowledge of the circuit part under-test is essential. Therefore, first read the circuit description in Section 2.
- Capacitors. Leakage can be traced with an ohmmeter adjusted to the highest resistance range. When testing take care of polarity and maximum allowed voltage. An open capacitor can be checked if the response for AC signals is observed. Also a capacitance meter can be used: compare the measured value with value and tolerance indicated in the parts list.

- Resistors. Can be checked with an ohmmeter after having unsoldered one side of the resistor from the p.c.b. Compare the measured value with value and tolerance indicated in the parts list.
- Coils and transformers. An ohmmeter can be used for tracing an open circuit. Shorted or partially shorted windings can be found by checking the wave-form response when HF signals are passed through the circuit. Also an inductance meter can be used.

NOTE: If a component must be replaced always use a direct-replacement. If not available use an equivalent after carefully checking that it does not degrade the instrument's performance. See also section 6.1. (replacement).

After replacement of a component the calibration of the instrument may be affected due to component tolerances. If necessary do the required adjustments.

6.6. ADAPTION OF MAINS (LINE) VOLTAGE

Before opening any part of the instrument it must be disconnected from all voltage sources. Access to the power supply unit:

Remove the two handle mounting screws.

- Bend the handle brackets outwards and remove it.

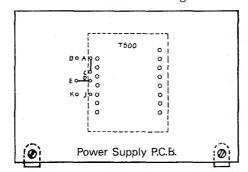
- Remove the two cabinet mounting screws which become visible now.

- Press firmly the two buttons of the rear cover until the click (the top cover will lift for approx. 2 mm).

- Now lift vertically the top cover out of the front and rear-cover.

- The power supply board is accessible now to adapt the mains voltage.

- Connections should be changed as follows:



NOTE:- FOR 240 V \sim CONNECT A-B & D-E FOR 220 V \sim CONNECT A-C & D-E FOR 120 V \sim CONNECT A-B, D-J & E-K

6.7. SAFETY INSPECTION AND TESTS AFTER REPAIR AND MAINTENANCE IN THE PRIMARY CIRCUIT

6.7.1. General directives

- Take care that the creepage distances and clearances have not been reduced.
- Before soldering, the wires should be bent through the holes of solder tags, or wrapped around the tag in the form of an open U, or, wiring shall be rigidly maintained by cable clamps or cable lacing.
- Replacing all insulating guards and plates.

6.7.2. Safety components

Components in the primary circuit may only be renewed by components selected by Philips.

6.7.3. Checking the protective earth connection

The correct connection and condition is checked by visual control and by measuring the resistance between the protective lead connection at the plug and the cabinet/frame. The resistance shall not be more than 0.1 Ω . During measurement the mains cable should be removed. Resistance variations indicate a defect.

6.7.4. Checking the insulation resistance

Measure the insulation resistance at $U=500\ V$ dc between the mains connections and the protective lead connections. For this purpose set the mains switch to **ON.** The insulation resistance shall not be less than $2\ M.\Omega.$

NOTE: 2 M Ω a minimum requirement at 40°C and 95% Relative Humidity. Under normal conditions the insulation resistance should be much higher (10...20 M Ω).

6.7.5. Checking the leakage current

The leakage current shall be measured between each pole of the mains supply in turn, and all accessible conductive parts connected together (including the measuring earth terminal). The leakage current is not excessive if the measured currents from the mentioned parts is ≤ 3.5 mA rms.

6.7.6. Voltage test

The instrument shall withstand, without electrical breakdown, the application of a test voltage between the supply circuit and accessible conductive parts that are likely to become energized. The test potential shall be 1500 V rms at supply-circuit frequency, applied for one second. The test shall be conducted when the instrument is fully assembled, and with the primary switch in the ON position.

During the test, both sides of the primary circuit of the instrument are connected together and to one terminal of the voltage test equipment; the other voltage test equipment terminal is to be connected to the accessible conductive parts.

6.8. SURVEY OF MEASURING POINTS

- To make fault finding easy, test points M1 to M26 are given below. These test points can be located on the PCB with the help of fig. 6.8 and fig. 6.9.
- Apply a sine wave signal of 120 mV peak to peak to YA(YB) input.
- Set AMPL/DIV in 20 mV position.

Measuring point	Location	Values to be measured	Remarks
M1	А3	100 mV p.p. superimposed on 1 V DC approx.	Signal on Channel A.
M2	A3 .	IV DC approx.	
М3	C3	100 mV p.p. superimposed on 1 V DC approx.	Signal on Channel B.
M4 .	C3	1 V DC approx.	
M5/M6	A4	400 mV p.p. superimposed on 8.1 V DC approx.	Signal on Channel A.
M7/M8	C4	400 mV p.p. superimposed on 8.1 V DC approx.	Signal on Channel B.
М9	A6	36 V p.p. on 0 VDC with trace at the centre	Signal on Channel A.
M10	A6 .	36 V p.p. on 0 VDC with trace at the centre	Signal on Channel B.
M11	D5	4 V DC (TTL high voltage) when CH A and CH B is ON and switch S1 in μs position. In ms position of S1, the Chopper frequency of 120 KHZ will be visible.	
M12	D4	TTL high voltage when CH.A is ON and CH.B is OFF	
M13	D4	TTL high voltage when CH.A is OFF and CH.B is ON	
M14	D4 .	Blanking pulse varying with TIME/DIV position.	, ·
M15	E2	1.2 V p.p. square wave	
M16	D4	4.5 V p.p. square wave	
M17	E3	1.2 V p.p. square wave with NORM/TV switch in TV position	
M18	D3/D4	With switch SI in us position needle like pulses and in ms position, no signal	
M19	E4	No trigger signal: square wave depending on TIME/DIV position	
		Trigger Signal: Square wave depending on TIME/DIV position and trigger signal.	
M20	E5	Sweep voitage (sawtooth) - 1 V up to + 1.2 V	

Measuring	Location	Values to be measured	Remarks
point			
M21	D6	Sawtooth	
M22	D6	2.5 V p.p. square wave	
M23	D6	Position control voltage - 1.5 V DC up to +4.5 V DC	
M24/M25	C6	Sawtooth 100 V p.p. (in X via A position - 60 V up to + 40 V)	
M26	Power Supply	Blanking pulse to CRT - 32 V p.p. square wave superimposed on 35 VDC approx.	
M31	C5	12 VDC in Normal 8 VDC in Invert	
M32	C5	8 VDC in Normal 12 VDC in Invert	
M33	B6	18 VDC with Add not selected 22 VDC with Add Selected.	

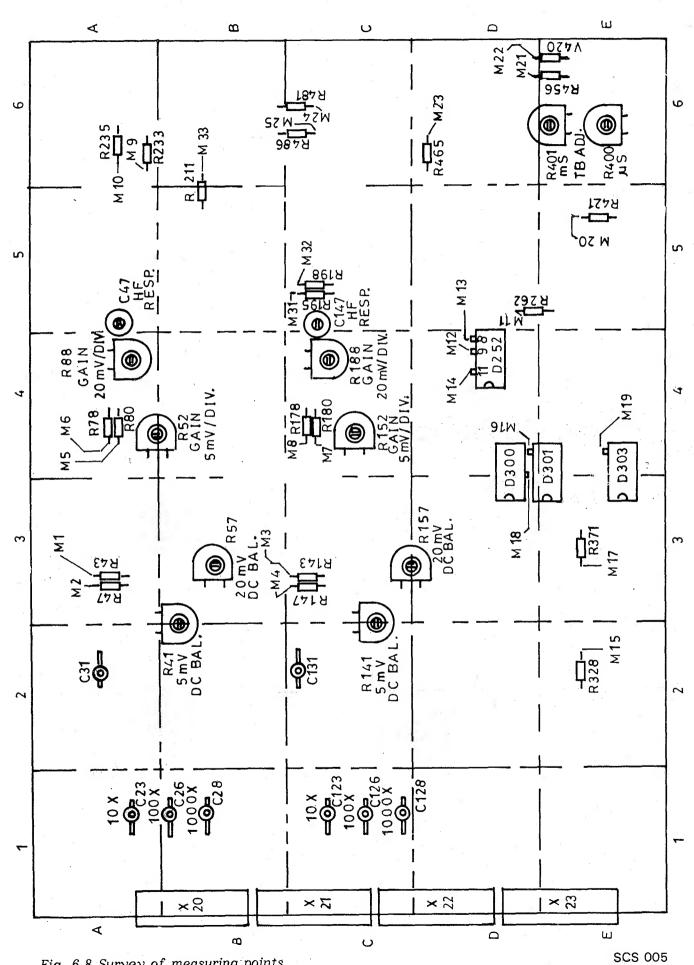
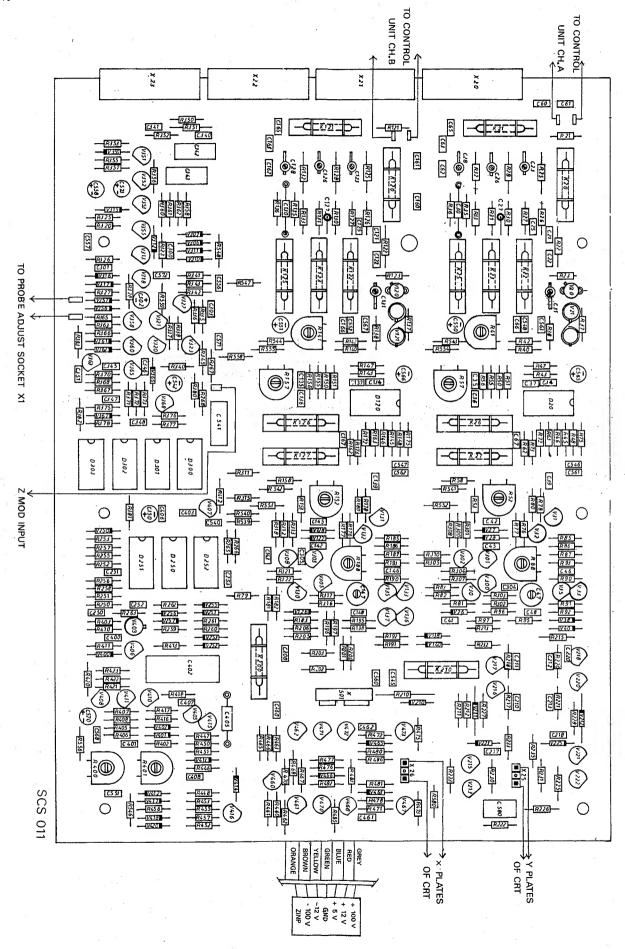


Fig. 6.8 Survey of measuring points



6.9. PIN CONNECTIONS (Refer to Fig. 8.1)

6.9.	PIN	CONNECTIONS (Refer to Fig. 8.1)		
	1.	AMPLIFIER UNIT	CONNECTOR NO./ PIN NO.	CONTROL UNIT
6.9.1		Coil K20, C60	X20-1	V50,V52,S8-3
		Coil K21, C61	X20-2	V70,V71,V72
		•		V55,V57,V70,S8-6
		Coil K22, C62	X20-3	
		Coll K23, C63	X20-4	V60,V62,V71,S8-9
		Coil K25, C65	X20-5	V65,V67,V72,S8-12
		Coil K26, C66	X20-6	V53,V58,V63,V68
		Coil K27, C67	X20-7	V51,V56,V61,V66
		D20 pins 2 & 11, R63,C39	X20-8	R67,V22
		D20 pins 4 & 8, R62,C34	X20-9	R68,V23
		+ 12V	X20-10	+ 12V
				- 12V
		- 12V	X20-11	
		Coil K120, C160	X20-12	V150,V152,S9-3
		Coil K121, C161	X21-1	V170,V171,V172
		Coil K122, C162	X21-2	V155,V157,V170,S9-6
		Coil K123, C163	X21-3	V160,V162,V171,S9-9
		Coil K125, C165	X21-4	V165,V167,V172,S9-12
		-	**	
		Coil K126, C166	X21-5	V153,V158,V163,V168
		Coil K127, C167	X21-6	V151,V156,V161,V166
		D120 pins 4 & 8, R162	X21-7	R168,V123
		D120 pins 2 & 11, R163	X21-8	R167,V122
		K200, K210	X21-9	R207
		V414	X21-10	51-1
		R203	X21-11	S2-1
		R202	X21-12	S2-2
		+ 5.7V	X22:1	+ 5.7V
		R252, C251	X22:2	S3-10
		R250, C250	X22:3	S3-5
		R463, C460	X22:4	R4
		V462, R468	X22:5	56-6
	•	R470	X22:6	S6-5
		D301 pin 4	X22:7	R270, S1-3
		D300 pin 12	X22:8	S17-1, S19-5
			X22:9	517-2,11
		D301 pin 13	.3	R332, R333
		R335, C303	/X22:18	· ·
		V312, V353	X22:11	S18-12
		V306, V307	X22:12	S18-10 ·
		R315, V307, V311	X23:1	S16-1
		R311, V306, V310	X23:2	S16-3
		V362, V360, V357	X23:3	S 17-6
			X23:4	S17-4
		V356, ·V358, V361		S19-8
		V356, V357	X23:5	
		V317, R452, R455	X 23:6	R425/R271
		V367, R378	X23:7	S5-6
		GND	X23:8	GND
		C401,V405,V401,R416,V402	X23:9	R415
		R401,R402,V400,R412	X23:10	S5-7
		R406, R400, R256	X23:11	S5-9
		R417	X23:12	R430-R433,
		1.717		R435-R438, R440
6.9.2	POV	VER SUPPLY UNIT (Refer to Fig. 1		
		+100V	X501-1	+100V
		+12V	X501-2	+12V
		+5V	X501-3	+5V
		OV	X501-4	OV
		-12V	X501-5	-12V
		-100V	X501-6	-100V
		Z input (D202-10)	X501-7	R530
	-	r cooket		
6.9.3	CRI	SOCKET	•	
		R481	X26-1	CRT Pin 7
		R486	X26-2	CRT Pin 9
		R235	X25-2	CRT Pin 11
		R233	X25-1	CRT Pin 12
		1122	•	

POSNR	DESCRI	PTION

С	140	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
Ċ	141		ELECTROLYTE	33UF	_	16V	-5322	124	21431
С	142	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
C	143	CAPACITOR		22PF	2%	100V	5322	122	34196
C	145	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
C	348	CAPACITOR	CERAMIC	3.9NF	10%	100V	4822	122	30098
С	350		ELECTROLYTE	33UF	_	16V	5322	124	21431
С	351	CAPACITOR		10NF	50%	100V	4822	122	31414
c	400	CAPACITOR		150PF	2%	100V	4822	122	31085
С	401	CAPACITOR		0.022UF	80%	63V	4822	122	30103
С	402	CAPACITOR	FOIL	2.2UF	5%	100V	5322	121	44246
С	403	CAPACITOR	FOIL	0.33UF	10%	63V	5322	121	42661
С	405	CAPACITOR	FOIL	2.2NF	5%	63V	4822	121	50415
С	406	CAPACITOR	CERAMIC	1NF	10%	100V	4822	122	30027
С	407	CAPACITOR	CERAMIC	1NF	10%	100V	4822	122	30027
С	408	CAPACITOR	CERAMIC	5.6PF	5%	100V	4822	122	32148
С	460	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	461	CAPACITOR	CERAMIC	0.68PF	0.25PF	500V	4822	122	31213
С	462	CAPACITOR	CERAMIC	0.68PF	0,25PF	500V	4822	122	31213
C,	500	CAPACITOR	FOIL	47NF	10%	400V	4822	121	40239
С	501	CAPACITOR	ELECTROLYTE	1000UF	50%	25V	4822	124	20786
С	502	CAPACITOR	ELECTROLYTE	33UF	- '	16V	5322	124	21431
С	503	CAPACITOR	ELECTROLYTE	33UF	-	.16V	5322	124	21431
С	506	CAPACITOR	ELECTROLYTE	33UF	-	16V	5322	124	21431
С	505	CAPACITOR	ELECTROLYTE	1000UF	50%	25V	4822	124	20786
С	507	CAPACITOR	ELECTROLYTE	100UF	50%	160V	5322	124	24221
C	508	CAPACITOR	ELECTROLYTE	100UF	50%	160V	5322	124	24221
С	510	CAPACITOR	FOIL	150N	10%	1.5KV	5322	121	44329
С	511	CAPACITOR	FOIL	150N	10%	1.5KV	5322	121	44329
$_{\mu}$ C $_{\mu}$	512	CAPACITOR	FOIL	150N	10%	1.5KV	5322	121	44329
С	513	CAPACITOR	FOIL	330NF	10%	250V	4822	121	40344
С	516	CAPACITOR	FOIL	10N	5%	2KV	5322	121	41603
· C	517	CAPACITOR	FOIL	220NF	10%	100V	4822	121	40232
С	518	CAPACITOR	FOIL	100NF	10%	100V	5322	121	40323
С	520	CAPACITOR	CERAMIC	10NF	50%	3000V	5322	122	50091
C	522	CAPACITOR	FOIL	68N	10%	250V	4822	121	41156
C	541	CAPACITOR	FOIL	100NF	10%	100V	5322	121	40323
С	542	CAPACITOR	ELECTROLYTE	33UF	-	16V	5322	124	21431
С	515	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	523.	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	540	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	545	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	546	CAPACITOR		0.022UF	80%	63V	4822	122	30103
С	547	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	548	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
, C	552	CAPACITOR	· · · · · · · · · · · · · · · · · · ·	0.022UF	80%	63V	4822	122	30103
C	553	CAPACITOR	(ERAMIC	0.022UF	80%	6.3V	4822	122	30103
С	556	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
C	557	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
С	560	CAPACITOR	CERAMIC	0.022UF	80%	63V	4822	122	30103
	-								

POSNR DESCRIPTION

R	221	RESISTOR METAL FILM	5K62	- 1%		5322	116	E 740E
R	222	RESISTOR METAL FILM	5K62	1%		5322	116	
R	223	RESISTOR METAL FILM	5K62	1%		5322	116	53495
R	225	RESISTOR METAL FILM	5K62	1%		5322	116	53495
R	226	RESISTOR METAL FILM	5K62	1%			116	53495
	007			1 /0		5322	116	53495
R	227	RESISTOR METAL FILM	121E	1%		4822	116	52955
R	228	RESISTOR METAL FILM	121E	1%		4822	116	52955
R	230	RESISTOR METAL FILM	110K	5%		4822	116	52455
R	231	RESISTOR METAL FILM	110K	5%		4822	116	52455
R	232	RESISTOR METAL FILM	200E	5%		4822	116	52405
R-	. 233	RESISTOR METAL FILM	205R	1%		E700	446	
R	235	RESISTOR METAL FILM	205R	1%		5322	116	53633
R	250	RESISTOR METAL FILM	3K6	5%		5322	116	53633
R	252	RESISTOR METAL FILM	3K6	5%		4822	116	52419
R	253	RESISTOR METAL FILM	3K6			4822	116	52419
			JNO	5%		4822	116	52419
R	256	RESISTOR METAL FILM	3K6	. 5%		4822	116	52419
R	251	RESISTOR METAL FILM	1K	5%		4822	116	52391
R	255	RESISTOR METAL FILM	1K	5%		4822	116	52391
R	257	RESISTOR METAL FILM	- 1K	5%		4822	116	52391
R	258	RESISTOR METAL FILM	1K	5%		4822	116	52391
R	259	PECICIOD METAL CULA					110	1211
R		RESISTOR METAL FILM	10K	5%		4822	116	52452
	260	RESISTOR METAL FILM	10K	5%	* *	4822	116	52452
- R	261	RESISTOR METAL FILM	10K	5%		4822	116	52452
R	262	RESISTOR METAL FILM	7K15	1%		5322	116	80125
R	263	RESISTOR METAL FILM	7K15	1%		5322	116	80125
R	265	RESISTOR METAL FILM	2K43	1%		E 700	.447	00400
R	266	RESISTOR METAL FILM	2K43		·	5322	116	80109
R	267	RESISTOR METAL FILM		1%		5322	116	80109
R	268	RESISTOR METAL FILM	200E	5% F.W.		4822 -	116	52405
R	270	RESISTOR METAL FILM	2K	5.% .		4822	116	52406
11	270	KESISTOR METAL FILM	1K	5%		4822	116	52391
R	300	RESISTOR METAL FILM	10K	5%		4822	116	52452
R	301	RESISTOR METAL FILM	10K	5%		4822	116	52452
R	302	RESISTOR METAL FILM	249R	1%		5322	116	53573
R	303	RESISTOR METAL FILM	249R	1%		5322	116	53573
R-	305	RESISTOR METAL FILM	1E	5%		4822	116	52385
_	70.			<i>77</i> a		4022	110	72707
R	306	RESISTOR METAL FILM	909R	1%		4822	116	53533
R	307	RESISTOR METAL FILM	4K64	1%		5322	116	53212
R	308	RESISTOR METAL FILM	7K32	1%		4822	116	53187
R	310	RESISTOR METAL FILM	1E	5%		4822	116	52385
R	311	RESISTOR METAL FILM	1E	5% <		4822	116	52385
R	312	RESISTOR METAL FILM	10K	E 0/		4000		50450
R	313	RESISTOR METAL FILM		5%		4822	116	52452
R	315	RESISTOR METAL FILM	10K ,	5%		4822	116	52452
R.	316	RESISTOR METAL FILM	1E	5%		4822	116	52385
R	312	RESISTOR METAL FILM	249R	1%		5322	116	53573
	- 14		249R	1%		5322	116	53573
R	318	RESISTOR METAL FILM	7K32	1%		5322	116	553 7 2
R	320	RESISTOR METAL FILM	825R	1%		5322	116	53541
R ·	321	RESISTOR METAL FILM	909R	1%		4822	116	53533
R	322	RESISTOR METAL FILM	4K64	1%			116	5321 2
			· · · · · ·			J766		JJ614

			 					
R	323	RESISTOR METAL FILM	1K05	4.0/		4000	446	50000
R	325	RESISTOR METAL FILM		1%		4822	116	52898
R	326		909R	1%		4822	116	53533
		RESISTOR METAL FILM	2K	5%		4822	116	52406
R	327	RESISTOR METAL FILM	 1K	5%	. "	4822	116	52391
R	328	RESISTOR METAL FILM	1K62	1%	:	5322	116	53257
R	332	RESISTOR METAL FILM	11K5	1%		5322	116	53319
R	333	RESISTOR METAL FILM	1K47	1%	-	5322	116	53185
R	335	RESISTOR METAL FILM	140R	1%		5322	116	53542
R	336	RESISTOR METAL FILM	2K49	1%	1	5322	116	50581
R	337	RESISTOR METAL FILM	36R5	1%		5322	116	80116
R	338	RESISTOR METAL FILM	36R5	1%	.t	5322	116	80116
R	340	RESISTOR METAL FILM	1K54	1%		5322	116	53571
R		RESISTOR METAL FILM	5K1	5%		4822	110	
R	342	RESISTOR METAL FILM	2K15					70126
R	343	RESISTOR METAL FILM		1%		5322	116	53239
			5K1	5%	2	4822	110	70126
R	345	RESISTOR METAL FILM	1K3	1%		5322	116	80102
R	346	RESISTOR' METAL FILM	6K8	5%		¥822	110	73129
R	347	RESISTOR METAL FILM	4K7	5%	- · · · · · · · · · · · · · · · · · · ·	¥822	116	52426
R	350	RESISTOR METAL FILM	360E	5%		822	116	52418
R	351	RESISTOR METAL FILM	845K	1%	5	322	116	80172
R	352	RESISTOR METAL FILM	402K	1%		5322	116	80118
R	353	RESISTOR METAL FILM	301K	1%	5	322	116	53328
R	355	RESISTOR METAL FILM	5K6	5%		822	110	73127
R	356 ·	RESISTOR METAL FILM	26R1	1%		322	116	53723
R	357	RESISTOR METAL FILM	5K6	5%		822	110	73127
R	358	RESISTOR METAL FILM	30K	5%	/	822	116	52466
R	360	RESISTOR METAL FILM	523R	1%				
R	361	RESISTOR METAL-FILM				322	116	80122
			523R	1%		322	116	80122
R	362	RESISTOR METAL FILM	2K32	1%		322	116	80108
R.	363	RESISTOR METAL FILM	10K	5%	4	822	116	52452
R	365	RESISTOR METAL FILM	1K	1%	4	822	116	53108
R	366	RESISTOR METAL FILM	1K	1%	4	822	116	53108
R	367	RESISTOR METAL FILM	3K83	1%	4	822	116	53079
R	368	RESISTOR METAL FILM	2K87	1%		322	116	55279
R	370	RESISTOR METAL FILM	4K02	1%	5	322	116	53558
R	371	RESISTOR METAL FILM	1M	5%	/1	822	116	52493
R	372	RESISTOR METAL FILM	10K	5%		822	116	52452
R-	373	RESISTOR METAL FILM	3K16	1%				
R						822	116	53021
	375 376	RESISTOR METAL FILM	10K	5%		822	116	52452
R	376	RESISTOR METAL FILM	1M	5%	. 4	822	116	52493
R	377	RESISTOR METAL FILM	196K	1%	5	322	116	53661
R	378	RESISTOR METAL FILM	10K	1%	4	822	116	53022
R	380	RESISTOR METAL FILM	2K	5%	4	822	116	52406
R	381	RESISTOR METAL FILM	5K1	5%	4	822	110	70126
R	382	RESISTOR METAL FILM	1K	5%	4	822	116	52391
R	383	RESISTOR METAL FILM	10K	5%	4	822	116	52452
R	400	POTMETER TRIMMER	22K	10%		322	100	10979
R	401	POTMETER TRIMMER	22K	10%		322	100	10979
R	402	RESISTOR METAL FILM	7K15	1%		322	116	54606
R	403	RESISTOR METAL FILM	1K	1%		822	116	53108
, ,	407	RESISTOR WILTAL FILM	117	1 /0	. 4	U	110	JJ 100

	٧	138	DIODE		IN4148			4822	130 -	30621	
	٧	140	DIODE		IN4148			4822	130	30621	
	٧	150	DIODE		IN4148			4822	130	30621	
	٧.	151	DIODE		IN4148			4822	130	30621	
	٧	152	DIODE		IN4148			4822	130	30621	
	٧	153	DIODE		IN4148			4822	130	30621	
•	٧	155	DIODE		IN4148			4822	130	30621	
	٧	156	DIODE		IN4148			4822	130	30621	
	٧	157	DIODE		IN4148			4822	130	30621	
	٧	158	DIODE		IN4148			4822	130	30621	
	V .	160	DIODE		IN4148			4822	130	30621	
	٧	161	DIODE		IN4148			4822	130	30621	
	٧	162	DIODE		IN4148	• *		4822	130	30621	
	٧	163	DIODE		IN4148			4822	130	30621	
	٧	165	DIODE		IN4148			4822	130	30621	
	٧	166	DIODE		IN4148			4822	130	30621	
	٧	167	DIODE		IN4148			4822	130	30621	
	٧ .	168	DIODE		IN4148			4822	130	30621	
	٧	170	DIODE		IN4148			4822	130	30621	
	٧	171	DIODE		IN4148			4822	130	30621	
	٧	172	DIODE		IN4148			4822	130	30621	
	٧	210	DIODE, REFERENCE ZENER		BZX79C22V			4822	130	34441	
	٧	211	DIODE, REFERENCE ZENER		BZX75C1V4			4822	130	34047	
	٧	212	DIODE, REFERENCE ZENER		BZX75C1V4			4822	130	34047	
	٧	213	TRANSISTOR		BF 199			4822	130	44154	
	٧	215	TRANSISTOR		BF 199			4822	130	44154	
	٧	216	TRANSISTOR		BF 199			4822	130	44154	
	٧	217	TRANSISTOR		BF199			4822	130	44154	
	٧	218	TRANSISTOR		BF199			4822	130	44154	
	٧ ,	220	TRANSISTOR	•	BF 199	:		4822	130	44154	
	٧	221	TRANSISTOR		BF 199			4822	130	44154	
	٧	222	TRANSISTOR		BF 199			4822	130	44154	
	٧	223	DIODE, REFERENCE ZENER		BZX79C56V			4822	130	34258	
	٧	225	DIODE, REFERENCE ZENER		BZX79C56V			4822	130	34258	
	٧	250	DIODE		IN4148			4822	130	30621	
	V	251	DIODE		IN4148			4822	130	30621	
	٧	252	DIODE		IN4148			4822	130	30621	
	٧	253	DIODE		İN4148			4822	130	30621	
	٧	255	DIODE		IN4148			4822	130	30621	
	٧	256	DIODE		IN4148			4822	130	30621	
	٧	257	DIODE		IN4148			4822	130	30621	
	٧	300	TRANSISTOR		BF 450	*		4822	130	44237	
	٧	301	TRANSISTOR		BF450			4822	130	44237	
	٧	.302	TRANSISTOR		BC558B			4822	130	44197	
	٧	303	TRANSISTOR		BF450			4822	130	44237	
	V.	305	TRANSISTOR		BF450			4822	130	44237	
	٧	306	DIODE		IN4148			.4822	130	30621	
	٧	307	DIODE		IN4148		*	4822	130	30621	•
	٧	308	TRANSISTOR		BC558B			4822	130	44197	
	٧	310	DIODE		IN4148			4822	130	30621	

V	311	DIODE:	IN4148	4822	130	30621
٧	312	DIODE	IN4148	4822	130	30621
٧	313	TRANSISTOR	BF 199	4822	130	44154
		*	BZX79C5V6	4822	130	34173
٧	315	DIODE, REFERENCE ZENER		4822	130	34167
٧	316	DIODE, REFERENCE ZENER	BZX79C6V2	4022		
٧	317	DIODE	IN4148	4822	130	30621
- V	318	TRANSISTOR	BC548C	4822	130	44196
٧	320	TRANSISTOR	BF 199	4822	130	44154
٧	321	TRANSISTOR	BF 199	4822	130	44154
٧	322	TRANSISTOR	BF199	4822	130	44154
٧	323	TRANSISTOR	BC558B	4822	130	44197
V	350	DIODE	IN4148	4822	130	30621
v	351	TRANSISTOR	BC558B	4822	130	44197
V	352	TRANSISTOR	BC558B	4822	130	44197
V	353	TRANSISTOR	BC558B	4822	130	44197
٧	355	TRANSISTOR	BC558B	4822	130	44197
٧	356	DIODE	IN4148	4822	130	30621
٧	357	DIODE	IN4148	4822	130	30621
٧	358	TRANSISTOR	BC548C	4822	130	44196
٧	360	TRANSISTOR	BC548C	4822	130	44196
٧	361	DIODE	IN4148	4822	130	30621
٧	362	DIODE	IN4148	4822	130	30621
V	363	TRANSISTOR	BC558B	4822	130	44197
v	365	TRANSISTOR	BC558C	4822	130	44196
v	366	DIODE	IN4148	4822	130	30621
٧	367	DIODE	IN4148	4822	130	30621
ý	368	TRANSISTOR	BC558B	4822	130	44197
٧	400	DIODE	IN4148	4822	130	30621
V	401	DIODE	IN4148	4822	130	30621
V	401		IN4148	4822	130	30621
		DIODE				
٧	403	TRANSISTOR	BSX20	4822	130	41705
٧	405	TRANSISTOR	BC558B	4822	130	44197
٧	406	TRANSISTOR	BC548C	4822	130	44196
٧	407	TRANSISTOR	BC548C	4822	130	44196
٧	408	TRANSISTOR	BC548C	4822	130	44196
٧	410	TRANSISTOR	BC548C	4822	130	44196
$^{\prime\prime}$ V	411	TRANSISTOR	BC548C	4822	130	44196
٧	412	DIODE	IN4148	4822	130	30621
٧	413	DIODE	IN4148	4822	130	30621
٧	414	DIODE	IN4148	4822	130	30621
V	415	TRANSISTOR	BC548C	4822	130	44196
٧	416	TRANSISTOR	BC548B	4822	130	44197
٧	417	DIODE	IN4148	4822	130	30621
V	417	DIODE	IN4148	4822	130	30621
				4822	130	30621
٧	420	DIODE	IN4148			
V	421	TRANSISTOR	BC548C	4822	130	44196 44197
٧	460	TRANSISTOR	BC548B	4822	130	44197
Α.	461	TRANSISTOR	BF 423	4822	130	441.96
٧	462	TRANSISTOR	BF423	4822	130	44196
٧	463	DIODE, REFERENCE ZENER	BZX75C2V1	4822	130	34049

POSNR DESCRIPTION

ORDERING CODE

V V V V	465 466 467 468 470	DIODE, REFERENCE ZENER DIODE, REFERENCE ZENER TRANSISTOR TRANSISTOR TRANSISTOR	BZX75C2V1 BZX79C6V2 BF 422 BF 422 BC548C	4822 4822 4822 4822 4822	130 130 130 130 130	34049 34167 41782 41782 44196	
V V 7-2 V V	471 472 473 501 502	TRANSISTOR TRANSISTOR TRANSISTOR DIODE, BRIDGE DIODE, BRIDGE	BC548C BF 422 BF 422 SKB2-08/LSA SKB-08/LSA	4822 4822 4822 5322 5322	130 130 130 130 130	44196 41782 41782 32031 32031	
V V V V	503 504 505 510 511	TRANSISTOR DIODE DIODE, REFERENCE ZENER DIODE DIODE	BC338/16 IN4148 BZX79-B5V6 BY584 BY584	4822 4822 4822 5322 5322	130 130 130 130 130	44121 30621 34173 32274 32274	
V V V V	512 513 517 518 520	DIODE DIODE TRANSISTOR TRANSISTOR DIODE	BY 584 BY 584 BF 422 BC 548C IN4148	5322 5322 4822 4822 4822	130 130 130 130 130	32274 32274 41782 44196 30621	
V V V V V V	521 522 523 525 526 540 541	DIODE DIODE TRANSISTOR DIODE, REFERENCE ZENER TRANSISTOR TRANSISTOR TRANSISTOR	BY 509 BAV21 BF 422 BZX75C1V4 BF 422 BC548C BC548C	4822 4822 4822 4822 4822 4822	130 130 130 130 130 130 130	41485 34189 41782 34047 41782 44196 44196	

INTEGRATED CIRCUITS

D	250	INTEGR. CIRCUIT	74LS132N	5322	209	85201
D	251	INTEGR. CIRCUIT	74LS02N	5322	209	85312
D	252	INTEGR. CIRCUIT	74LS74	4822	209	80782
D.	300	INTEGR. CIRCUIT	74LS132N	5322	209	85201
D	301	INTEGR. CIRCUIT	74LS86N	5322	209	84997
D	302	INTEGR. CIRCUIT	74122N	5322	209	84231
D	303	INTEGR. CIRCUIT	74LS74	4822	209	80782
D	500	INTEGR. CIRCUIT	7812	4822	209	81016
Ď	501	INTEGR. CIRCUIT	7912T	5322	209	81856

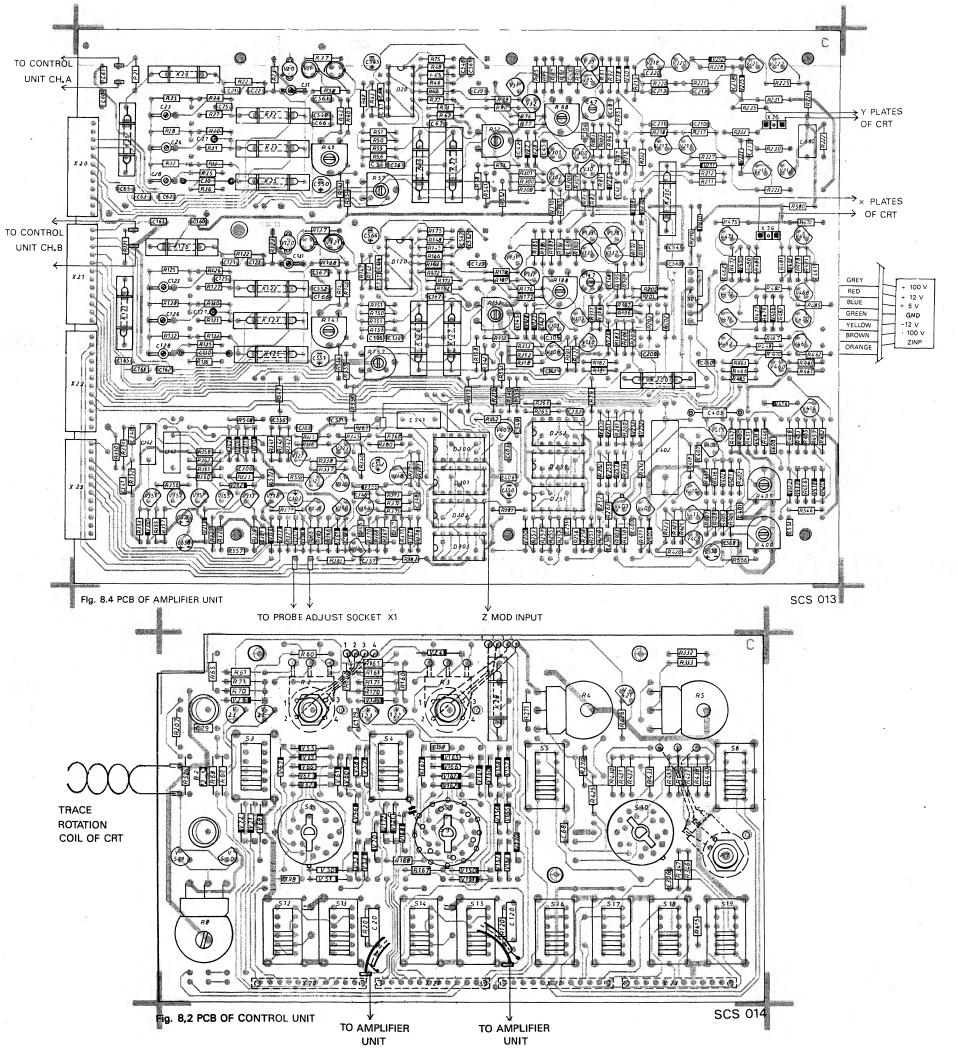
MISCELLANEOUS

PO	POSNR DESCRIPTION			ORDERING CODE				
		BNC CONNECTOR SOLDER TAG HEX. NUT CONTROL UNIT POWER SUPPLY UNIT	5322 5322 5322 5322 5322	267 290 506 216 216	10004 30204 14005 51147 51146			
X	20	MOLEX CONNECTOR	5322	266	40147			
X	20	12 PINS FEM MOLEX CONNECTOR 12 PINS MA	5322	265	40624			
X	21	MOLEX CONNECTOR 12 PINS FEM	5322	266	40147			
X	21	MOLEX CONNECTOR 12 PINS MA	5322	265	40624			
X	22	MOLEX CONNECTOR 12 PINS FEM	5322	266	40147			
Х	22	MOLEX CONNECTOR 12 PINS MA	5322	265	40624	<u>.</u>		
x x	23	MOLEX CONNECTOR 12 PINS FEM	5322	266	40147			
X	23	MOLEX CONNECTOR 12 PINS MA	5322	265	40624			
X	25	MOLEX CONNECTOR HOUSING 2 PINS FEM	5322	265	61106			
X X	25 25	MOLEX CONNECTOR PINS MOLEX CONNECTOR 2 PINS MA	5322 5322	265 265	40623 61107	-		
X	26	MOLEX CONNECTOR HOUSING 2 PINS FEM	5322	265	61106	·		
X X	26 26	MOLEX CONNECTOR PINS MOLEX CONNECTOR 2 PINS MA	5322 5322	265 265	40623 61107			
Χ	501	MOLEX CONNECTOR WITH CABLE	5322	321	22263	(1)		
Χ	· 501	MOLEX CONNECTOR 7 PINS MA	5322	265	40237			
		REED RELAY CONTACT	5322	280	24126			
		CATHODE RAY TUBE 150 BT31	5322	131	20106			
		POWER ON LED	4822	130	31911			
		THERMAL FUSE	5322	252	20114			
F	500	MAINS FUSE	4822	253	30009			
K K K K	20 21 22 23 25	REED RELAY ASSLY	5322 5322 5322 5322 5322	280 280 280 280 280	20276 20276 20276 20276 20276			
K K K	26 27 28 120 121	REED RELAY ASSLY	5322 5322 5322 5322 5322	280 280 280 280 280 280	20276 20276 20276 20276 20276			

POSNR DESCRIPTION

K	122	REED RELAY ASSLY	5322	280	20276
K	123	REED RELAY ASSLY	5322	280	20276
K	125	REED RELAY ASSLY	5322	280	20276
K	126	REED RELAY ASSLY	5322	280	20276
K	127	REED RELAY ASSLY	5322	280	20276
S	3	SLIDE SWITCH	5322	277	24077
S	4	SLIDE SWITCH	5322	277	24077
S	5	SLIDE SWITCH	5322	277	24077
S	6	SLIDE SWITCH	5322	277	24077
S	7	DPST SWITCH	5322	101	40132
S	8	ROTARY AMPL/DIV SWITCH	5322	273	34121
S	9	ROTARY AMPL/DIV SWITCH	5322	273	34121
S	10	ROTARY TIME/DIV SWITCH	5322	273	34119
S	12	SLIDE SWITCH	5322	277	24077
Ś	13	SLIDE SWITCH	5322	277	24077
S	14	SLIDE SWITCH	5322	277	24077
S	15	SLIDE SWITCH	5322	277	24077
S	16	SLIDE SWITCH	5322	277	24077
S	17	SLIDE SWITCH	5322	277	24077
5	18	SLIDE SWITCH	5322	277	24077
S	19	SLIDE SWITCH	5322	277	24077
Т	500	MAINS TRANSFORMER	5322	146	40385

8. CIRCUIT DIAGRAMS and PCB LAY-OUTS.



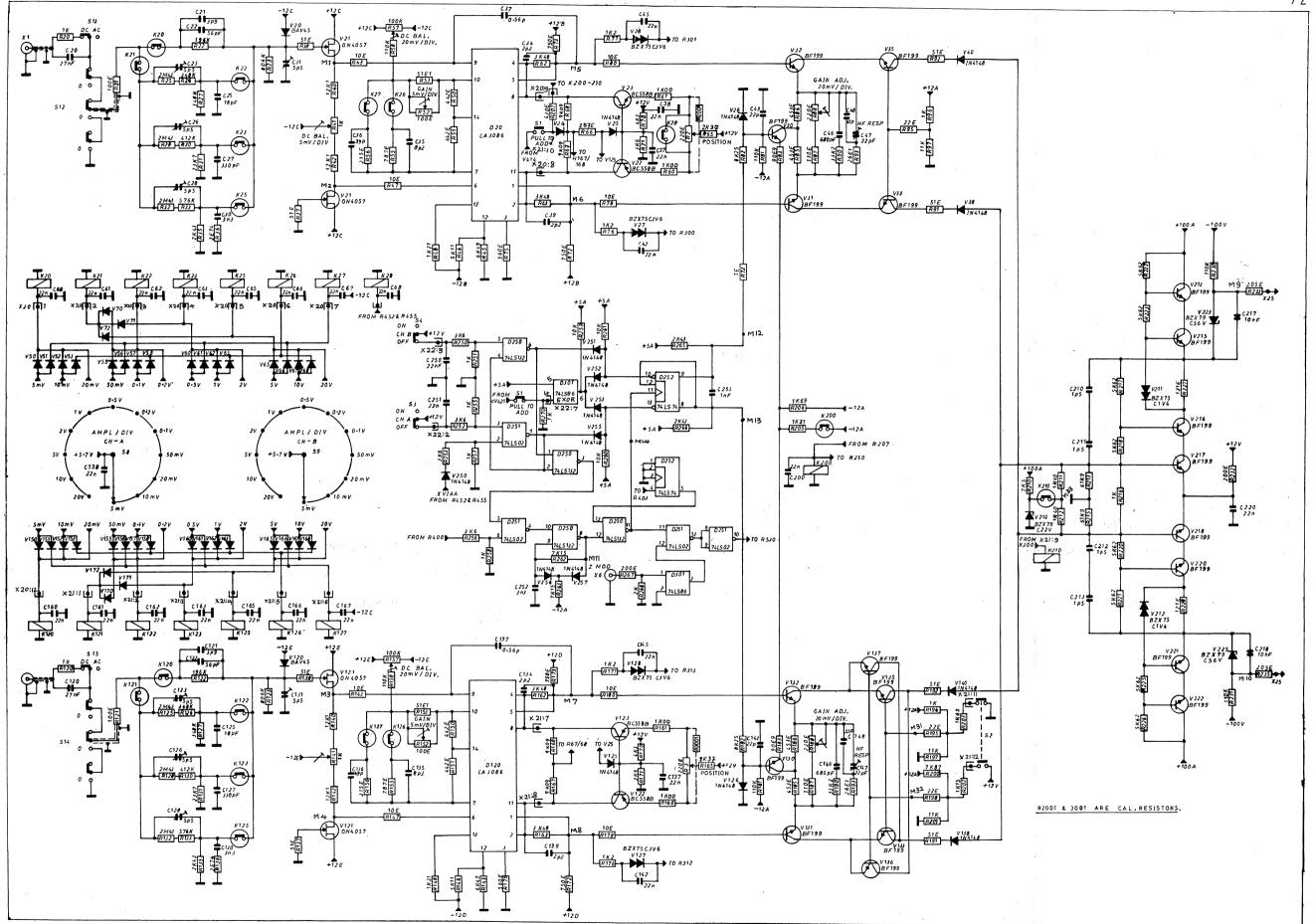
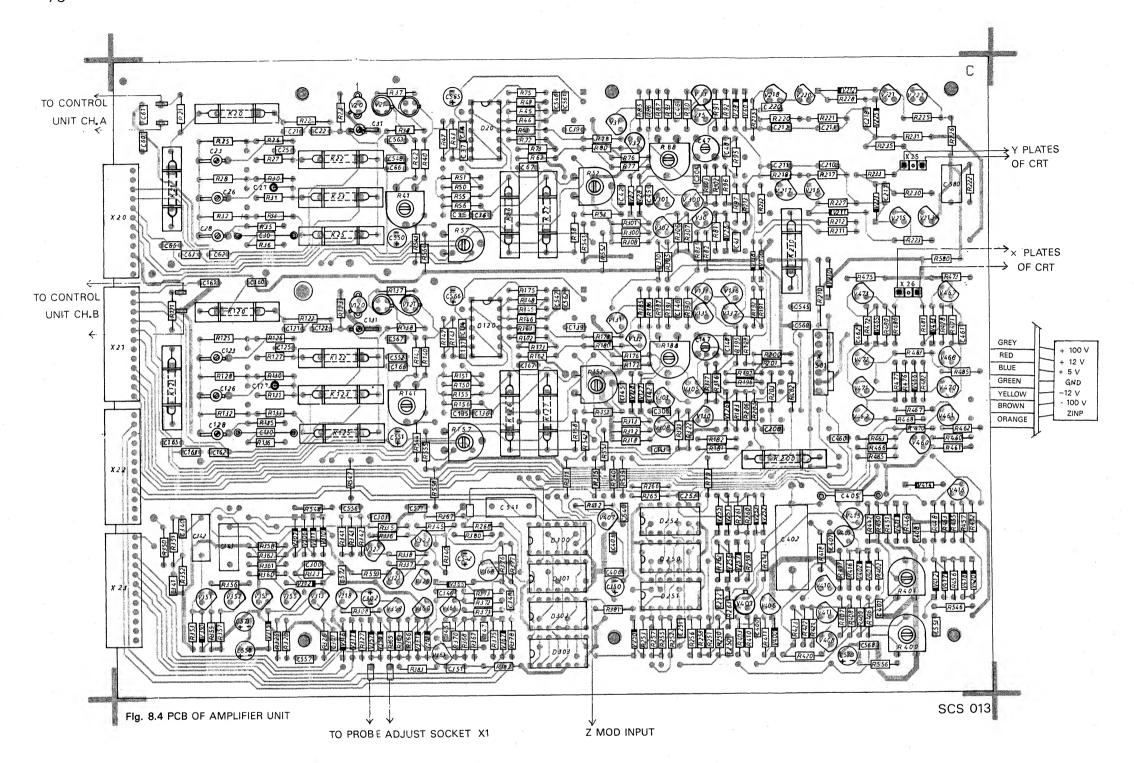
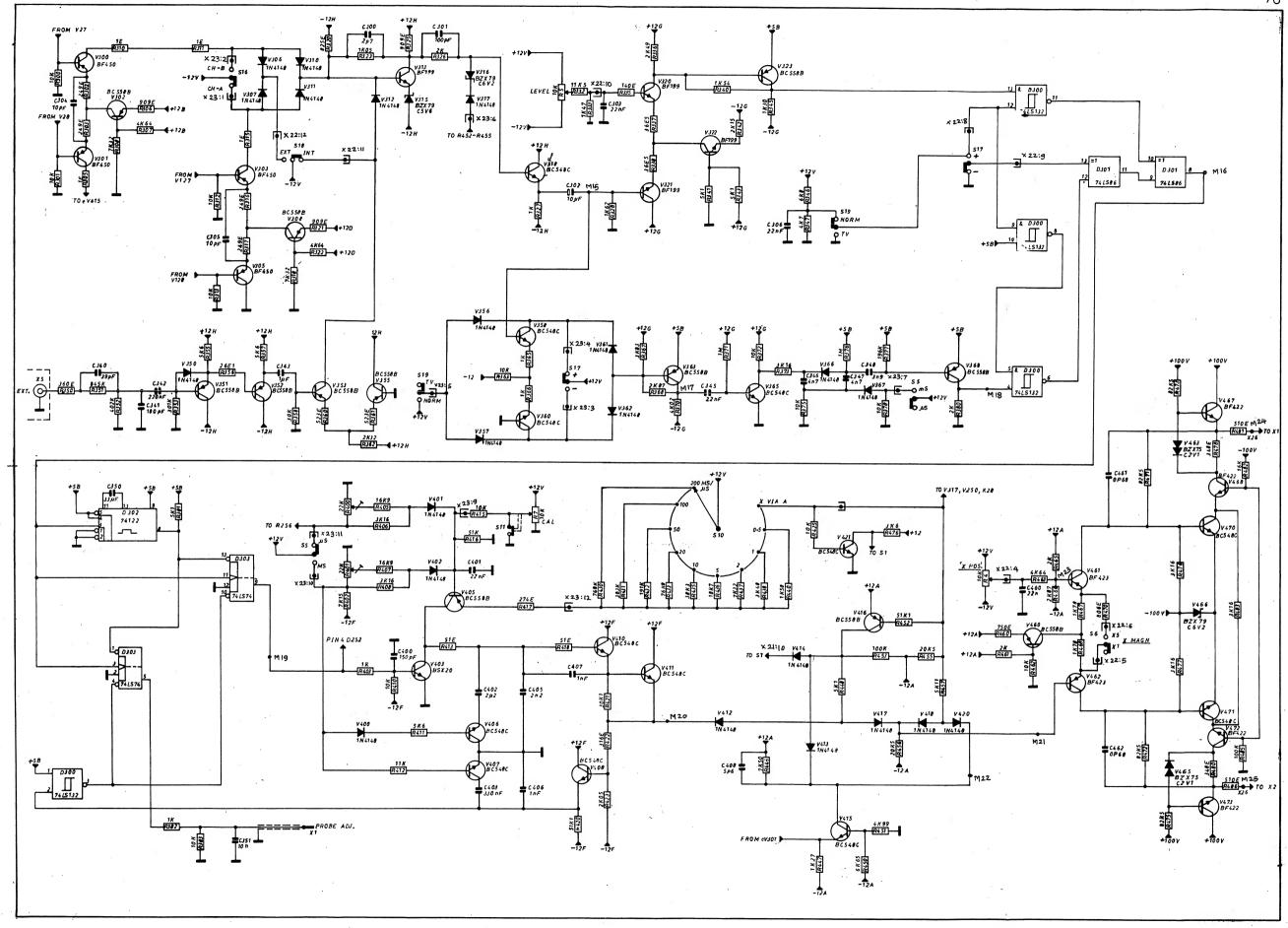


Fig. 8.3 CIRCUIT DIAGRAM VERTICAL CHANNEL







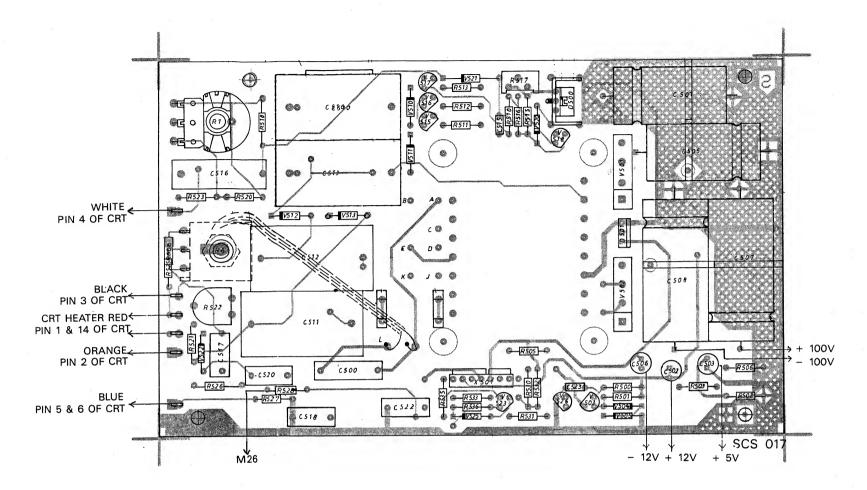
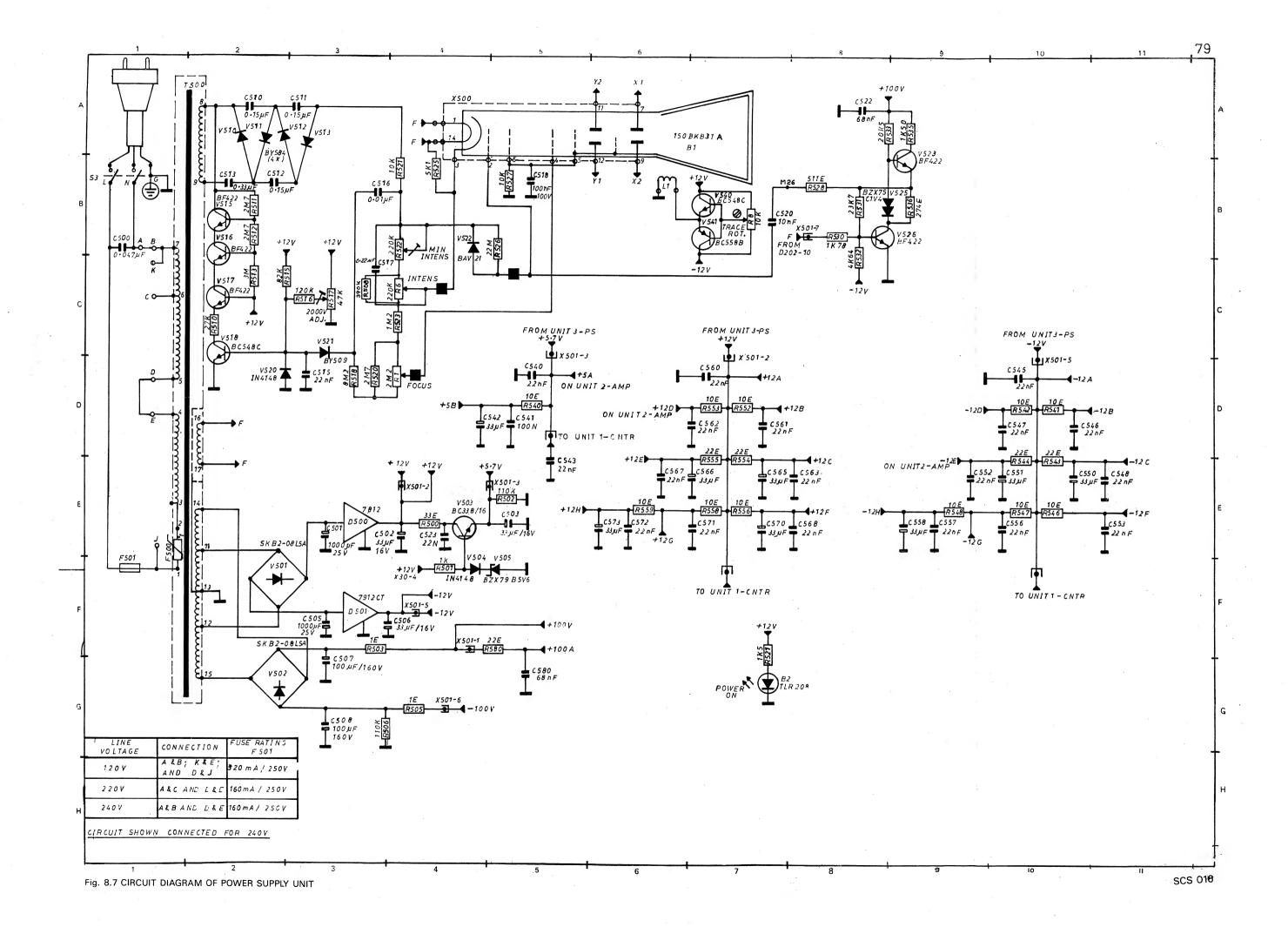


Fig. 8.6 PCB OF POWER SUPPLY UNIT



CODING SYSTEM OF FAILURE REPORTING FOR QUALITY ASSESSMENT OF T & M INSTRUMENTS

(excl. potentiometric recorders)

The information contents of the coded failure description is necessary for our computerized processing of quality data.

Since the reporting of repair and maintenance routines must be complete and exact, we give you an example of a correctly filled-out PHILIPS SERVICE Job sheet.

1	②	3		4			
Country	Day Month Year	Typenumber	/Version	Factory/Serial no.			
3 2	1 5 0 4 7 5	0 P M 3 2 6	0 0 2	D O 0 0 7 8 3			
	CODE	D FAILURE DESCRI	PTION	6			
(5)							
Nature d	of call Location	Component	t/sequence no. Ca	ategory			
—				1 0			
Installat		T S 0 6 R 0 0 6	3 1 2	Job completed			
Pre sale Preventi	· \		 				
mainten Correcti	ance 0 2	3 3 0 0		Working time (8)			
mainten		4		1 2 Hrs			
Other]			
Detailed des	cription of the informat	ion to be entered in th	ne various boxes:	·			
	3 2 = Switzerland	,, ,,, ,,, ,,,,,,,,,,,,,,,,,,,,,,,,,,					
②Day Month	n Year 1 5 0 4 7	5 = 15 April 1975					
গ্রType numl	ber/Version O P M	3 2 6 0 0 2 =	Oscilloscope PN	1 3260, version 02 (in later			
a) rype nam	bei/ version of i in	0 2 0 0 0 2	•	is number is placed in front of			
			the serial no)				
4)Factory/Se	erial number D O 0	0 7 8 3 = DO 78	33 These data are	mentioned on the type plate of			
(). uoto.,,,ot	57.ar 1.a.1.251 [2] 51.51	-111-1-1	the instrument	., .			
	call: Enter a cross in thure description	ne relevant box					
Location	,	Component/sequen	ice no.	Category .			
	7		_	I			
	J						
	oxes are used problem area.	These six boxes are pinpoint the faulty		0 Unknown, not applicable (fau not present, intermittent or	ΙŢ		
	de of the part	A. Enter the comp		disappeared)			
	fault occurs, e.g. unit	designation as used		1 Software error			
	nical item no	diagram. If the des	_	2 Readjustment			
	(refer to 'PARTS	alfa-numeric, the le		3 Electrical repair (wiring, solde	r		
LISTS' in th	e manual). 101 for Unit 1	written (starting fr		joint, etc.) 4 Mechanical repair (polishing,			
•	OA for Unit A	the figures must be		filing, remachining, etc.)			
	75 for item 75	such a way that th	e last digit	5 Replacement (of transistor,			
	not numbered, do not	occupies the right-r		resistor, etc.)			
	ur boxes; see Example	the four right-hand B. Parts not identi		6 Cleaning and/or lubrication			
Job sheet.		circuit diagram:	ried in the	7 Operator error 8 Missing items (on pre-sale test	1		
		990000 Unknown/	Not applicable	9 Environmental requirements a			
		990001 Cabinet or		not met			
			lem, grip, rail,	100			
		graticule, e 990002 Knob (inc					
		etc.)	, dia kilos, sep,				
		990003 Probe (on	y if attached				
•		to instrum		=			
		990004 Leads and					
		990005 Holder (va fuse, boar					
		990006 Complete					
		1	t. unit, etc.)	* *			
		990007 Accessory		1 1 2			
		1	ype number)				
		990008 Document supplement		,			
		990009 Foreign o					
		990099 Miscellane	•	I			

① Job completed: Enter a cross when the job has been completed.

® Working time: Enter the total number of working hours spent in connection with the job (excluding travelling, waiting time, etc.), using the last box for tenths of hours.

1	١	I	4	2	_	1 2	workin		/1	h	12	:-	١
					_	1.2	WOLKII	IQ HOL	11 S II	- 11	14	1111111.	

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